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EXPLOITING VARIABLE IMPEDANCE IN DOMAINS WITH
CONTACTS

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All magic comes with a price.

— Rumpelstiltskin

ABSTRACT

The control of complex robotic platforms is a challenging task, especially in designs with high levels of kinematic redundancy. Novel variable impedance actuators (VIAs) have recently demonstrated that, by allowing the ability to simultaneously modulate the output torque and impedance, one can achieve energetically more efficient and safer behaviour. However, this adds further levels of actuation redundancy, making planning and control of such systems even more complicated.

VIAs are designed with the ability to mechanically modulate impedance during movement. Recent work from our group, employing the optimal control (OC) formulation to generate impedance policies, has shown the potential benefit of VIAs in tasks requiring energy storage, natural dynamic exploitation and robustness against perturbation. These approaches were, however, restricted to systems with smooth, continuous dynamics, performing tasks over a predefined time horizon. When considering tasks involving multiple phases of movement, including switching dynamics with discrete state transitions (resulting from interactions with the environment), traditional approaches such as independent phase optimisation would result in a potentially suboptimal behaviour.

Our work addresses these issues by extending the OC formulation to a multi-phase scenario and incorporating temporal optimisation capabilities (for robotic systems with VIAs). Given a predefined switching sequence, the developed methodology computes the optimal torque and impedance profile, alongside the optimal switching times and total movement duration. The resultant solution minimises the control effort by exploiting the actuation redundancy and modulating the natural dynamics of the system to match those of the desired movement. We use a monopod hopper and a brachiation system in numerical simulations and a hardware implementation of the latter to demonstrate the effectiveness and robustness of our approach on a variety of dynamic tasks.

The performance of model-based control relies on the accuracy of the dynamics model. This can deteriorate significantly due to elements that cannot be fully captured by analytic dynamics functions and/or due to changes in the dynamics. To circumvent these issues, we improve the performance of the developed framework by incorporating an adaptive learning algorithm. This performs continuous data-driven adjustments to the dynamics model while re-planning optimal policies that reflect this adaptation. The results presented show that the augmented approach is able to handle a range of model discrepancies, in both simulation and hardware experiments using the developed robotic brachiation system.

LAY SUMMARY

Recent technological developments have contributed to the development of a wide range of robotic platforms. These modern systems are used in a wide range of applications and differ from their classical counterparts in terms of complexity and capabilities. One such capability is the modulation of impedance, using variable impedance actuators (VIAs). This capability brings improved performance in terms of energetically more efficient and safer behaviour. However, this adds further complexity, making planning and control of such systems even more complicated.

While previous studies addressed the control of systems with VIA capabilities, the results focused on systems with smooth, continuous dynamics, performing tasks over a predefined time horizon. When considering tasks involving multiple phases of movement, including switching dynamics with discrete state transitions (resulting from interactions with the environment, i.e., touch-down, lift-off), as required by the new generation of robotic platforms, traditional approaches would result in a potentially suboptimal behaviour.

In our work we address these issues by extending the OC formulation to a multi-phase scenario for robotic systems with VIAs. Given a predefined locomotion pattern, the developed methodology computes the optimal torque and impedance profile, alongside the optimal switching times and total movement duration. We use a monopod hopper and a brachiation system in numerical simulations and a hardware implementation of the latter to demonstrate the effectiveness and robustness of our approach on a variety of dynamic tasks.

The method devised is model-based, hence its performance relies on the accuracy of the dynamics model. This can deteriorate significantly due to elements that cannot be fully captured by standard mathematical models used and/or due to changes in the dynamics. To circumvent these issues, we improve the performance of the developed framework by incorporating an adaptive learning algorithm. This method is able to adjust the model of the system online, using newly collected data, which reflects the system's behaviour. The results presented show that the augmented approach is able to handle a range of model discrepancies, in both simulation and hardware experiments using the developed robotic brachiation platform.

DECLARATION

I declare that this thesis was composed by myself, that the work contained herein is my own except where explicitly stated otherwise in the text, and that this work has not been submitted for any other degree or professional qualification except as specified.

Edinburgh, 2016

Andreea Rădulescu

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PUBLICATIONS

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- Radulescu, A. and Howard, M. and Braun, D.J. and Vijayakumar, S. (2012). "Exploiting variable physical damping in rapid movement tasks." *Advanced Intelligent Mechatronics (AIM), 2012 IEEE/ASME International Conference on.*
- Nakanishi, J. and Radulescu, A. and Vijayakumar, S. (2013). "Spatio-temporal optimization of multi-phase movements: Dealing with contacts and switching dynamics." *Intelligent Robots and Systems (IROS), 2013 IEEE/RSJ International Conference on.*
- Radulescu, A. and Nakanishi, J. and Vijayakumar, S. (2016). "Optimal Control of Multi-Phase Movements with Learned Dynamics." *International Conference on Man-Machine Interactions (ICMMI), 2016 IEEE.*
- Nakanishi, J. and Radulescu, A. and Vijayakumar, S. (2016). "Spatio-temporal stiffness optimization with switching dynamics." *Autonomous Robots, 2016*

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INTRODUCTION

Robotics research has seen a lot of progress since the days of the first serial industrial manipulators. Modern robotic systems come in diverse configurations depending on their function. They are used in various fields ranging from the entertainment industry to health-care and operate in environments highly dangerous to humans (e.g., space and deep sea exploration, search and rescue missions).

As a consequence of this wide range of applications, complex designs have emerged (e.g., multi-fingered robotic arms or bipedal humanoid robots), which are characterised by a tree-type kinematic structure and multiple constraints. The control of these complex robotics platforms is a challenging task, due to the high levels of kinematic redundancy and the discontinuity in the dynamics, introduced by mechanical contact with the environment.

1.1 VARIABLE IMPEDANCE ACTUATORS

By comparison, biological systems can achieve locomotion and manipulation tasks with significant ease. They exert control over their limbs with notable versatility, compliance and energy efficiency. This is achieved despite significant levels of delay and noise affecting the biological motor systems [Faisal et al., 2008]. Such efficiency is due in part to the considerable utilisation of joint impedance modulation. This is mainly realised by co-contracting antagonistic muscle pairs in a manner adapted to the task at hand [Mitrovic et al., 2010a]. Various studies indicate that humans make use of impedance control in counteracting the effect of destabilising external forces in early stages of dynamics learning and as an accuracy increasing technique in general [Burdet et al., 2001; Franklin et al., 2007]. Biological systems use motions that exploit, rather than overcome, the underlying non-linearities of the physical systems.

The presence of compliant elements in the structure of robotic systems was investigated in [Sweet and Good, 1985], for an industrial robotic arm equipped with flexible transmission. In an previous study [McCallion et al., 1979], an end-effector design with passive compliance (implemented using a set of springs) was intro-

duced, in order to efficiently solve a "peg-in-the-hole" alignment task for assembly. Subsequently, various designs of industrial robots with inbuilt compliance (e.g., compliant, position adaptive wrist) were developed and used in assembly tasks, due to their ability to handle small misalignments between parts [Cho et al., 1987]. The potential benefits of impedance modulation for a variety of scenarios (e.g., prosthetics, tool manipulation) were suggested as early as [Hogan, 1984].

Inspired by these studies and by the aforementioned capabilities of biological systems, the robotics community has recently developed a new generation of actuators equipped with an additional mechanically adjustable compliant mechanism [Groothuis et al., 2012; Petit et al., 2010; Schiavi et al., 2008; Visser et al., 2011; Vitiello et al., 2008]. These variable impedance actuators (VIAs) can provide simultaneous modulation of impedance and output torque, with the purpose of achieving dynamic and flexible robotic movements. However, this adds further levels of actuation redundancy, making planning and control of such systems even more complicated.

The benefits of impedance modulation have come under investigation in several studies. The duality principle introduced in [Anderson and Spong, 1988], stating that a good rule would be for the robotics system to produce an impedance inversely proportional to that of the environment, was successfully used for simple tracking tasks [Hogan, 1984]. Such simple presets are viable solutions only in specific cases, while in a more general set-up the robot has to adapt to changes in task and the structure of the world.

This can only be achieved by continuous and fast modulation of impedance. Recent work, employing the optimal control (OC) formulation to generate impedance policies, has shown the potential advantage of VIAs in terms of noise rejection [Mitrovic et al., 2009] and energy efficiency in highly dynamic, explosive tasks [Braun et al., 2012b; Howard et al., 2010]. The benefits of their use include high dynamic range (e.g., due to the ability to store energy in spring-like actuators) [Braun et al., 2012b] and a stable and fast response (since compliance is built into the actuator mechanically, sensory feedback is not as crucial in responding to perturbations). The results are consistent with observed patterns displayed by humans and other biological systems in adapting their impedance under similar conditions [Mitrovic et al., 2010c].

1.2 SYSTEMS IN DOMAINS WITH CONTACTS

These studies were based on traditional OC approaches and thus, focused on systems with smooth, continuous dynamics, performing tasks over a predefined time horizon. When considering tasks involving multiple phases of movement, including switching dynamics with discrete state transitions (resulting from interactions with the environment), traditional OC approaches such as independent phase optimisation would result in a potentially suboptimal behaviour. In order to cope with discontinuities introduced by varying contacts, several modifications, such as the hybrid dynamics formulation, were constructed. These methods have proven feasible in a variety of tasks involving contacts [Bätz et al., 2010; Grizzle et al., 2001; Long et al., 2011; Rosa Jr. et al., 2012].

Several studies of impedance modulation in the context of domains with contacts showed that actuators equipped with mechanical compliance provide a significant *energy efficiency* improvement, due to their energy storage capabilities. In [Stramigioli et al., 2008], the energy input requirements of the motors is reduced by exploiting the compliance offered by the mechanical design. In [Vanderborght et al., 2006], the authors employ a compliance actuator and control it to modulate the natural dynamics of the system (to match that of the desired trajectory). Combined with trajectory tracking, the approach significantly lowered the required control effort. Though the study focused on an isolated robotic joint, the results demonstrate how an optimised impedance modulation could improve the performance of locomotion systems. In [Nakanishi et al., 2011] stiffness and temporal optimisation is used at individual phase level, for periodic movements with an emphasis on exploiting the intrinsic dynamics of the system for control efficiency.

The use of impedance modulation in scenarios involving interaction with the environment was shown to provide several *safety* benefits [Van Damme et al., 2009]. In [Goris et al., 2011], a robot with variable stiffness actuation is designed specifically for safe human-robot interaction. The "peg-in-the-hole" problem from [McCallion et al., 1979] was revisited in [Tsumugiwa et al., 2002] and approached from a human-robot cooperation perspective, where virtual stiffness control is used to facilitate a successful task execution with increased precision.

Other advantages of variable impedance capabilities were observed in terms of *robustness* and *adaptability*. These are often required by tasks involving unpredictable changes in the environment and noise [Catalano et al., 2011; Yang et al., 2011]. The need for *adaptability* and *force accuracy* becomes crucial in rehabilita-

tion and assistive technologies, such as exoskeletons [Veneman et al., 2007] and prosthetics [Blaya et al., 2004].

Our work builds upon these prior efforts by extending the OC formulation to a multi-phase scenario and incorporating temporal optimisation capabilities (for robotic systems with VIAs). For a predefined switching sequence associated with the task, the developed method outputs the optimal torque and impedance profile, alongside the optimised temporal aspect (i.e., optimal switching times and total movement duration). The obtained solution is characterised by minimal control effort, realised by exploiting the actuation redundancy and modulating the natural dynamics of the system to match those of the desired movement.

The use of continuously more advanced variable VIAs and control techniques may lead to a new generation of robotic systems capable of close interaction with humans. Their increased manipulation and locomotion performances could approach those of the biological systems from which they were originally inspired.

BRACHIATION SYSTEMS Our main platform, used for demonstrating the benefits our developed methodology in both simulation and real hardware implementation, is a two-link brachiation system.

Brachiation is a type of locomotion, encountered in biological systems (e.g., primates, mainly gibbons), where movement is achieved using the arms by swinging from one handhold to another [D'Août and Vereecke, 2011]. The use of muscle activation and impedance modulation is crucial for the successful execution of these tasks. Various studies [Michilsens et al., 2009; Usherwood and Bertram, 2003] have shown how active control of the muscle activation (and thus, modulation of impedance) is used in these cases, in order to reach the desired target for this arboreal locomotion.

The control of robotic brachiation systems has been the topic of numerous studies (Sec. 4.3) [Gomes and Ruina, 2005; Kajima et al., 2004; Nakanishi et al., 2011; Rosa Jr. et al., 2012; Saito et al., 1994]. The utilisation of brachiation in practical robotics applications was demonstrated for a wide range of tasks, such as bridge inspection [Mazumdar and Asada, 2010] or maintenance of power and transmission lines [Rocha and Sequeira, 2004; Toussaint et al., 2009].

In our works, we focus on brachiation as locomotion task where the successful reach of the desired handhold configuration is highly dependent on the efficient use of the impedance modulation capabilities and exploitation of the natural dynamics of the system.

1.3 ADAPTIVE DYNAMICS LEARNING

The performance of model-based control is highly dependent on the accuracy of the dynamics models employed, which are traditionally obtained from mechanical engineering insights. However, certain elements cannot be fully captured by analytic dynamics functions. Examples include friction from the joints or resulting from cable movement [Siciliano and Khatib, 2008], which can vary in time. The presence of flexible elements (e.g., springs used in VIAs designs) increases the model complexity and the difficulty of model identification. Additionally, changes in the behaviour of the system can occur due to wear and tear or due to the use of a tool (thus modifying the mechanical chain structure) [Sigaud et al., 2011].

On-line adaptation of models provides a solution for capturing all these properties. Early approaches, such as on-line parameter identification [Sciavicco and Villani, 2009], which tunes the parameters of a predefined model (dictated by the mechanical structure) using data collected during operation, proved sufficient for accurate control and remained a popular approach for a long time [Atkeson et al., 1986; Khalil and Dombre, 2004]. The increased complexity of recent robotic systems demands novel approaches capable of accommodating significantly non-linear and unmodelled robot dynamics. Successful non-parametric model learning methods use supervised learning to perform system identification with only limited prior information about its structure. By removing the restriction to a fixed model structure, the model complexity can adapt in a data driven manner.

We improve the performance of the developed framework by incorporating an adaptive learning algorithm. This performs continuous data-driven adjustments to the dynamics model while re-planning optimal policies that reflect this adaptation. By engaging these technique in the context of multiphase variable impedance movements we build on prior efforts to employ adaptive dynamics learning in improving the performance of robot control [Mitrovic et al., 2010a; Nakanishi et al., 2005; Nguyen-Tuong and Peters, 2011]. We present results showing that the augmented approach is able to handle a range of model discrepancies, in both simulation and hardware experiments using the developed brachiation system.

1.4 THESIS OUTLINE

In this opening chapter, we introduced the motivations and goals for our work, and presented the corresponding setting for the developed methodology we employ throughout the thesis.

In Chapter 2, we outline the relevant background on OC theory. The main formulation of an OC problem and the principal approaches for solving it are presented, with a focus on methods well suited for systems with multiple degrees of freedom. We present the Iterative Linear Quadratic Gaussian (iLQG) method, upon which our approach is built. Finally, we give a few brief remarks regarding practical issues and we outline the changes presented in later chapters.

In Chapter 3, we start by introducing the concept of VIAs, actuators that allow for simultaneous modulation of the joint position and impedance. We show the difficulty of planning and control for systems with such capabilities. We provide a summary of their benefits, as indicated by previous approaches. We follow with our study into the little researched sub-field of variable damping (alongside stiffness modulation) in the context of OC. Though they are redundant elements from a mechanical perspective, we show that there is a clear advantage of having variable damping capabilities. The amplitude of this effect is dependent on the particularities of the hardware embodiment and the task at hand.

Original contributions:

- *First time use of simultaneous and independent stiffness and damping modulations in a single actuator design.*
- *A demonstration of the benefits of exploiting the variable physical damping through OC in the context of rapid movement tasks.*

In Chapter 4, we provide an overview of the theoretical background regarding modelling and control of systems in domains with contacts. We indicate currently uninvestigated aspects and present the developed approach addressing some of the issues highlighted. We provide detailed investigations emphasising the benefits of temporal and impedance modulation optimisation. We present both numerical evaluations and experimental results, for validation in a physical scenario, of brachiation tasks in a two-link brachiating robot with a variable stiffness actuator (VSA) developed in our laboratory.

Original contributions:

- *Optimisation framework for multi-phase movement tasks featuring switching dynamics and contacts with the environment for systems with variable impedance capabilities.*
- *Showcasing the benefits of temporal and impedance modulation optimisation.*
- *Real hardware validation of the proposed approach, alongside the corresponding numerical results from simulation experiments.*

In Chapter 5, we extend the evaluation of our approach to a monopod hopper. We discuss the additional difficulties introduced by the increasingly complex dynamics of this platform. We conclude with a set of simulation experiments demonstrating the capabilities of this framework on numerical simulations of a monopod hopper system, equipped with VSAs.

Original contributions:

- *Numerical simulations demonstrating the success of the proposed approach on a platform with increased complexity of the dynamics (i.e., a monopod hopper with VIA capabilities).*
- *A demonstration of the robustness against perturbations of the optimal feedback controller for a continuous hopping task.*

In Chapter 6 we augment our approach by addressing the challenging problem of model learning for control. We employ a (locally weighted non-parametric) adaptive learning method to model the discrepancies in the system dynamics. This allows for continuous adjustments to the dynamics model, while re-planning policies that reflect this adaptation. We present results showcasing the performance of the augmented approach on a range of model discrepancies in both simulations and hardware experiments using the developed brachiation system.

Original contributions:

- *Improved performance of the developed multi-phase optimisation framework by incorporating an adaptive learning approach.*
- *Real hardware validation of the extended methodology on the developed brachiation robot with VSA capabilities.*

In Chapter 7 we summarise the presented work and suggest possible research directions for the future.

PUBLICATION SUMMARY

The results detailed in Chapter 2 were published in [Radulescu et al., 2012]. Results presented in Chapters 3 and 4 were included in [Nakanishi et al., 2013]. Part of the work in Chapter 6 was published in [Radulescu et al., 2016]. Finally, a publication based on Chapters 3 and 4 has recently been accepted for publication.

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- Radulescu, A. and Nakanishi, J. and Vijayakumar, S. (2016). "Optimal Control of Multi-Phase Movements with Learned Dynamics." *International Conference on Man-Machine Interactions (ICMMI), 2016 IEEE*.
- Nakanishi, J. and Radulescu, A. and Vijayakumar, S. (2016). "Spatio-temporal stiffness optimization with switching dynamics." *Autonomous Robots, 2016:1-19*

OPTIMAL CONTROL

In this chapter we discuss the relevant background on optimal control (OC) theory. We first state the basic formulation of an OC problem and provide a brief overview of the main approaches for solving it. We focus on methods well suited for systems with multiple degrees of freedom. Next, we present our method of choice (iLQG), upon which our approach is built. At the end, we give a few brief remarks regarding practical issues and we outline the modifications presented in later chapters.

2.1 INTRODUCTION

Optimal control (OC) is concerned with finding a control law and the corresponding state trajectories for a given system, such that a chosen optimality criterion is satisfied. In order to devise the control logic, a mathematical model of the dynamical system (*system dynamics*) and a performance index (*cost function*) are required.

Let $\mathbf{x}(t) \in \mathbb{R}^n$ denote the state of a plant (consisting of joint angles \mathbf{q} and velocities $\dot{\mathbf{q}}$) and the control signal $\mathbf{u}(t) \in \mathbb{R}^m$ applied at time t . The state-space representation of the controlled system dynamics in the absence of discontinuities is expressed as:

$$d\mathbf{x} = \mathcal{F}(\mathbf{x}, \mathbf{u})dt + \mathcal{G}(\mathbf{x}, \mathbf{u})d\omega \in \mathbb{R}^n, \quad (2.1)$$

where \mathcal{F} represents the dynamics function, $\omega \in \mathbb{R}^w$ is assumed to be the Brownian motion noise, which is transformed by a possibly state- and control-dependent matrix $\mathcal{G}(\mathbf{x}, \mathbf{u})$ (*diffusion coefficient*). For fully deterministic systems, we consider $\mathcal{G}(\mathbf{x}, \mathbf{u}) = 0$ and reduce the dynamics to $\dot{\mathbf{x}} = \mathcal{F}(\mathbf{x}, \mathbf{u})dt$.

An OC problem can accept constraints on the values of the control variables (e.g. imposed by the limitations of the actuators). In this case, the control inputs $\mathbf{u} \in \mathbb{R}^m$ are drawn from the set of admissible controls:

$$\mathbf{u} \in [\mathbf{u}_{\min}; \mathbf{u}_{\max}], \quad (2.2)$$

where \mathbf{u}_{\min} and \mathbf{u}_{\max} are vectors expressing the lower and the upper bounds on \mathbf{u} , respectively.

A cost functional encodes the criteria for optimisation. This is a function of the state and control variables, which can be expressed for a finite horizon $t \in [0, T]$ task as:

$$J = \rho(\mathbf{x}(T)) + \int_0^T r(\mathbf{x}(t), \mathbf{u}(t), t) dt, \quad (2.3)$$

where $\rho(\cdot)$ and $r(\cdot)$ are the terminal and running costs, respectively. These are selected according to the application to encode the goals of the task.

An optimal control is a set of differential equations describing the paths of the control variables that minimise the cost functional. The general stochastic optimal control problem is then, given the initial state $\mathbf{x}(0) = \mathbf{x}_0$ and desired final state $\mathbf{x}(T) = \mathbf{x}_T$, finding:

$$\mathbf{u}^* = \arg \min_{\mathbf{u}} (J), \quad (2.4)$$

while satisfying the defined dynamics equations (2.1) and control constraints (2.2). In the case of stochastic systems, the minimisation is performed on the expected cost (on both the running and the final costs). This OC problem can be formulated for both discrete and continuous domains [Todorov, 2006].

The solution can be provided in either open-loop control form, where the output control law \mathbf{u} is independent of the state variables, or in feedback control form, where \mathbf{u} is a function of the current state.

2.2 OPTIMAL CONTROL THEORY

Next, we give a brief overview of relevant findings in the area of OC. For further details we refer the interested reader to some of the papers and well known reference books [Bryson and Ho, 1975; Dyer and McReynolds, 1970; Stengel, 1994].

There are two different ways of deriving the optimal control: either by solving the Hamilton-Jacobi-Bellman (HJB) equation [Stengel, 1994] (also called Bellman's optimality principle), or by using Pontryagin's Minimum Principle (PMP) [Ross, 2009].

Bellman's optimality principle introduces the concept of value function:

$$\mathbf{V}(\mathbf{x}(\sigma), \sigma) = \rho(\mathbf{x}(T)) + \int_{\sigma}^T r(\mathbf{x}(t), \mathbf{u}(t), t) dt, \quad (2.5)$$

which is the cost that will result from initialising the system in state $\mathbf{x}(\sigma)$ at time σ and following the control law \mathbf{u} . The principle states that optimal trajectories remain optimal for intermediate points in time. For the evolution of the system from t to $t + dt$, this is formulated as:

$$\mathbf{V}(\mathbf{x}(t), t) = \min_{\mathbf{u}} \{r(\mathbf{x}(t), \mathbf{u}(t), t)dt + \mathbf{V}(\mathbf{x}(t + dt), t + dt)\}. \quad (2.6)$$

The last term in (2.6) can be approximated using the first order Taylor expansion as:

$$\mathbf{V}(\mathbf{x}(t + dt), t + dt) \approx \mathbf{V}(\mathbf{x}(t), t) + \dot{\mathbf{V}}(\mathbf{x}(t), t)dt + \nabla_{\mathbf{x}}\mathbf{V}(\mathbf{x}(t), t)\dot{\mathbf{x}}(t)dt, \quad (2.7)$$

where ∇ is the gradient operator.

For the given formulation of the problem, we form the HJB equation (or simply Bellman equation for the discrete case) by replacing (2.7) inside (2.6):

$$\dot{\mathbf{V}}(\mathbf{x}(t), t) + \min_{\mathbf{u}} \{\nabla_{\mathbf{x}}\mathbf{V}(\mathbf{x}(t), t) \cdot \mathcal{F}(\mathbf{x}(t), \mathbf{u}(t)) + r(\mathbf{x}(t), \mathbf{u}(t), t)\} = 0, \quad (2.8)$$

while setting the terminal condition:

$$\mathbf{V}(\mathbf{x}(T), T) = \rho(\mathbf{x}(T)). \quad (2.9)$$

The HJB formulation provides sufficient conditions for optimality; the solutions can be obtained by either value or policy iteration. Bellman's optimality principle renders a global solution, which can include a globally valid feedback control law for stochastic scenarios [Bellman and Kalaba, 1965].

Solving the HJB equation involves a discretisation of the state-action space, which can be non-trivial in practice. A too-sparse discretisation will result in poor performance (due to poor representation of the actual state-action space). A too fine one would lead to an exponential increase of the computational requirements (for large degrees of freedom (DOFs) and complex systems). This issue has been referred to by Bellman as "the curse of dimensionality".

To bypass the curse of dimensionality global methods rely on function approximation, but this reduces the chances of convergence. Given the large dimension-

ality of the state space of robotic systems, local methods, based on PMP, have traditionally been used.

The central idea behind the minimum principle states that the optimal trajectory will have neighbouring solutions that do not result in a smaller cost. Thus, any derivative of the cost function, taken along the optimal trajectory, should be zero.

Given the standard OC problem, the minimum principle is typically written in a compact form using the Hamiltonian function:

$$H(\mathbf{x}(t), \mathbf{u}(t), \mathbf{p}(t), t) := r(\mathbf{x}(t), \mathbf{u}(t), t) + \mathbf{p}(t)^\top \mathcal{F}(\mathbf{x}(t), \mathbf{u}(t)), \quad (2.10)$$

where $\mathbf{p}(t)$ is the *costate* (gradient of the optimal cost-to-go function), with $\mathbf{p}(T) = \frac{\partial}{\partial \mathbf{x}} \rho(\mathbf{x}(T))$. This allows the cost function in (2.3) to be reformulated as:

$$J = \rho(\mathbf{x}(T)) + \int_0^T [H(\mathbf{x}(t), \mathbf{u}(t), \mathbf{p}(t), t) - \mathbf{p}(t)^\top \mathcal{F}(\mathbf{x}(t), \mathbf{u}(t))] dt. \quad (2.11)$$

If $\mathbf{x}^*(t)$ and $\mathbf{u}^*(t)$ with $t \in [0, T]$ represent the optimal solution for the given problem then, according to the minimum principle, the following conditions must be met:

$$\begin{aligned} \dot{\mathbf{x}}(t) &= \left. \frac{\partial}{\partial \mathbf{p}} H \right|_{(\mathbf{x}=\mathbf{x}^*, \mathbf{u}=\mathbf{u}^*)} \\ \dot{\mathbf{p}}(t) &= - \left. \frac{\partial}{\partial \mathbf{x}} H \right|_{(\mathbf{x}=\mathbf{x}^*, \mathbf{u}=\mathbf{u}^*)} \\ 0 &= \left. \frac{\partial}{\partial \mathbf{u}} H \right|_{(\mathbf{x}=\mathbf{x}^*, \mathbf{u}=\mathbf{u}^*)} \end{aligned} \quad (2.12)$$

The minimum principle supplies necessary conditions for optimality; it refers only to quantities along a specific trajectory. Methods based on the minimum principle are called local, trajectory based or open-loop methods. Finding the solution for the equations in (2.12) can be achieved via numerical methods such as relaxation (solving a boundary value problem), gradient descent or shooting [Polak, 2012]. Due to the local nature of the solutions, which provide only open-loop control laws, they do not generalise easily to stochastic problems.

An important milestone in OC theory was the *Linear Quadratic Regulator (LQR)*, formulated by [Kalman et al., 1960], which provides *linear feedback of the state variables* for a system with linear dynamics ($\mathcal{F}(\mathbf{x}, \mathbf{u}) = \mathbf{Ax} + \mathbf{Bu}$) and quadratic performance index (in both \mathbf{x} and \mathbf{u}).

The LQ problem is described by:

$$\begin{aligned}
\text{system dynamics} & : \dot{\mathbf{x}} = (\mathbf{A}\mathbf{x} + \mathbf{B}\mathbf{u})dt, \\
\text{running cost} & : r(\mathbf{x}, \mathbf{u}) = \frac{1}{2}\mathbf{u}^\top \mathbf{R}\mathbf{u} + \frac{1}{2}\mathbf{x}^\top \mathbf{Q}\mathbf{x}, \\
\text{final cost} & : \rho(\mathbf{x}) = \frac{1}{2}\mathbf{x}^\top \mathbf{Q}^f \mathbf{x},
\end{aligned} \tag{2.13}$$

where \mathbf{R} is symmetric positive definite, \mathbf{Q} and \mathbf{Q}^f are positive semi-definite and \mathbf{u} is unconstrained. The corresponding Hamiltonian ¹:

$$H = \mathbf{u}^\top \mathbf{R}\mathbf{u} + \mathbf{x}^\top \mathbf{Q}\mathbf{x} + \mathbf{p}(t)^\top (\mathbf{A}\mathbf{x} + \mathbf{B}\mathbf{u}), \tag{2.14}$$

is minimised w.r.t \mathbf{u} , with $\mathbf{V}(t) = \mathbf{p}(t)\mathbf{x}(t)$. This involves solving the Riccati differential equation:

$$-\dot{\mathbf{V}}(t) = \mathbf{A}^\top \mathbf{V}(t) + \mathbf{V}(t)\mathbf{A} - \mathbf{V}(t)\mathbf{B}\mathbf{R}^{-1}\mathbf{B}^\top \mathbf{V}(t) + \mathbf{Q}, \tag{2.15}$$

by initialising it with $\mathbf{V}(T) = \mathbf{Q}^f$ and integrating the ordinary differential equation (ODE) backwards in time. The global optimal control law is given by:

$$\mathbf{u} = -\mathbf{R}^{-1}\mathbf{B}^\top \mathbf{V}(t). \tag{2.16}$$

This approach was later extended to the case of Gaussian noise, under the name of *Linear Quadratic Gaussian (LQG) compensator* [Athans, 1971]. The benefits of the LQ approach are: (i) it provides an analytic form of the solution via the Riccati equations and (ii) it also offers a globally valid feedback control law.

In the case of systems that do not conform to the LQ framework, solving the HJB equations requires a discretisation of the state-action space, which is not a simple task to achieve for realistic situations (i.e., continuous state-action space). Several methods to avoid this problem, by using sampling techniques, have been explored successfully for the case of smooth enough dynamics and cost functions [Atkeson, 2007].

Another approach for avoiding the curse of dimensionality is to contract the space of interest to a narrow region in the vicinity of a nominal optimal trajectory. In the vicinity of such trajectories, the problem can be approximated using the LQ methodology and solved using (2.15) and (2.16). The final solution is obtained by computing the optimal trajectory with the corresponding feedback law and

¹ The factor 1/2 is omitted in order to simplify the derivation.

iteratively improving this solution until convergence. A well know example of such an approach is the *Iterative Linear Quadratic Gaussian (iLQG)* method [Todorov and Li, 2005]. This algorithm serves as the technique of choice in the development of our framework. It will be explained in detail in the following section, after a brief overview of related iterative methods.

2.3 ITERATIVE OPTIMAL CONTROL METHODS

As mentioned in the preceding sections, most numerical methods for OC can be separated into two distinct categories. Global methods, based on the HJB equation, provide a globally valid feedback control law (and, thus, can cope with stochastic dynamics), but they do not scale well to higher dimensions due to the curse of dimensionality. Local methods, based on PMP, avoid this by solving a set of ODE, but they only return a locally optimal open-loop control law (therefore, they cannot handle unexpected changes during execution).

Early approaches in optimal motor control focused on the open-loop formulation [Li and Todorov, 2004]. Modern robotic set-ups often feature changing environments and robot-human interaction, which can introduce noise and delays. Biological systems control their limbs with versatility, compliance and energy efficiency, in the presence of significant levels of delay and noise [Faisal et al., 2008], which is only possible due to an intricate feedback control scheme. This inspired a shift towards optimal feedback control methods [Li and Todorov, 2004] in OC for robotic systems. Since such systems can have a large number of DOFs, global methods do not scale easily to their case.

A good compromise between local and global optimisation methods is obtained by algorithms that compute an optimal trajectory together with a locally valid feedback law. Solving an ODE with boundary conditions (resulting from the minimum principle) can be performed explicitly for specific cases (e.g., for an LQ problem). Most numerical computing environments (e.g., MatLab, Octave) provide solvers for such problems.

Coordinate Gradient Descent (CGD) uses PMP to compute the gradient of the total cost wrt. the nominal control sequence and performs an optimised conjugate gradient descent to solve the posed problem. Coordinate descent is based on the idea that the minimisation of a multi-variable cost function can be achieved by minimising it along one direction at a time.

Differential Dynamic Programming (DDP) [Dyer and McReynolds, 1970; Jacobson and Mayne, 1970] performs regular dynamic programming in the vicinity of the nominal trajectory using a second order approximation. Solutions are obtained by iteratively improving nominal trajectories, each iteration performing a sequence of backward and forward sweeps in time. In the backward sweep a control law is obtained by approximating the time-dependent value function along the current nominal trajectory. In the forward sweep an improved nominal control sequence is computed. These iterations are repeated until the cost converges.

For a more detailed presentation and comparison of these methods we refer the interested reader to the study in [Todorov and Li, 2003].

2.3.1 The iLQG Method

The iLQR (*Iterative Linear Quadratic Regulator*) framework first introduced in [Li and Todorov, 2004] and generalised as the iLQG (*Iterative Linear Quadratic Gaussian*) method in [Todorov and Li, 2005] is a fast approximate solver of optimal control problems, closely related to the DDP approach. The method is based on approximating the optimal control problem as linear-quadratic and performing iterative improvements of the solutions around a nominal trajectory.

Compared with previously existing algorithms (CGD, DDP), the iLQG method was shown to provide faster convergence. Combined with multiple initialisations, to overcome local minimal issues, it was able to converge to a better solution in terms of computation time and cost minimisation [Li and Todorov, 2004].

This section will present the details of the iLQG algorithm, in the form we used as starting point for our development. We consider the standard OC task as described by (2.1) : (2.4). The finite horizon allocated is discretised as $t = (p - 1)\Delta t$ and $t \in [0, T]$ with $p \in [1 : P]$ are the discretisation steps and Δt is the time step (simulation rate).

The algorithm starts with the initial state \mathbf{x}_0 and an user-supplied nominal initial control sequence $\bar{\mathbf{u}}_p := \bar{\mathbf{u}}((p - 1)\Delta t)$. The corresponding state trajectory $\bar{\mathbf{x}}_p$ is generated by running the control sequence through the deterministic forward dynamics:

$$\bar{\mathbf{x}}_{p+1} = \bar{\mathbf{x}}_p + \Delta t \mathcal{F}(\bar{\mathbf{x}}_p, \bar{\mathbf{u}}_p). \quad (2.17)$$

The dynamics (2.1) are approximated locally using a linear model around the nominal trajectory $(\bar{\mathbf{x}}, \bar{\mathbf{u}})$:

$$\delta \mathbf{x}_{p+1} = (\mathcal{F}_x)_p \delta \mathbf{x}_p + (\mathcal{F}_u)_p \delta \mathbf{u}_p + \mathbf{C}_p(\delta \mathbf{u}_p) \boldsymbol{\xi}_p, \quad (2.18)$$

where $\delta \mathbf{x}_p := \mathbf{x}_p - \bar{\mathbf{x}}_p$, $\delta \mathbf{u}_p := \mathbf{u}_p - \bar{\mathbf{u}}_p$, $\boldsymbol{\xi}_p \sim \mathcal{N}(0, I_w)$ and:

$$(\mathcal{F}_x)_p = \mathbf{I} + \Delta t \left. \frac{\partial \mathcal{F}}{\partial \mathbf{x}} \right|_{\bar{\mathbf{x}}_p}, \quad (\mathcal{F}_u)_p = \Delta t \left. \frac{\partial \mathcal{F}}{\partial \mathbf{u}} \right|_{\bar{\mathbf{u}}_p}, \quad (2.19)$$

$$\begin{aligned} \mathbf{C}_p(\delta \mathbf{u}_p) &= [\mathbf{c}_{1,p} + \mathbf{C}_{1,p} \delta \mathbf{u}_p, \dots, \mathbf{c}_{w,p} + \mathbf{C}_{w,p} \delta \mathbf{u}_p], \\ \mathbf{c}_{i,p} &= \sqrt{\Delta t} \mathcal{G}^{[i]}, \quad \mathbf{C}_{i,p} = \sqrt{\Delta t} \frac{\partial \mathcal{G}^{[i]}}{\partial \mathbf{u}}, \end{aligned} \quad (2.20)$$

where $\mathcal{G}^{[i]}$ is the i -th column of the matrix \mathcal{G} (as defined in (2.1)). The last term in (2.18) refers to the case when the dynamics are affected by control dependent noise. Since the covariance of Brownian motion grows linearly with time, the standard deviation of the discrete-time noise scales with $\sqrt{\Delta t}$.

Similarly the cost is approximated as quadratic:

$$\begin{aligned} \delta J &= \rho_x^T \delta \mathbf{x}_P + \delta \mathbf{x}_P^T \rho_{xx} \delta \mathbf{x}_P + \sum_{p=1}^P (\mathbf{r}_x^T \delta \mathbf{x} + \mathbf{r}_u^T \delta \mathbf{u}) \\ &+ \sum_{p=1}^P (\delta \mathbf{x}^T \mathbf{r}_{xx} \delta \mathbf{x} + \delta \mathbf{x}^T \mathbf{r}_{xu} \delta \mathbf{u} + \delta \mathbf{u}^T \mathbf{r}_{uu} \delta \mathbf{u}), \end{aligned} \quad (2.21)$$

with the cost-to-go:

$$\begin{aligned} v_p(\delta \mathbf{x}_p) &= \Delta t (r_p) + \delta \mathbf{x}_p^T (r_x) + \frac{1}{2} \delta \mathbf{x}_p^T (r_{xx}) \delta \mathbf{x}_p + \\ &\delta \mathbf{u}_p^T (r_u) + \frac{1}{2} \delta \mathbf{u}_p^T (r_{uu}) \delta \mathbf{u}_p + \delta \mathbf{u}_p^T (r_{xu}) \delta \mathbf{x}_p. \end{aligned} \quad (2.22)$$

Here the subscript notations \mathbf{x} and \mathbf{u} denote partial differentiation of the term with respect to the subscript (e.g., ρ_x is the Jacobian of ρ with respect to \mathbf{x} and ρ_{xx} is the Hessian of ρ with respect to \mathbf{x}). Inside the sums the suffix $p \in [1 : P]$ is omitted:

$$\begin{aligned} r_x &= \Delta t \left. \frac{\partial r}{\partial \mathbf{x}} \right|_p; & r_u &= \Delta t \left. \frac{\partial r}{\partial \mathbf{u}} \right|_p; \\ r_{xx} &= \Delta t \left. \frac{\partial^2 r}{\partial \mathbf{x}^2} \right|_p; & r_{xu} &= \Delta t \left. \frac{\partial^2 r}{\partial \mathbf{x} \partial \mathbf{u}} \right|_p; & r_{uu} &= \Delta t \left. \frac{\partial^2 r}{\partial \mathbf{u}^2} \right|_p; \\ \rho_x &= \Delta t \left. \frac{\partial \rho}{\partial \mathbf{x}} \right|_p; & \rho_{xx} &= \Delta t \left. \frac{\partial^2 \rho}{\partial \mathbf{x}^2} \right|_p. \end{aligned} \quad (2.23)$$

Taking the two equations (2.18) and (2.21), we then form a local LQR subproblem that can be solved efficiently via a modified Riccati-system [Li and Todorov, 2004]. The latter computes the required changes in the control towards the minimisation of the cost:

$$\delta \mathbf{u}_p = \mathbf{l}_p + \mathbf{L}_p \delta \mathbf{x}_p, \quad (2.24)$$

which is then used to update the nominal command trajectory:

$$\bar{\mathbf{u}} \leftarrow \bar{\mathbf{u}} + \delta \mathbf{u}. \quad (2.25)$$

The new nominal state trajectory $\bar{\mathbf{x}}$ is computed via numerical simulation, and the whole process is then repeated until convergence (i.e., $\delta J \approx 0$). Inside the main iLQG loop, a factor β is used for a modified Levenberg-Marquardt optimisation.

Computing $\delta \mathbf{u}_p$ (Solution to the local LQR problem)

The optimal cost-to-go function ($v_p(\delta \mathbf{x})$) is defined as the cost expected to accumulate if the system (2.18) is initialised in state $\delta \mathbf{x}$ at time p and evolves under the current control law:

$$v_p(\delta \mathbf{x}) = s_p + \delta \mathbf{x}^\top \mathbf{s}_p + \frac{1}{2} \delta \mathbf{x}^\top \mathbf{S}_p \delta \mathbf{x}. \quad (2.26)$$

The cost-to-go (2.26) is obtained by approximating backwards in time, starting with the conditions at the final step $p = P$, where the parameters are defined as: $s_P = \rho$, $\mathbf{s}_P = \rho_{\mathbf{x}}$ and $\mathbf{S}_P = \rho_{\mathbf{xx}}$. For $p < P$, the parameters are updated recursively, as described in the following procedure.

For each time step $p = (P - 1) : -1 : 1$:

a) Compute the terms \mathbf{g} , \mathbf{G} and \mathbf{H} :

$$\begin{aligned} \mathbf{g} &:= (\mathbf{r}_{\mathbf{u}})_p + (\mathcal{F}_{\mathbf{u}}^\top)_p (\mathbf{s})_{p+1} + \sum_i \mathbf{C}_{i,p}^\top \mathbf{S}_{p+1} \mathbf{c}_{i,p}, \\ \mathbf{G} &:= (\mathbf{r}_{\mathbf{xu}})_p + (\mathcal{F}_{\mathbf{u}}^\top)_p (\mathbf{S})_{p+1} (\mathbf{I} + (\mathcal{F}_{\mathbf{x}})_p), \\ \mathbf{H} &:= (\mathbf{r}_{\mathbf{uu}})_p + (\mathcal{F}_{\mathbf{u}}^\top)_p (\mathbf{S})_{p+1} (\mathcal{F}_{\mathbf{u}})_p + \sum_i \mathbf{C}_{i,p}^\top \mathbf{S}_{p+1} \mathbf{C}_{i,p}. \end{aligned} \quad (2.27)$$

b) Obtain the affine control law by minimising:

$$\alpha(\delta \mathbf{u}, \delta \mathbf{x}) = \delta \mathbf{u}^\top (\mathbf{g} + \mathbf{G} \delta \mathbf{x}) + \frac{1}{2} \delta \mathbf{u}^\top \mathbf{H} \delta \mathbf{u}, \quad (2.28)$$

with respect to \mathbf{u} , which results in:

$$\delta \mathbf{u} = -\mathcal{H}^{-1}(\mathbf{g} + \mathbf{G}\delta \mathbf{x}). \quad (2.29)$$

The term \mathcal{H}^{-1} is a modified version of the Hessian \mathbf{H} obtained using modified Cholesky decomposition (in order to avoid negative eigenvalues, that would cause the cost function to become arbitrarily negative):

$$\begin{aligned} [\mathbf{V}, \mathbf{D}] &= \text{eig}(\mathbf{H}) & (2.30) \\ \text{(the eigenvectors } (\mathbf{V}) \text{ and eigenvalues } (\mathbf{D}) \text{ of } \mathbf{H}) \\ \text{If } (\mathbf{D}(i, i) < 0) : \mathbf{D}(i, i) &= 0 \forall i \\ \mathbf{D} &= \mathbf{D} + \mathbf{I}\beta \\ \mathcal{H}^{-1} &= \mathbf{V}\mathbf{D}^{-1}\mathbf{V}^T \end{aligned}$$

The control law mentioned in (2.24) has a special form consisting of an open loop component (\mathbf{l}_p) and a feedback element ($\mathbf{L}_p\delta \mathbf{x}_p$). These correspond to the first and second terms of the solution in (2.29).

c) Update the cost-to-go approximation parameters:

$$\begin{aligned} \mathbf{S}_p &= \mathbf{r}_{\mathbf{xx}} + (\mathbf{I} + (\mathcal{F}_x)_p)^T \mathbf{S}_{p+1} (\mathbf{I} + (\mathcal{F}_x)_p) + \mathbf{L}_p^T \mathbf{H} \mathbf{L}_p + \mathbf{L}_p^T \mathbf{G} + \mathbf{G}^T \mathbf{L}_p, \\ \mathbf{s}_p &= \mathbf{r}_x + (\mathbf{I} + (\mathcal{F}_x)_p)^T \mathbf{s}_{p+1} + \mathbf{L}_p^T \mathbf{H} \mathbf{l}_p + \mathbf{L}_p^T \mathbf{g} + \mathbf{G}^T \mathbf{l}_p, & (2.31) \\ s_p &= r + s_{p+1} + \frac{1}{2} \sum_i \mathbf{c}_{i,p}^T \mathbf{S}_{p+1} \mathbf{c}_{i,p} + \frac{1}{2} \mathbf{l}_p^T \mathbf{H} \mathbf{l}_p + \mathbf{l}_p^T \mathbf{g}. \end{aligned}$$

After the solution has converged, the algorithm returns the optimal control sequence $\bar{\mathbf{u}}$, the corresponding state trajectory $\bar{\mathbf{x}}$ and the optimal feedback control law $\bar{\mathbf{L}}$. A pseudocode of the iLQG method is given in Algorithm 1; for more details, we refer the reader to [Todorov and Li, 2005].

2.3.2 Practical Issues and Limitations

The iLQG algorithm, along with most optimal control algorithms can only handle smooth dynamics and cost functions. We note that obtaining good optimisation results with iLQG requires a fair amount of practical experience and a good un-

Algorithm 1 Pseudocode of the iLQG algorithm

GIVEN:

- Model of plant dynamics (2.1)
- Encoding of task through cost function (2.3)

INITIALISE:

- discretise time as $p = 1 \dots P$ with $\Delta t = T/(P - 1)$
- $\bar{\mathbf{u}}_p = \bar{\mathbf{u}}_0$, corresponding $\bar{\mathbf{x}}_p = \bar{\mathbf{x}}_0$
- Levenberg-Marquardt constant $\beta = \beta_{init}$

REPEAT:

- Linearise dynamics around $\bar{\mathbf{x}}_p$ (2.18)
- Find quadratic approximation of the cost (2.21)
- solve the obtained local LQR problem $\rightarrow \delta \mathbf{u}_p$

Solution to local LQR problem

- for $p = (P - 1) \dots 1$
 - compute the terms \mathbf{g} , \mathbf{G} and \mathbf{H} (2.27)
 - compute \mathcal{H}^{-1} (modified Cholesky decomposition) (2.30)
 - compute $\delta \mathbf{u}_p = -\mathcal{H}^{-1}(\mathbf{g} + \mathbf{G}\delta \mathbf{x}_p)$
 - compute the cost-to-go parameters $\mathbf{S}_p, \mathbf{s}_p, s_p$ (2.31)
- apply control law forward in time to the linearised system ($\delta \mathbf{x}_{p+1} = \mathcal{F}_x \delta \mathbf{x}_p + \mathcal{F}_u \delta \mathbf{u}_p$ with $\delta \mathbf{x}_1 = 0$)
- compute $\tilde{\mathbf{u}}_p = \bar{\mathbf{u}}_p + \delta \mathbf{u}_p$, corresponding $\tilde{\mathbf{x}}_p$ and δJ
- **If** (convergence - i.e. $\delta J \approx 0$)
STOP
- **Else** - (update Levenberg-Marquardt constant β)
 - If** ($\delta J > 0$) : $\bar{\mathbf{u}}_p = \tilde{\mathbf{u}}_p$, $\bar{\mathbf{x}}_p = \tilde{\mathbf{x}}_p$, decrease β
 - Else** : increase β
 - End If**
- **End If**

UNTIL convergence

derstanding of the given optimisation task. This is due to the large number of open parameters, such as cost function parameters, convergence constants or simulation parameters, which all influence the optimisation outcome. As mentioned previously, the obtained results will also depend on the chosen initial control sequence, due to the local nature of the method. This problem can be overcome using multiple initialisation and selecting the optimal solution (out of the obtained batch) at the end.

When the control limits are reached within the solution, the algorithm resets the feedback gains to zero. This allows iLQG to cope with bounds on the control input (in a deterministic case), while in a stochastic scenario this problem can be avoided by carefully shaping the cost function accordingly (i.e., adjusting the control cost weights).

In addition to requiring smooth dynamics, in the currently existing implementation, iLQR/iLQG is designed to admit only predefined finite horizon tasks. The task of choosing the time is left to the user and requires a significant amount of tuning.

2.4 SUMMARY

In this chapter, we reviewed the main elements of OC theory. We summarised the formulation and the main approaches, namely local and global methods. We then presented the case of iterative techniques that constitute a good compromise between the two. We focused on iLQG as our method of choice for investigation and starting point for development.

In the next chapter, we introduce the concept of impedance modulation, which creates redundancy in the controls. We present how this added DOF in the controls can be exploited using OC and the benefits of impedance modulation, as demonstrated by previous investigations. We expand on existing work by applying OC to a system capable of damping modulation, alongside that of stiffness and position.

OPTIMAL CONTROL OF SYSTEMS EQUIPPED WITH VARIABLE IMPEDANCE ACTUATORS

In this chapter we introduce the concept of actuators that allow for simultaneous modulation of the joint position and impedance, called *Variable Impedance Actuators (VIAs)*. We present the difficulties raised by the task of planning for systems with such redundancy in the controls. Then, we give a brief overview of their benefits and previously used approaches.

We follow with a study on OC for an actuator capable of simultaneous and independent physical damping and stiffness modulation. Using iLQG (Sec. 2.3.1), we explore how variable physical damping can be exploited in the context of rapid movements. Several numerical simulation results are presented, in addition to a hardware experiment.

3.1 RELATED WORK ON THE DESIGN OF VIAS

The traditional approach regarding joint actuation used in robotic systems in the past implies stiff dynamics. A stiff (non-VIA) actuator is capable to move to a specific position (or track a trajectory) with high accuracy; once the desired location is reached, in the absence of a new command, it holds that position, regardless of the external forces applied (within the design bounds).

This type of operation is acceptable in the case of robots acting in highly constrained environments, performing a well defined task that does not require adaptation, nor interaction with humans. Recent developments opened new areas where robotic systems can be employed, some of which feature a changing environment and working in close proximity to humans. This created a need for multifaceted systems, able to meet these demands.

To address this need, several designs offering compliant actuation were implemented [Vanderborght et al., 2012]. Most research has focused on the design of *Variable Stiffness Actuators (VSAs)* [Ham et al., 2009]. These new implementations showed significant improvement in terms of safety due to inbuilt compliance [Zinn

et al., 2004], energy efficiency, robustness to perturbations [Vanderborght et al., 2013] and dynamic range [Hurst et al., 2004].

From the perspective of how the modulation of impedance is achieved we distinguish :

- i **active impedance**, in which the hardware uses software control to emulate a desired impedance behaviour [Loughlin et al., 2007]. This requires an actuator, a sensor and a controller which are fast enough to create the desired result. This technology is available in systems such as the DLR KUKA [Bischoff et al., 2010] or the SARCOS robot [Bilodeau and Papadopoulos, 1998]. While this approach has the capability of modulating the impedance in an on-line fashion, for a theoretically infinite range and speed, it is unable to store energy or exploit the natural dynamics [Vanderborght et al., 2013].
- ii **passive impedance**, where the actuator contains a compliant element. This can exhibit either fixed or adaptable impedance capabilities. Actuators in the latter category are able to use their elastic components to store energy [Vanderborght et al., 2013]. They require separate motors for position control and impedance modulation. Most of the studies on design and control of such VIA focus on stiffness modulation. In the next section we describe some of the available implementations for these variable stiffness actuators (VSA).

VARIABLE STIFFNESS ACTUATORS In terms of mechanical design for adaptable stiffness actuators in ii), three categories are distinguished [Vanderborght et al., 2013]. Spring preload VSAs operate by modulating the pretension on a spring (or a set of springs). A popular approach is the antagonistic design, with antagonistic motors (e.g., Edinburgh SEA (Series Elastic Actuator) [Mitrovic et al., 2010b] Fig. 3.1 (c)). There, two opposing actuators are used to modulate the position (by moving in the same direction) and stiffness (by acting in opposite directions).

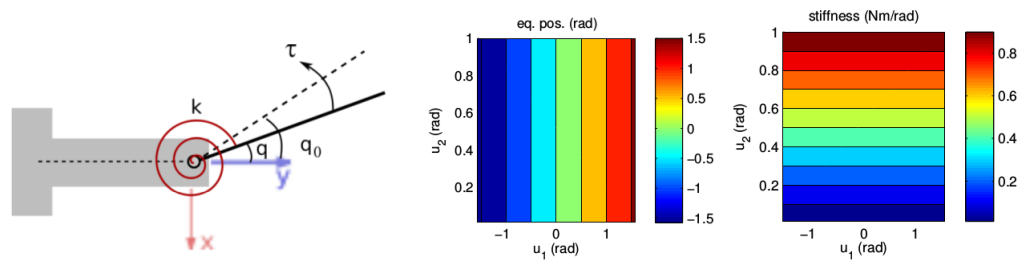
This has some deficiencies in terms of energy consumption: relying on co-contraction to increase stiffness requires the use of non-backdriveable actuators, otherwise motor power must be used continuously to maintain a stable joint position at high stiffness. Even if the appropriate actuators are selected, the energy cost of changing stiffness from compliant to stiff can still be considerably high.

Using a design in which the motors are not opposed in the antagonistic way allows for less energy consumption, a smaller volume and lower mass [Okada et al., 2001; Zinn et al., 2004]. The series pretension design consists of a spring,

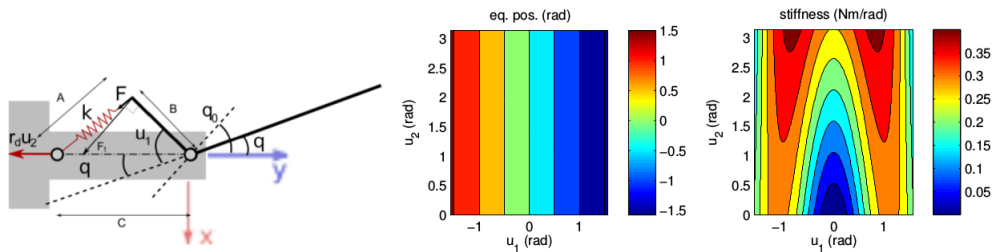
which can be pre-stretched, in series with the actuator [Ham et al., 2007] (Fig. 3.1 (b)), but it also requires a significant amount of energy to increase stiffness.

The second design type relies on varying the transmission ratio between the load and the spring to modulate the stiffness. Lastly, VSAs can also function by changing the physical properties of a spring. For an in-depth overview of each category and their various sub-classes we refer the reader to the studies in [Vanderborght et al., 2012, 2013].

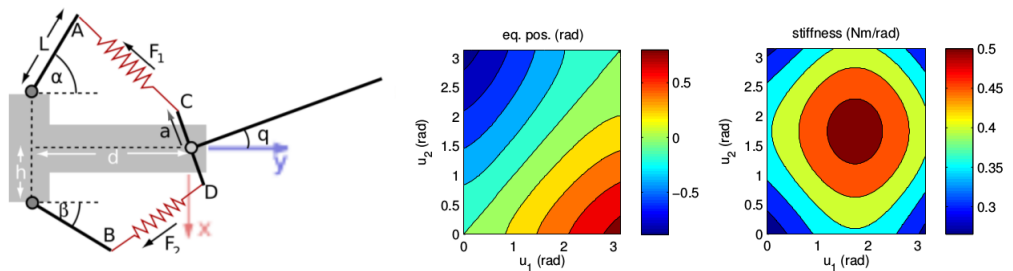
These VSA configurations often increase the complexity of the system dynamics by adding coupling between motion and the impedance features and by introducing non-linearities (particularly if the system has multiple DOFs). Thus, the complexity of the planning and control problem increases correspondingly.



(a) Ideal one-link VSA



(b) MACCEPA



(c) Edinburgh SEA

Figure 3.1: Variable Stiffness Actuators Left: Model of the system. Right: Equilibrium position and stiffness as a function of commands \mathbf{u} (evaluated at $q = 0, \dot{q} = 0$).

3.2 MODEL OF A ROBOTIC SYSTEM WITH COMPLIANT ACTUATION

In this section we present a general model of a robotic system equipped with VSA capabilities. We introduce the dynamics of the systems and the added complexity of redundant actuation. We then address the question of control for such systems.

3.2.1 Robot Dynamics

In a robotic system with n degrees of freedom, the equation of the motion system has the following general form:

$$\mathbf{M}(\mathbf{q}) \ddot{\mathbf{q}} + \mathbf{C}(\mathbf{q}, \dot{\mathbf{q}}) \dot{\mathbf{q}} + \mathbf{g}(\mathbf{q}) + \mathbf{D}\dot{\mathbf{q}} = \boldsymbol{\tau}(\mathbf{q}, \mathbf{u}) + \boldsymbol{\tau}_f, \quad (3.1)$$

where $\mathbf{q}, \dot{\mathbf{q}}, \ddot{\mathbf{q}} \in \mathbb{R}^n$ are the joint angles, velocities and accelerations, $\mathbf{M} \in \mathbb{R}^{(n \times n)}$ is a symmetric positive definite mass matrix, $\mathbf{C} \in \mathbb{R}^n$ represents centrifugal and Coriolis forces, $\mathbf{g} \in \mathbb{R}^n$ are the gravitational forces, $\mathbf{D} \in \mathbb{R}^{(n \times n)}$ is the viscous damping matrix, $\boldsymbol{\tau} \in \mathbb{R}^n$ the joint torques from the actuators, $\mathbf{u} \in \mathbb{R}^m$ the motor positions and $\boldsymbol{\tau}_f \in \mathbb{R}^n$ represents the torques due to joint friction.

3.2.2 Compliant Actuation

If the motors are directly connected to the links ($\mathbf{u} = \mathbf{q}$), the joint torques correspond directly to the motor torques and can be considered part of the control inputs [Siciliano and Khatib, 2008]. This is not the case for actuators with compliant elements, where the joint torques $\boldsymbol{\tau}$ cannot be directly controlled and the torque function is generally dependent on the positions of the joints (\mathbf{q}) and the motor commands (\mathbf{u}) (as expressed in (3.1)). There are often range constraints on the control commands:

$$\mathbf{u} \in [\mathbf{u}_{\min}; \mathbf{u}_{\max}], \quad (3.2)$$

with \mathbf{u}_{\min} and \mathbf{u}_{\max} the lower and upper bounds on the motor positions.

As mentioned previously, due to the nature of the VSAs, the relation between the joint torques, joint angles and motor commands can be quite complicated ¹:

$$\boldsymbol{\tau}(\mathbf{q}, \mathbf{u}) = \mathbf{A}^\top(\mathbf{q}, \mathbf{u})\mathbf{F}(\mathbf{q}, \mathbf{u}), \quad (3.3)$$

where $\mathbf{A} \in \mathbb{R}^{(p \times n)}$ ($p \geq n$) is the moment-arm matrix, defined by the geometry of the actuators, and $\mathbf{F} \in \mathbb{R}^p$ are the corresponding forces due to the elastic elements [Braun et al., 2012a] (determined by the physical characteristics of these elements). The joint stiffness matrix ($\mathbf{K} \in \mathbb{R}^{(n \times n)}$) is defined as:

$$\mathbf{K}(\mathbf{q}, \mathbf{u}) = -\frac{\partial \boldsymbol{\tau}(\mathbf{q}, \mathbf{u})}{\partial \mathbf{q}} = \frac{\partial \mathbf{A}^\top}{\partial \mathbf{q}}\mathbf{F} + \mathbf{A}^\top \frac{\partial \mathbf{F}}{\partial \mathbf{q}}. \quad (3.4)$$

In general this is a non-linear function of the state and commands and, depending on the implementation, the stiffness of the individual joints may be coupled [Howard et al., 2011]. The formulation in (3.4) observes the dependencies between the control inputs (\mathbf{u}) and the joint stiffness (\mathbf{K}).

This dependency can allow independent modulation of the joint torques and of the (passive) joint stiffness of the actuators, assuming the joint torque function (3.3) is redundant with respect to the motor positions (i.e., $m > n$).

In an idealised VSA design (Fig. 3.1 (a) left) the stiffness and equilibrium position are directly controllable and independent, such that any combination is realisable (within the actuator design capabilities). This is illustrated in Fig. 3.1 (a) right panel where, any combination of equilibrium position ($q_0 = u_1$) and stiffness ($k = u_2$) is possible. However, real mechanisms rarely exhibit such idealised behaviour.

By comparison, in the Mechanically Adjustable Compliance and Controllable Equilibrium Position Actuator (MACCEPA [Ham et al., 2007] Fig. 3.1 (b)) and Edinburgh series elastic actuator (Edinburgh SEA [Mitrovic et al., 2010b] Fig. 3.1 (c)) present much more intricate dependencies between the commands, equilibrium position and stiffness. In the case of the MACCEPA the joint torque (3.3) is given by:

$$\tau(q, u_1, u_2) = \kappa_s BC \sin \alpha \left(1 + \frac{r u_2 - (C - B)}{\sqrt{B^2 + C^2 - 2BC \cos \alpha}} \right), \quad (3.5)$$

¹ In several formulations the torque can depend on the joint velocity $\dot{\mathbf{q}}$ (e.g., when considering viscoelastic forces). The formulation used remains valid in those scenarios.

where B , C are shown in the diagram (Fig. 3.1 (b), left panel), r is the radius of the drum attached to the pretensioning servo, $\alpha = u_1 + q$, and κ_s is the spring constant of the spring. While the equilibrium position is only influenced by the position of the first motor, the stiffness has a complex, non-linear dependency on both the equilibrium position and spring pre-tension commands (Fig. 3.1 (b), right panel).

Similarly, in the Edinburgh SEA (Fig. 3.1 (c)) the joint torque takes the form in:

$$\boldsymbol{\tau}(\mathbf{x}, \mathbf{u}) = -\hat{\mathbf{z}}^\top (\mathbf{a} \times \mathbf{F}_1 - \mathbf{a} \times \mathbf{F}_2), \quad (3.6)$$

where $\hat{\mathbf{z}}$ is the unit vector along the rotation axis, $\mathbf{a} = (a \cos q, a \sin q, 0)^\top$, \mathbf{F}_i (with $i \in \{1, 2\}$) are the forces due to the two springs. In this antagonistic design, there is a highly non-linear relationship between the motor commands and the joint equilibrium position and stiffness (Fig. 3.1 (c), right panel).

As shown by these examples, even a relatively simple VSA designs can present quite a challenging task in terms of position and stiffness modulation. Simple pre-sets, such as the duality principle [Anderson and Spong, 1988], are viable solutions only for specific cases; in a real environment the robot has to adapt to changes in the structure of the world and the tasks. This can only be achieved by continuous and fast modulation of stiffness. The OC formulation can provide a good methodology for obtaining an adequate control strategy for such devices.

3.2.3 Optimal Control Formulation

In order to formulate an OC problem (Chapter 2.1), the system dynamics and a performance index (cost function) are required. Given the previously described robotic system with VSA, an adequate state-space representation would be:

$$\dot{\mathbf{x}} = \mathbf{f}(\mathbf{x}, \mathbf{u}), \quad (3.7)$$

where:

$$\mathbf{f} = \begin{bmatrix} \mathbf{x}_2 \\ \mathbf{M}^{-1}(\mathbf{x}_1) (-\mathbf{C}(\mathbf{x}_1, \mathbf{x}_2)\mathbf{x}_2 - \mathbf{g}(\mathbf{x}_1) - \mathbf{D}\dot{\mathbf{x}}_2 + \boldsymbol{\tau}(\mathbf{x}_1, \mathbf{x}_3)) \end{bmatrix} \quad (3.8)$$

and $\mathbf{x} = [\mathbf{x}_1^\top, \mathbf{x}_2^\top]^\top = [\mathbf{q}^\top, \dot{\mathbf{q}}^\top]^\top \in \mathbb{R}^{2n}$. The range constraints on the motor positions (3.2) can be incorporated as control constraints.

Given a cost functional (2.3), encoding the desired task, the optimal control \mathbf{u}^* defined in (2.4) can then be obtained using the methodology presented in Sec. 2.2.

Using the optimal control command obtained \mathbf{u}^* and the corresponding trajectory $\mathbf{x}^* = [\mathbf{q}^{*\top} \dot{\mathbf{q}}^{*\top}]^\top$, the optimal torques $\boldsymbol{\tau}^*$ and stiffness \mathbf{K}^* are obtained from (3.3) and (3.4). The framework resolves the control redundancy (i.e., $\mathbf{u} \in \mathbb{R}^m$, $m > n$) taking into consideration the specifics of the system (3.1), actuator (3.3) and task (2.3) at hand [Nakamura and Hanafusa, 1987].

3.3 PREVIOUS WORK ON OPTIMAL CONTROL OF VSAS

As shown in the previous section, VSA configurations often increase the complexity of the system dynamics by adding coupling between the motion and the impedance features and by introducing non-linearities (particularly if the system has multiple DOFs).

Various studies showed that, by fully including stiffness modulation in the optimisation problem, additional improvements can be achieved in terms of noise rejection [Mitrovic et al., 2009]. In [Braun et al., 2012b] the energy efficiency of a throwing movement is raised by controlling the control of actuator's ability to store and release energy. The same capability is explored for periodic tasks such as brachiation [Nakanishi et al., 2011; Nakanishi and Vijayakumar, 2013]. In [Howard et al., 2010] the authors show how a general impedance strategy for a hitting task can be adapted to the particularities of a hardware implementation (i.e., actuator and system dynamics). By devising a system specific solution, the dynamics are fully exploited, ensuring a better task performance. Similarly, [Garabini et al., 2011] showed how a much higher link velocity can be obtained when the VSA capabilities are exploited for a hitting movement.

These results are consistent with observed patterns displayed by humans and other biological systems in adapting their impedance under similar conditions [Mitrovic et al., 2010c].

While stiffness modulation has been the focus of various studies, the damping component of the impedance has been often overlooked. Our hypothesis is that a VIA able to achieve simultaneous and independent stiffness and damping modulation would be able to combine the benefits of both approaches. There are, of course, several difficulties, the most salient being the increase in complexity of the system dynamics, and thus, of the planning and control problem.

In [Radulescu et al., 2012] we combine variable physical stiffness and damping in a single actuator using insights from adjustable dampers in haptic interfaces [Mehling et al., 2005; Srikanth et al., 2008]. In the next section, we present the results obtained, which indicate that damping modulation, alongside those of stiffness and position, can improve performance.

3.4 SIMULTANEOUS AND INDEPENDENT STIFFNESS AND DAMPING MODULATION

In this section we first discuss the issue of damping in the context of variable physical impedance actuation and present the hardware implementation employed in our experiments. We then introduce the task used in our investigation and the control framework employed for its realisation.

3.4.1 *Damping in Variable Stiffness Actuators*

To investigate the effects of damping modulation in this framework, we select the simplest system available: a one-link compliant joint system moving on a horizontal plane. The equation of motion (3.1) for such a system becomes:

$$M\ddot{q} = \tau + \tau_f. \tag{3.9}$$

The torque from the actuators is defined in our case as $\tau = \tau_k + \tau_b$, with τ_k and τ_b representing the stiffness and damping torques, respectively.

An appropriate damping magnitude is required for the accurate control of the VIA system. If the damping level is too low, oscillations occur during rapid movements. If the level is too high, the response of the system becomes slow, rendering many of the advantages of (variable) passive stiffness, such as energy storage, unexploitable. One may also say that, in general, it is natural to consider adding variable damping to a system with variable stiffness in order to achieve an appropriate response (e.g., under-, over-, or critical damping) as the stiffness varies (for a detailed analysis, see Appendix A). In the following, we explore these issues with respect to a simple example of a variable stiffness actuator (VSA).

3.4.2 Example: Mechanically Adjustable Compliance and Controllable Equilibrium Position Actuator

We selected for our experiments, as an example of a VSA that is easy to build and intuitive to operate, the Mechanically Adjustable Compliance and Controllable Equilibrium Position Actuator (MACCEPA), illustrated in Fig. 3.2.

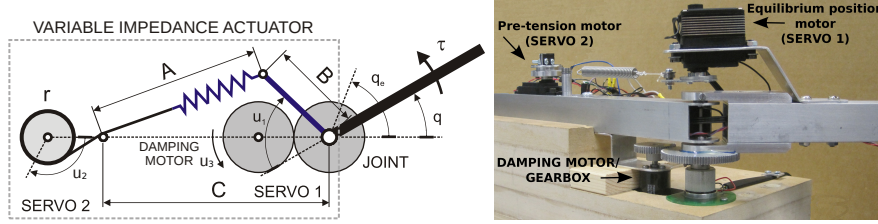


Figure 3.2: Mechanically Adjustable Compliance and Controllable Equilibrium Position Actuator (MACCEPA) [Ham et al., 2007] with Variable Damping: model (left) and robot (right). Parameters: $r = 0.01\text{m}$, $B = 0.03\text{m}$, $C = 0.13\text{m}$, link mass: 0.125kg , link length: 0.295m , centre of mass location on the link: 0.1475m

The MACCEPA design allows simultaneous equilibrium position and joint stiffness control by using two independently controlled servomotors (Fig. 3.1 (b), right). The first servomotor (attached to the free-moving link) adjusts the position of a lever of length B , while the second (attached to the base) adjusts the position of a winding drum of radius r attached via a cord to the spring (Fig. 3.2, left).

The torque around the joint (3.5) is generated when a discrepancy is introduced between the angle of the first servomotor u_1 and that of the joint q (i.e., $\alpha := u_1 - q \neq 0$), thus extending the spring away from its rest length $L_0 = C - B$. The restorative spring force is $F = -\kappa_s(l - l_0)$, where κ_s is the spring constant and l is the resultant length of the spring. The latter is given by the length $A = B^2 + C^2 - 2BC\cos\alpha$ (Fig. 3.2, left) plus any additional extension caused by the pretensioning servo u_2 , i.e., $l = A + ru_2$. The equilibrium position can be directly controlled by setting $q_0 = u_1$, while the stiffness ($k := -\partial\tau/\partial q$) is:

$$k(q, u_1, u_2) = \kappa_s BC \cos \alpha \left(1 + \frac{r u_2 - (C - B)}{\sqrt{B^2 + C^2 - 2BC \cos \alpha}} \right) - \frac{\kappa_s B^2 C^2 \sin^2 \alpha (r u_2 - (C - B))}{(B^2 + C^2 - 2BC \cos \alpha)^{\frac{3}{2}}}. \quad (3.10)$$

For full details of the derivation of the torque relationship (3.5) we refer the reader to [Ham et al., 2007]. We note that, as with many VSA designs, the joint

torque and stiffness have a non-linear dependence on the joint and motor configurations.

In addition to this existing design, we add a third control variable u_3 , corresponding to damping modulation. Thus, for the combined system (MACCEPA with Variable Damping, MACCEPA-VD) the torque relationship corresponding to τ from (3.9) is:

$$\tau(q, \dot{q}, u_1, u_2, u_3) = \tau_k(q, u_1, u_2) + \tau_b(\dot{q}, u_3). \quad (3.11)$$

The details of the mechanical design used are presented in Sec. 3.4.5.

3.4.3 *Selecting the System Response*

An adequate task-dependant system response is required for the efficient control and operation of a VSA such as the MACCEPA. For periodic tasks, (e.g., running), we may desire an underdamped response, which will encourage oscillation through the natural dynamics. On the other hand, in tasks such as tracking, we may require a response close to critical damping, to secure reaching the target reference as quickly as possible while avoiding overshooting.

For one-link linear-time-invariant (LTI) systems, this is relatively straightforward, as the damping response is characterised by the damping ratio ξ :

$$\xi = b/2\sqrt{kM}, \quad (3.12)$$

where M represents the inertia, b the damping and k the stiffness. In an LTI system selecting, for example, a critically damped response simply requires choosing b such that $\xi = 1$.

However, systems equipped with VSA (e.g., $k = k(q, u_1, u_2)$) are not LTI by definition. Choosing the damping response for such set-ups is non-trivial, and requires the modulation of the damping b in a complex way. This is evident in the case of the MACCEPA, given the nonlinear dependence of the stiffness (3.10) on the joint and motor configurations.

Furthermore, since such systems are not LTI, the damping response is not well-characterised through ξ as defined by (3.12) and so, for example, ensuring $\xi = 1$ (e.g., by coupling the stiffness and damping through (3.12)), may not result in the fastest non-oscillatory response (i.e., critical damping). This expectation is confirmed by the results obtained in our experiments (presented in Sec. 3.4.7).

3.4.4 *Related Work*

Active, passive and semi-active damping are all viable ways to modulate the dissipative properties of a dynamical system. The most frequently adopted design is *active damping* realised through velocity feedback. One of the major deficiencies of this implementation is the significant energy cost. Additionally, while the implementation of this method can be realised entirely by software control, its stable dissipative operation is limited by the bandwidth of the feedback loop.

Alternatively, *passive damping* can be implemented using dissipative mechanical components (i.e., dampers, brakes), that are guaranteed to be passive regardless of the excitation. Though they do not present bandwidth limitation, their damping characteristics cannot be changed during task execution. A compromise between these two approaches is represented by active modulation of the passive damping.

This method, called *semi-active damping*, is mainly realised using magnetorheological (MR) dampers or frictional dampers (FD). In the case of the former, damping is controlled with a magnetic field, that solidifies the magnetic fluid (a suspension of iron particles and oil as a base fluid) which fills the damper [Guglielmino et al., 2005; Unsal et al., 2004]. The main disadvantage of this approach is the large amount of additional space required. FD designs, on the other hand, make use of Coulomb friction between two moving bodies by controlling the normal contact force. A common realisation is through piezo-electric actuators [Laffranchi et al., 2010; Unsal et al., 2004]. Semi-active damping requires additional elements in mechanical design, but presents higher bandwidth and less energy consumption compared to active damping.

In [Laffranchi et al., 2010], semi-active damping modulation in robotic systems is suggested as an approach to suppress unwanted oscillation caused by the compliance of the joint, together with a specific design of the FD type. The introduction of damping modulation has been shown to improve the performance of robots acting in human environments [Laffranchi et al., 2013].

3.4.5 *Mechanism Design*

In order to implement our variable damping actuator we considered an alternative design which exploits a DC motor damping effect. The basic concept behind the approach is to attach a back-drivable motor/gearbox unit to the joint (see Fig. 3.2), that operates as a generator during the motion. In this way the mechanical energy

is converted into the electrical domain where it is dissipated through a resistor. To our knowledge, this was the first time simultaneous and independent stiffness and damping modulation were used in a single actuator. In the following, we outline the details of the variable damping mechanism in our actuator.

A Variable Passive Electrical Damping

Shorting (i.e., connecting together) the output terminals of electric motors creates a motor braking effect whereby rotations of the output shaft, due to an externally applied torque, are opposed by a torque associated with the current induced within the motor. We use this electrical effect in order to manipulate the passive mechanical damping in our variable impedance system.

The electronics and equivalent circuit corresponding to our variable passive damper is depicted in Fig. 3.3. In this case, applying a voltage V will generate current I according to:

$$L\dot{I} + R_e I + V_{emf} = V, \quad (3.13)$$

where L is the inductance of the circuit, R_e the effective resistance and V_{emf} is the back electromotive force (e.m.f.) generated when the motor armature rotates. The relations between (i) the angular velocity of the output shaft $\dot{\theta}$ and V_{emf} and (ii) the motor torque τ_m (reflected through the gear reduction) and (iii) the current I are given by:

$$V_{emf} = n\kappa_q \dot{\theta}, \quad \tau_m = n\kappa_\tau I, \quad (3.14)$$

where κ_q and κ_τ are the speed and torque constants of the motor and $n : 1$ defines the gear reduction.

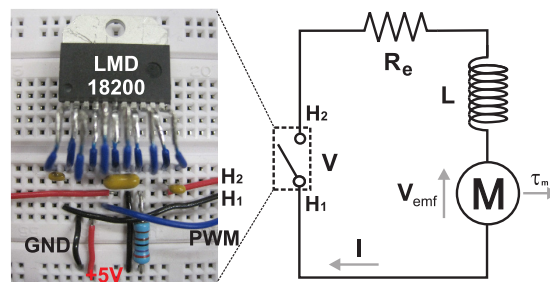


Figure 3.3: H-bridge damping circuit (left) and equivalent circuit (right)

We now seek to formulate the relation between the motor torque and the operation of the electrical circuit. From (3.14) we observe that opening the electrical

circuit (i.e., $I = 0$) reduces the motor torque to zero $\tau_m = 0$, while closing the circuit, generates a torque dependent on V and $\dot{\theta}$. Additionally, if we assume that the time constant $\tau_{RL} = L/R$ of the equivalent (RL) circuit is small ² we can perform a steady-state approximation of the current as: $I \approx I_{ss} = (V - V_{emf})/R_e$, and analytically approximate the torque relation:

$$\tau_m = n\kappa_\tau I_{ss} = \frac{n\kappa_\tau}{R_e} V - \frac{n^2\kappa_\tau\kappa_{\dot{q}}}{R_e} \dot{\theta}. \quad (3.15)$$

This connection shows that τ_m constitutes of a voltage dependent driving torque (first term) and a motion dependent braking torque (second term). Correspondingly, by shorting the terminals of the circuit together (i.e., $V = 0$), we can generate viscous damping at the motor output shaft:

$$\tau_b(\dot{\theta}) = -b_e \dot{\theta} = -\frac{n^2\kappa_\tau\kappa_{\dot{q}}}{R_e} \dot{\theta}. \quad (3.16)$$

This represents a purely passive damping provided by the motor in the absence of any velocity feedback. As indicated by (3.16), the damping constant b_e depends on the resistance of the equivalent circuit R_e . Thus, by modulating this resistance, we can achieve a variable passive damping effect. This can be obtained by different means, for example, we could add a mechanical variable resistor, digital potentiometer, or another integrated circuit across the motor terminals to adjust R_e . In such a scenario the resistance of the equivalent circuit is the sum of the internal resistance of the motor and the variable resistance component.

In our design, we do not employ variable resistance directly, but instead we emulate its effect using an H-bridge (NSC: LMD18200) circuit. This circuit opens and closes the connection between the motor terminals according to a PWM (pulse-width modulation) signal with controllable duty cycle $u_3 \in [0, 1]$. Consequently, the damping is modulated according to the proportion of time the connection is open or closed: if the connection is open the damping effect vanishes (i.e., $u_3 = 0 \Rightarrow \tau_b = 0$), while if the connection is closed maximal damping is achieved ($u_3 = 1 \Rightarrow \tau_b = -b_e \dot{\theta}$). Experiments characterising the damping effect in our robotic device are reported in Sec. 3.4.7.c. For details of the system identification and estimation of the damping used in our work, please see Appendix B.

² According to (3.13), the dynamics of the current follow an exponential tendency given by $e^{-t/\tau_{RL}}$. For our set-up, $\tau_{RL} = L/R = 7 \times 10^{-5}$ (i.e., $L = 1.37 \times 10^{-3} \text{H}$, and $R_e = 20.2 \Omega$), making the steady-state approximation of the current and the associated algebraic relation of the torque (3.15) feasible.

The variable damping mechanism described can be incorporated into numerous VSAs by rigidly connecting the DC motor shaft to the rotating part of the joint. In our implementation we use a spur gear rigidly attached to the free link to pick up the joint rotation of the MACCEPA (see Fig. 3.2 (right)).

3.4.6 Control Framework

As discussed in Sec. 3.4.3, the non-linear, time-varying nature of many VIAs renders the appropriate selection of the damping a non-trivial task. In order to avoid this difficulty, we use an optimal control approach to devise strategies for exploiting this additional degree of freedom. We employ the iLQG control framework (as described in Sec. 2.3.1).

In order to apply this control framework to the MACCEPA-VD, we define the state and control inputs as $\mathbf{x} = (q, \dot{q})^T = (x_1, x_2)^T$ and $\mathbf{u} = (u_1, u_2, u_3)^T$, respectively. In this representation, we obtain \ddot{q} from (3.9) and replace it in (2.1), which gives us:

$$\dot{\mathbf{x}} = \begin{pmatrix} \dot{q} \\ \ddot{q} \end{pmatrix} = \begin{pmatrix} x_2 \\ ((\tau(\mathbf{x}, \mathbf{u}) + \tau_f)/J) \end{pmatrix}. \quad (3.17)$$

The admissible command range, imposed by the mechanical stops in the servos and PWM restrictions, is $(-\pi/2, 0, 0)^T \preceq \mathbf{u} \preceq (\pi/2, \pi/2, 1)^T$. To encode the objective of rapid, accurate reaching to some target, we define the cost function:

$$J = \int_0^T (q(t) - q^*)^2 + \epsilon \mathbf{u}(t)^T \mathbf{u}(t) dt, \quad (3.18)$$

where q^* is the target point in joint space, and T is the maximum time duration allowed for the movement.

We note that, the first term in (3.18) has several influences on the solution. Firstly, it promotes rapid movement by penalising solutions for time spent away from the target point q^* . Secondly, it ensures the system comes to a halt at the target (if the target is reached before the final time T , the optimal solution is to remain there to avoid incurring further costs). Finally, it penalises overshoot (i.e., deviations for which $q(t) - q^* > 0$). The second term acts to regularise solutions, with ϵ set to a small constant value (in our experiments $\epsilon = 10^{-8}$).

3.4.7 Experiments: Exploiting Variable Physical Damping in Rapid Movement Tasks

Having introduced the problem of variable damping modulation, we now investigate how this can be exploited in the context of optimal control of rapid movements on VIAs.

First, we present numerical results illustrating the benefits of variable damping over (optimal) fixed rate damping schemes while the stiffness varies due to the movement and/or the control. Finally, we showcase the experiment in hardware, verifying the effectiveness of variable damping in the MACCEPA-VD.

A Optimal Exploitation of Variable Damping

Our first numerical experiment focuses on highlighting the benefits of optimal damping modulation compared to traditional fixed rate damping schemes in the context of rapid movements. For this, we compare reaching behaviour on the MACCEPA-VD with: (i) variable damping (with the damping command optimised with respect to (3.18)), (ii) fixed damping (also optimised with respect to (3.18)) and (iii) damping coupled to the stiffness through (3.12) according to $\xi = 1$.

To keep the analysis simple, we start by comparing behaviours for which the equilibrium position command is fixed at the target $u_1 = q^* = \pi/6$ rad and the spring pre-tension command is fixed to $u_2 = \pi/8$ rad. We note that, due to the dependence of the stiffness k on the joint position q (3.10), the movement produces a natural (uncontrolled) variation of stiffness.

To evaluate performance, for each system, we calculate the cost incurred under (3.18), and the settling time, defined as the time at which (i) the absolute distance to the target is less than a threshold value of $|q^* - q(t)| < 0.01$ rad, and (ii) the velocity is $|\dot{q}(t)| < 0.02$ rad/s. The corresponding results are displayed in Fig. 3.4 and Table 3.1.

	Fixed ξ	Fixed b	Variable b
Cost	0.068	0.048	0.042
Settling time (s)	1.800	1.460	1.300

Table 3.1: Cost and settling time for reaching with fixed $\xi = 1$, fixed damping and variable damping.

When looking at Table 3.1, we first notice that the set-up in which there is coupling between stiffness and damping ($\xi = 1$) produces the slowest response and incurs the greatest cost. This confirms the discussion in Sec. 3.4.3: because the

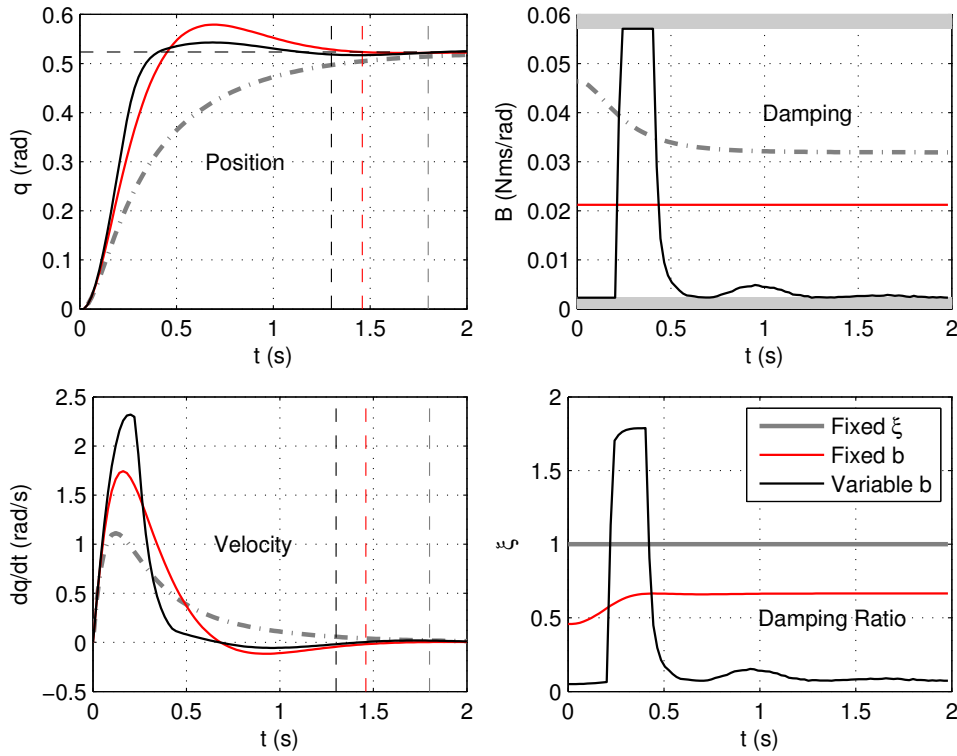


Figure 3.4: Joint position, velocity and damping profiles for reaching with (i) critical (ii) optimal fixed and (iii) optimal variable damping. The vertical dashed lines indicate the settling times for the three behaviours and the grey shaded area the damping bounds. The damping ratio over the duration of the movement is shown in the bottom right panel.

stiffness and damping are state and control dependent, $\xi = 1$ does not ensure critical damping, and in fact appears to result in an *overdamped response* in this case. For the scheme of optimal fixed damping the response is faster and (with a small overshoot) settles well before the 2 s time limit. We note that for the MACCEPA actuator (3.5) the joint stiffness drops as the link approaches equilibrium ($q \rightarrow q_0$), resulting in a decreasing stiffness torque τ_k and an inherent stabilisation effect.

From Table 3.1, we observe that the best performance in terms of cost and settling time is achieved by the optimal damping modulation scheme. Looking at the behaviour generated with this approach (Fig. 3.4), we notice how the device achieves faster convergence to the target, with less overshoot. The optimal damping profile makes use of the full range of damping (Fig. 3.4, top right) starting with low levels during the initial phase of the movement in order to achieve a high initial velocity (Fig. 3.4 bottom left), and then switching to the maximum damping as the target is approached in order to brake. The small overshoot that occurs causes the damping to be lowered once again (in order to facilitate return to the target),

before finally settling to a low level when the device has halted. The mentioned small overshoot is an artefact of our choice of cost function (3.18) which does not take into account the end velocity.

B Variable Damping and the Speed of Motor Dynamics

Having shown the benefit of optimal damping modulation when the equilibrium position and spring pre-tension are fixed, our next goal is investigating its effect when they are allowed to vary according to control. We therefore repeated the experiment from the preceding section, this time allowing the equilibrium position and stiffness (pre-tension) to be modulated. The latter were also optimised with respect to (3.18), within servo command ranges $u_1 \in [-\pi/3, \pi/3]$ rad and $u_2 \in [0, \pi/4]$ rad. The results are presented in the left column of Fig. 3.5.

From Fig. 3.5 (left column), we observed that, when equilibrium position and stiffness are allowed to vary in addition to damping, the ability to modulate the damping does not provide a significant improvement over having it fixed: the settling time for variable damping is 0.36s compared to 0.42s for fixed rate damping, and the costs are 0.0365 and 0.0368, respectively.

This close similarity in terms of performance is caused by the fact that in the case of this redundantly actuated system, the equilibrium position and stiffness can also be used to accelerate and brake the system, even if the damping is fixed. When presented with a constant rate damping command (Fig. 3.5, left column, thin black line) the system is able to accelerate by setting the position command beyond the target and increasing the stiffness. As the system is approaching the target, the command for the position is brought back in order to brake. When the system overshoots, the procedure is repeated until a stable level is achieved. This allows the variable stiffness system to compensate for having a fixed rate damping.

One disadvantage of relying on such a strategy, however, is that it expects the actuator to have the ability to very rapidly vary the equilibrium position and stiffness. In practice, this is not the case with the MACCEPA-VD: though damping modulation (i.e., modulating the PWM signal) is very fast (nearly instantaneous), the servomotors which control the position and stiffness modulation present much slower dynamics.

In order to investigate this more realistic scenario we repeated the experiment with a more accurate model that incorporates these delays in the servomotor response. Empirically we have found a third-order filter model (as described in [Wada et al., 2009]) to be a good model of the real servo responses. The results

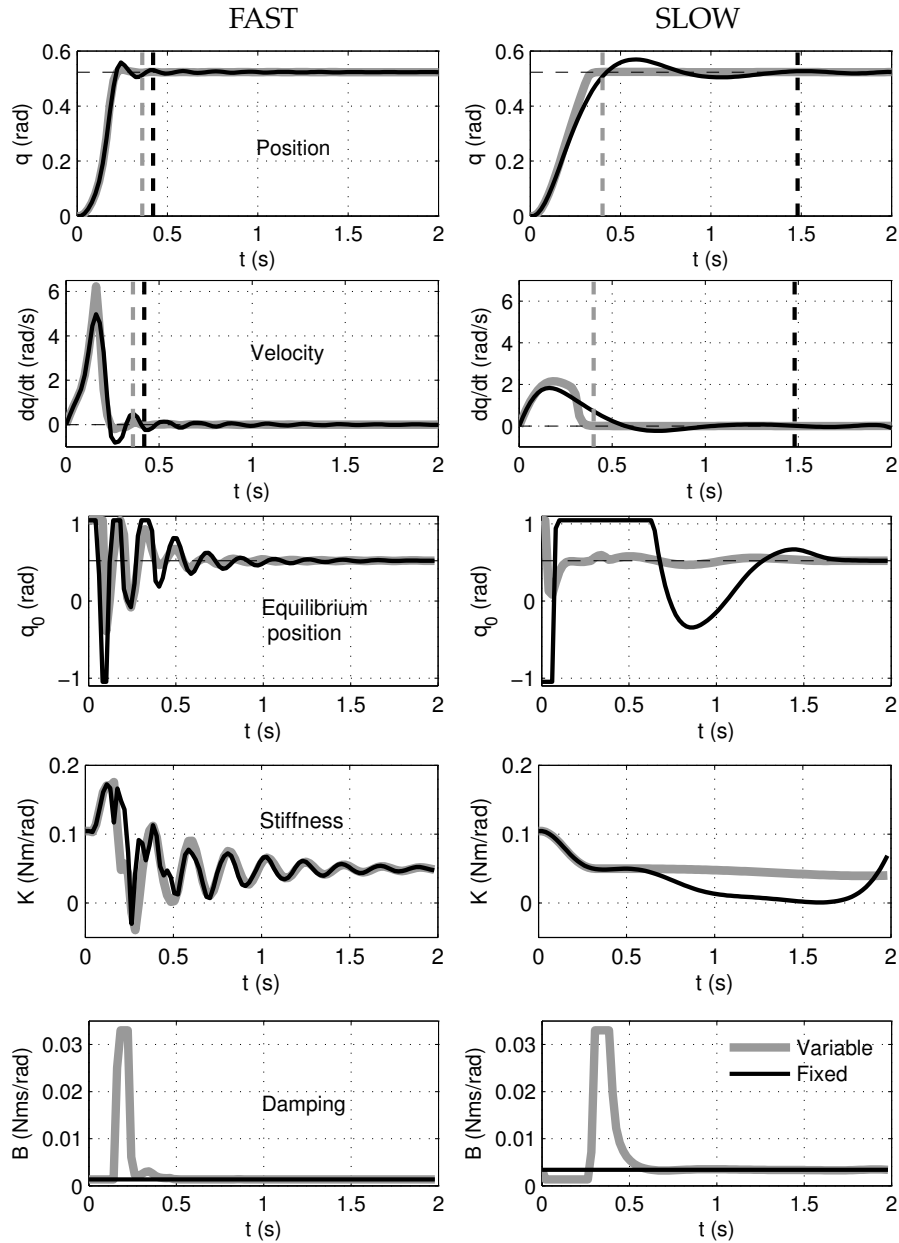


Figure 3.5: Joint position, velocity, equilibrium position, stiffness and damping profiles for reaching with (i) optimal fixed rate damping (black) and (ii) optimal variable damping (grey) for variable equilibrium position and spring pre-tension. Results for fast (left column) and slow (right column) motor dynamics.

obtained while repeating the optimisation with this model of the MACCEPA-VD are displayed in the right column of Fig. 3.5.

As it can be observed, there is now a much bigger difference in performance. Although the equilibrium position and stiffness commands are used in the same way, the slower motor dynamics hinder the fixed rate damping system from achieving the required braking for rapid settling at the target (Fig. 3.5, right column, thin

black line). In contrast, the variable damping system achieves comparable performance to that of the fast motor dynamics (settling time 0.40s) and much better than that of the fixed rate damping scheme (difference in settling time is 0.52s and improvement with respect to the cost is cca. 8.7%).

c Hardware Experiment

Lastly, we report experiments verifying the effectiveness of our proposed damping mechanism and show its performance in a rapid reaching task with a hardware implementation of the MACCEPA-VD.

Our implementation consists of two servomotors (Hitec HSR-5990TG) controlled with 50 Hz PWM signals from a micro-controller (Atmel ATmega328). Simultaneously, the micro-controller also controls an H-bridge circuit (NSC:LMD18200, as detailed in Sec. 3.4.5) which modulates the effective resistance between the terminals of a DC motor/gearbox unit (Maxon A-max 22/110125) - (see Fig. 3.2). The servomotor positions are measured from their internal potentiometers and the joint angle is measured with a rotary encoder (Melexis MLX90316GO).

1) *Passive Response Test:* Before proceeding with the rapid reaching task we first verify the performance of our variable damping mechanism. We perform this by testing its passive response with different damping commands. Namely, we conducted tests whereby the position and stiffness commands were fixed ($u_1 = \pi/3$ rad and $u_2 = \pi/2$ rad) and the actuator link was held (at rest) at $q = 0$ rad and then released. The recorded joint trajectories are plotted in Fig. 3.6 for damping commanded at the maximum ($u_3 = 1$), mid-range ($u_3 = 0.5$) and minimum ($u_3 = 0$) values. Overlaid are predictions of a model based on a system-identification of the device (for details on the latter, please see the Appendix B).

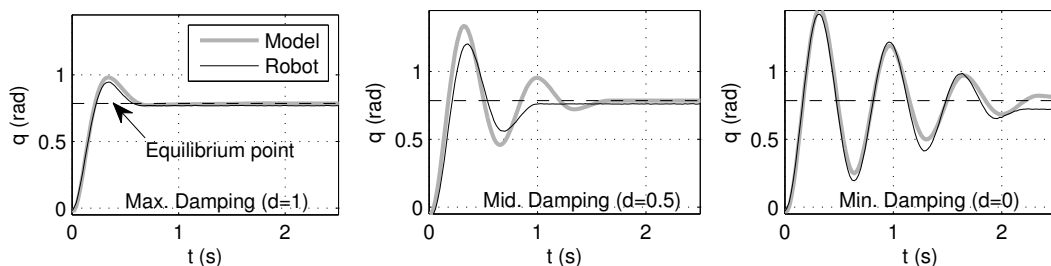


Figure 3.6: System response (black) and model predictions (grey) for maximum ($d = 1$), mid-range ($d = 0.5$) and minimum ($d = 0$) damping commands.

As it can be observed, a significant damping range is achieved with the proposed mechanism, approaching critical damping at its maximum (Fig. 3.6, left)

and an underdamped response at its minimum (Fig. 3.6, right). The tests also reveal a close fit between the model and the hardware response at each of the damping levels.

2) *Rapid Reaching in Hardware*: Finally, we present a comparison of the reaching behaviour of the MACCEPA-VD with fixed versus variable damping, as realised on the hardware system. In this experiment the target is set to $q^* = \pi/4$ rad and the maximum reaching duration $T = 1.5$ s.

The iLQR method was used to devise the optimal (open loop) command sequence for the three control inputs $(u_1, u_2, u_3)^T$, starting at rest with $q = 0$ rad and $u_1 = u_2 = 0$ rad. We compare the cases where the damping is (i) fixed at the optimal constant value and (ii) allowed to vary freely according to the optimisation. The results are plotted in Fig. 3.7. (A video recording of the experiment is available in the provided supplementary material).

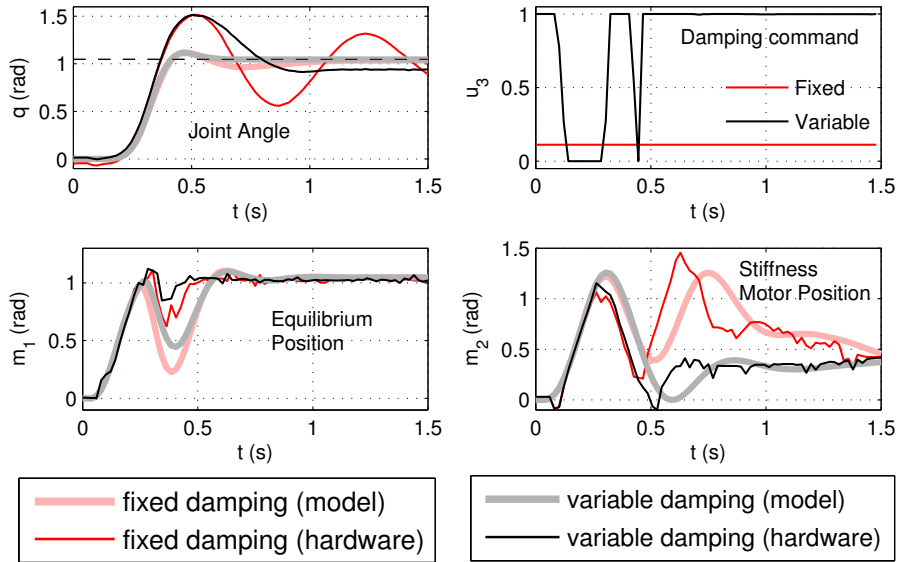


Figure 3.7: Model predictions (light, thick lines) and the real robot response (sharp, thin lines). Shown are joint angle, damping command and motor positions when reaching to a target of $q^* = \pi/4$ rad (black dashed line) on the MACCEPA-VD actuator with optimal (i) fixed (red lines) and (ii) variable damping (dark lines).

As it can be seen, (Fig. 3.7, top left), the ability of the system to reach and stabilise at the goal is heavily dependent on the whether or not the ability to modulate the damping is available. In both scenarios we observe that the solutions make use of all of the control degrees of freedom available (Fig. 3.7, motor positions).

However, only the system equipped with variable damping capabilities is able to stabilise within the 1.5s time period (Fig. 3.7, top left). We also note that, as pre-

dicted in Sec. 3.4.7.b, the fixed rate damping scheme tries to compensate by using the equilibrium position and pre-tension motors to brake. In the real device, the relatively slow dynamics of the servomotors prevent it from doing so successfully.

3.5 DISCUSSION AND CONCLUSIONS

In this chapter we introduced the concept of VIAs. We discussed the difficulties raised by the task of planning for systems equipped with them and presented an overview of previous approaches. We highlighted the benefits of their use, as outlined by preceding investigations.

A novel variable damping actuator for use in compliantly actuated systems was presented in [Laffranchi et al., 2010]. The device was used to remove unwanted oscillations occurring in such designs. In our work, we presented a mechanism for implementing variable damping in variable physical impedance devices. As opposed to the aforementioned device, this was the first time simultaneous and independent stiffness and damping modulations were used in a single actuator design.

We outlined how variable physical damping can be exploited through optimal control in the context of rapid movement tasks. The simulation and hardware experiments suggest a significant benefit of (i) variable damping over fixed ratio damping schemes and (ii) variable damping over variable stiffness when considering the relative speed at which the two can be modulated.

The investigation focused on a rapid reaching task for a single link system. Future work should extend these experiments, in order to illustrate the role of damping in multi-link variable impedance control and explore the relative benefits of using variable damping versus variable stiffness in the context of energy optimal behaviour.

In the next chapter we will build up on these results by addressing the optimal control problem of robotic systems with VIAs in more complex scenarios. We will present our optimisation framework, developed for handling the difficult domain of switching dynamics and discontinuous state transitions.

SYSTEMS WITH SWITCHING DYNAMICS

Classical robotic setups involve fixed base systems (e.g., industrial manipulators - Fig. 4.1, left). The methods addressed in the previous chapter can be easily applied there. Modern robotics, however, employs more complex, floating base hardware designs, such as locomotion robots (Fig. 4.1, right).

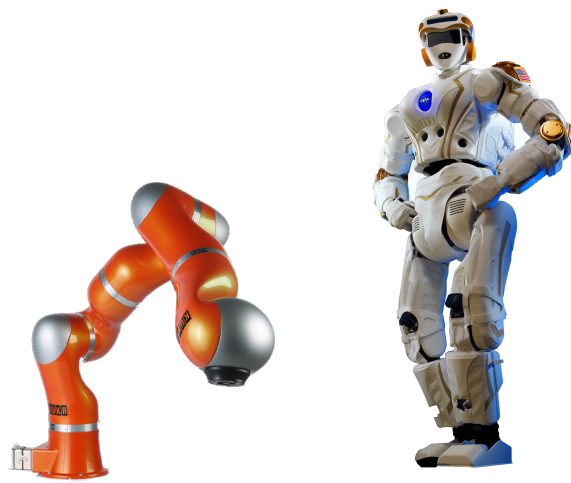


Figure 4.1: Fixed Base (left, KUKA lightweight robot (LWR) [Loughlin et al., 2007]) vs. Floating Base Systems (right, NASA Valkyrie [Radford et al., 2015]).

During their operation, these systems can exhibit a change in the number and positions of contacts with the environment. As a result, the complexity of the control and planning problem increases correspondingly.

In this chapter, we introduce the theoretical background for modelling and control of such systems, indicating currently uninvestigated aspects. We present our systematic methodology for movement optimisation with multiple phases, including contacts, for robotic systems with VIAs. We follow with a numerical evaluation of the proposed framework, in a multi-phase brachiation movement, with switching dynamics and a realistic VSA model.

To demonstrate the applicability of the proposed framework under various physical conditions, we present experimental results of brachiation tasks in a two-link brachiating robot with variable stiffness actuation developed in our laboratory.

The contacts between the robotic systems and the environment increase the complexity of the planning and control problem. Thus, we begin by giving an overview of the main approaches for modelling them.

4.1 CONTACT MODELLING

As two bodies come into contact, the generic collision between them is described by the forces acting (i) along the line of impact (normal to the tangential plane at the contact point [Meriam and Kraige, 2012]) and (ii) along the tangential plane [Brach, 2007]. The *coefficient of restitution* acts along the normal direction (i) and is a measure of the kinetic energy remaining at the end of the impact for the objects to recoil, compared to the energy dissipated during the impact [Brach, 2007]. It is formally defined as the ratio between the relative speed of the objects before and after the collision [McGinnis, 2005]. The tangential coefficients (ii) refer to the friction acting along the point (or area) of collision.

In dealing with contact analysis, the approaches are traditionally separated into two main categories: *discrete* and *continuous* methods [Gilardi and Sharf, 2002]. In the case of *discrete (impulse momentum)* methods, the impact is instantaneous. The forces are impulsive, the kinetic variables have discontinuous changes, while no displacement occurs during the impact and the other finite forces are deemed negligible. Parameters such as the coefficient of restitution and impact ratio are used to model the energy transfer [Wang and Mason, 1992]. On the tangential direction, kinematic constraints, securing that no displacement occurs during the impact, are introduced. Several models (e.g., Newton's [Whittaker, 1970], Poisson's [Routh et al., 1960], Stronge's [Stronge, 1991]) are available, differentiated by the way in which the aforementioned parameters are defined. The discrete formulation is mainly employed for rigid body collisions. It does not extend easily to general scenarios (e.g., multiple contact points) and relies on Coulomb's law for modelling friction, which has been reported to lead to inconsistencies or multiple solutions [Gilardi and Sharf, 2002].

In the case of *continuous (forced based)* methods, the interaction forces act continuously; thus, they can be inserted in the equation of motion, for the duration of the contact [Gilardi and Sharf, 2002]. The continuous analysis can be split into two approaches: i) considering the contact model as an explicit functional relation between the contact forces and the generalised coordinates and their rates [Deguet et al., 1998; Vukobratović and Potkonjak, 1999], or ii) taking into account

the deformation caused by the contact [Bathe and Bouzinov, 1997; Farahani et al., 2000]. The continuous formulation is not restricted to using Coulomb’s friction model and allows generalisation. However, it can require tuning of the contact parameters, which is non-trivial [Gilardi and Sharf, 2002].

Contacts with the environment introduce discontinuities in the dynamics of the robotic systems and several methods for modelling their effect have been developed. While *continuous methods* allow a better generalisation, they necessitate significant more tuning of the parameters than their *discrete* counterparts. Given a system operating in domains with contacts the question of control becomes more complicated. In the following section we discuss the main approaches for applying OC in such scenarios.

4.2 OPTIMAL CONTROL OF SYSTEMS IN DOMAINS WITH CONTACTS

The traditional OC formalism introduced in Sec. 2.2 is designed to handle smooth, continuous systems dynamics for tasks over a predetermined time horizon. In order to cope with discontinuities introduced by varying contacts, several modifications were developed. In the following section, we present some of the resulting approaches and their applications.

4.2.1 Hybrid Dynamics Formulation

Dynamics with intermittent contacts and impacts, such as locomotion and juggling, are often modelled as hybrid dynamical systems [Bätz et al., 2010; Grizzle et al., 2001; Long et al., 2011; Rosa Jr. et al., 2012]. This formulation consist of sets of piece-wise continuous dynamics (4.1) and discontinuous state transition (4.2), which encode the effect of a contact event:

$$\dot{\mathbf{x}} = \mathbf{f}_i(\mathbf{x}, \mathbf{u}), \quad i \in I = \{1, 2, \dots, M\}, \quad (4.1)$$

$$\mathbf{x}^+ = \Delta(\mathbf{x}^-), \quad (4.2)$$

where $\mathbf{f}_i : \mathbb{R}^n \times \mathbb{R}^m \rightarrow \mathbb{R}^n$ is the i -th continuous dynamics function, $\mathbf{x} \in \mathbb{R}^n$ is the state vector, $\mathbf{u} \in \mathbb{R}^m$ is the control input vector and I is the set of indices for the continuous dynamics functions. When the dynamics switch from function \mathbf{f}_{i-1} to \mathbf{f}_i an instantaneous discontinuous state transition is assumed according to

the impact map Δ which maps a pre-impact condition (\mathbf{x}^-) to a post-impact state (\mathbf{x}^+).

Traditionally, when discussing the switching dynamics of the hybrid dynamical systems, two classes emerge: time-based and state-based switching formulation. In the time-based switching case, the discontinuous state jump occurs at a predefined time : $\dot{\mathbf{x}} = \mathbf{f}_i(\mathbf{x}, \mathbf{u})$ for $t_{i-1} \leq t < t_i$. When the switching time is met, the mapping Δ acts upon the state.

In state-based switching, the discontinuous jump arises when certain state conditions are met. These systems can be more formally described using hybrid automata, which is an extension of the finite-state machine formalism [Lunze et al., 2009]. Each discrete state is associated with a differential (or difference) equation, describing the continuous evolution of the system in that state. Depending on the complexity of the system, the encoding of the switching rules can be quite elaborate.

4.2.2 Control Approaches using Hybrid Dynamics

From a control theoretic perspective, a significant effort was made to address optimal control problems of various classes of hybrid systems [Branicky et al., 1998; Shaikh and Caines, 2007; Sussmann, 1999; Xu and Antsaklis, 2002, 2003a].

An extension of the Pontryagin's Minimum Principle (Sec. 2.2), was developed for the hybrid setting in the form of the Hybrid Minimum Principle (HMP) [Piccoli, 1998]. The HMP was applied to specific classes of systems, while considering impact maps as either controls or constraints. By introducing realistic constraints on the dynamics, the validity of the solution (for direct application on a hardware platform) was ensured.

In [Long et al., 2011] state based switching is used and impacts were treated as instantaneous state transition. The open-loop solution was obtained by first order gradient descent. The convergence rate is much lower than that offered by second order methods but, obtaining an explicit expression for the second order derivative proved too exhausting even for simple cases, as the authors note.

Solving hybrid optimal control problems with state based switching is challenging even if the number and the sequence of switching are known a priori. This is partially due to the need for additional constraints ensuring the states lie on the switching surface at the time of switching. Finding this time is an ambitious task,

as it depends on the control commands and finding an analytic expression for it is generally difficult [Xu and Antsaklis, 2003b].

In spite of the substantial amount of research on hybrid dynamics, examples in the OC literature are confined to low-dimensional and simple dynamical systems, and only a few robotic applications can be found for optimisation of movements over multiple phases [Buss et al., 2002; Long et al., 2011].

4.2.3 *Alternative Approaches to using Hybrid Dynamics*

Instead of using hybrid dynamics modelling, different optimisation approaches were proposed for dealing with multiple contact events, such as model predictive control (MPC) [Camacho and Alba, 2013]. The method delivers an optimal solution for a finite time horizon, but only applies it for the first step, after which the process is repeated.

For example, [Erez et al., 2011] combines offline trajectory optimisation and on-line MPC. The limit cycle of a periodic movement is found by offline optimisation with an infinite-horizon average cost, while on-line MPC is used to obtain an optimal feedback control law. Another study [Kulchenko and Todorov, 2011], generalises MPC from the usual finite horizon (or receding horizon) setup to a first-exit setting (i.e., a solution is found based on the assumption of an exit state), which avoids dealing with discontinuities in the on-line optimisation phase. A modified iLQG (Sec. 2.3.1) algorithm is used for on-line optimisation by running the forwards pass not for a predefined number of steps, but until it hits a terminal state. However, as observed in [Rawlik, 2013], first-exit strategies can result in a potentially suboptimal solution.

The work in [Tassa et al., 2012] proposes an iLQG based MPC with smooth approximation of contact forces without the need of switching dynamics. In [Posa et al., 2014], the contact forces are explicitly included as constraints (using complementarity conditions) and directly optimised, together with the trajectories and control commands, using sequential quadratic programming. As noted in the previous section, such state-based switching approaches can be very challenging, as they require introducing additional constraints.

In [Morimoto and Atkeson, 2003] Minimax Differential Dynamic Programming is used to find the control policy for a bipedal robotic system. The optimisation is performed over the continuous dynamics for one step of walking. An impulse map is used to compute the velocity change at the ground contact in the terminal cost

in order to ensure the periodicity of the trajectory (heel strike is always assumed to occur at the fixed final time). However, the cost function is formulated for tracking a given nominal trajectory obtained by a hand-tuned controller, while minimising the control cost.

So far we have reviewed the major challenges posed by the introduction of changing contacts for the planning and control problem. We have showed how these contacts can be encoded and presented several approaches used to obtain an OC solution for systems operating under these conditions. We have not yet addressed the added difficulty of undertaking this task for systems with VIAs.

4.3 OPTIMAL IMPEDANCE MODULATION AND PASSIVE DYNAMICS EXPLOITATION IN DOMAINS WITH CONTACTS

For robotic systems performing a periodic task, actuators equipped with mechanical compliance allow for energy storage [Stramigioli et al., 2008]. Exploiting the compliance of the mechanical design reduces the energy input of the motor. In [Sreenath et al., 2011] the energy consumption of the bipedal robot MABEL is lowered by using the compliance caused by negative work at the time of impact, instead of activating the actuators. In robot running, [Karssen and Wisse, 2011] presents numerical studies to demonstrate that an optimised nonlinear leg stiffness profile could improve robustness against disturbances. The importance of exploitation of the intrinsic passive dynamics for efficient actuation and control is discussed in the study of passive dynamic walking, where biped robots with no or minimal actuation can exhibit human-like natural walking behaviour [Collins et al., 2005].

Hybrid dynamics are an established way of describing the behaviour of realistic robotic systems operating in domains with contacts. In our work we employ this formulation alongside time-based switching for encoding the system dynamics. We choose time-based switching with the aim of improving the tractability of the optimisation problem (due to the difficulties raised by the state-based switching formulation). The first system we use to showcase the benefits of our approach is a two-link brachiation robot.

The brachiation system incorporates all the elements we are interested in addressing in our work (i.e., switching dynamics, caused by contacts with the envi-

ronment, VIA capabilities). Thus, it constitute a good test platform for showcasing the benefits of our developed framework.

Next, we present relevant work conducted on robotic brachiation, in which the system uses a form of locomotion inspired by that of primates, defined by swinging from handhold to handhold like a pendulum [Eimerl and DeVore, 1966; Saito et al., 1994]. Explicit exploitation of the passive dynamics with the help of gravity is required in order to achieve the task.

CONTROL OF BRACHIATION SYSTEMS Designing a brachiating controller is a challenging problem since the system is underactuated (i.e., there is no actuation at the handhold gripper). Significant efforts were made regarding control for a class of underactuated systems, especially from a theoretical perspective [Luca and Oriolo, 2002; Lynch et al., 2000; Nakamura et al., 1997; Spong, 1995]. A significant proportion this work [Lynch et al., 2000; Nakamura et al., 1997] focused on motion planning for underactuated manipulators in a horizontal plane, and thus not focusing on the influence of gravity. In these scenarios, the dynamic coupling of link inertia is exploited during movement, rather than the passive dynamics under gravity, as in the case of manipulators moving in the vertical plane.

In [Nakanishi et al., 2000] a brachiation system is controlled using active cancellation of the plant dynamics with input-output linearisation to force the robot to mimic the specified pendulum-like motion described in terms of target dynamics. In [Nakanishi et al., 2011] temporal and stiffness optimisation for a brachiation task are shown to result in lower control effort. In [Meghdari et al., 2013], a multiple shooting method is used to find optimal trajectories and the final time for a single swing movement. The resulting behaviour is shown to reduce the control effort in comparison to the method proposed in [Nakanishi et al., 2000].

In [Gomes and Ruina, 2005] passive brachiation is achieved by finding numerical solutions for the initial conditions which lead to periodically continuous locomotion without any joint torques. The work in [Rosa Jr. et al., 2012] extends these results by investigating open-loop stable (unactuated downhill and powered uphill) brachiation of a two-link model in the context of hybrid systems control, including switching and discontinuous transitions. Motivated by this work [Gomes and Ruina, 2005], we show that highly dynamic tasks, such as brachiation, can be achieved by exploiting passive dynamics with *simultaneous* stiffness and temporal optimisation in a compliantly actuated robot.

4.4 DEVELOPED FRAMEWORK OUTLINE

As discussed in the previous sections, although a significant amount of research addressed OC of hybrid systems, few applications focusing on movement optimisation over multiple switching events can be found and optimal impedance modulation has not come under close study.

In our work we address these issues. We have already used OC of VIA, in the form of stiffness and damping modulation in a single actuator, for a fixed base robotic system (Sec. 3.4). We extend this research to robotic systems in domains with contacts. Since the actuator is equipped with mechanical compliance, it can be used to store energy [Stramigioli et al., 2008].

Thus, we suggest an approximate approach to the hybrid optimal control problem. We model the multiple phases of movement as time-based switching hybrid dynamics assuming that the sequence of switching is known. In this set-up, we address the following aspects:

1. Time-based switching dynamics with continuous control input
2. Discrete state transition
3. Realistic plant dynamics with a VSA model
4. Composite cost function for a multi-phase movement task
5. Optimisation of control command and stiffness
6. Optimisation of switching instances
7. Optimisation of final time (total movement duration)

The proposed formulation also delivers an optimal *feedback* control law, while many trajectory optimisation algorithms typically provide only optimal *feed forward* controls. While elements addressed in our approach have been treated separately in previous works [Caldwell and Murphey, 2011a; Egerstedt et al., 2003; Xu and Antsaklis, 2004], this is, to our knowledge, the first investigation to consider all these issues at the same time.

The rest of this chapter is organised as follows: first, the hybrid optimal control problem with time-based switching is formulated. Next, we present an iterative solution method by extending the iterative linear quadratic regulator (iLQR) algorithm [Li and Todorov, 2007] (Sec. 2.3.1), in order to incorporate switching dynamics and discrete state transition. In addition, we present a temporal optimisation algorithm for hybrid dynamics.

In the evaluation, we start by introducing a two-link brachiation system equipped with a VSA with switching dynamics and discrete state transition. We showcase the benefits and the effectiveness of our approach with experiments in simulation and on a real hardware implementation.

4.5 OPTIMISATION FRAMEWORK FORMULATION

4.5.1 Hybrid Dynamics with Time-Based Switching and Discrete State Transition

Consider a class of hybrid systems, as introduced in Sec. 4.2.1. Employing the time based switching formulation [Xu and Antsaklis, 2003b] (4.1), (4.2) can be rewritten in a more intuitive representation as in [Caldwell and Murphey, 2011b]:

$$\dot{\mathbf{x}} = \mathbf{f}_{i_j}(\mathbf{x}, \mathbf{u}), \quad T_j \leq t < T_{j+1} \quad (4.3)$$

$$\mathbf{x}(T_j^+) = \Delta(\mathbf{x}(T_j^-)), \quad (4.4)$$

with the finite number of switching given as K and the timed switching sequence σ in $[T_0, T_f]$ is defined as [Xu and Antsaklis, 2003b] :

$$\sigma = ((T_0, i_0), (T_1, i_1), \dots, (T_K, i_K)), \quad (4.5)$$

where $T_0 \leq T_1 \leq \dots \leq T_K \leq T_f = T_{K+1}$ (i.e., monotonically increasing switching instances) and $i_j \in I$ for $j = 0, \dots, K$. At $t = T_j$, the subsystem switches from i_{j-1} to i_j and subsystem i_j is active for $T_j \leq t < T_{j+1}$. In our work, the sequence of switching is assumed to be given. Figure 4.2 depicts a schematic diagram of a hybrid system we consider in this investigation.

4.5.2 Robot Dynamics with Variable Stiffness Actuation

Using the general formulation introduced above, an individual rigid body dynamics model (Sec. 3.2) is defined for each phase of the movement. Thus, we consider the robot dynamics with variable stiffness actuation from (3.1):

$$\mathbf{M}_i(\mathbf{q})\ddot{\mathbf{q}} + \mathbf{C}_i(\mathbf{q}, \dot{\mathbf{q}})\dot{\mathbf{q}} + \mathbf{g}_i(\mathbf{q}) + \mathbf{D}_i\dot{\mathbf{q}} = \boldsymbol{\tau}(\mathbf{q}, \mathbf{q}_m), \quad (4.6)$$

where i denotes the i -th subsystem. At the same time, the positions of the motors (\mathbf{q}_m) are not always instantaneously modulated though the motor commands (\mathbf{u}),

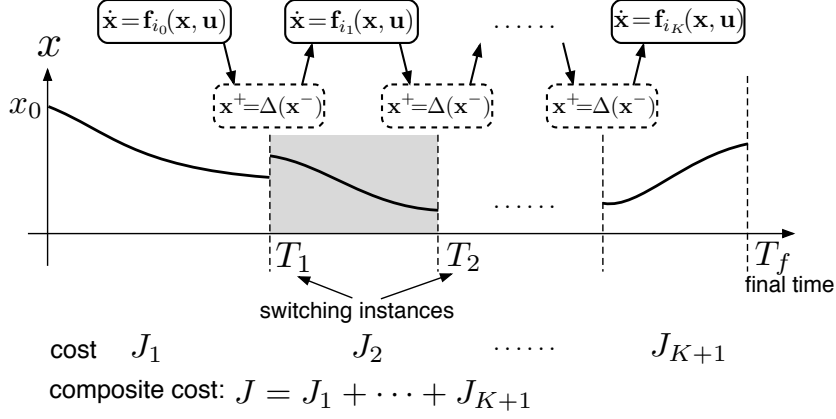


Figure 4.2: A hybrid system with time-based switching dynamics and discrete state transition with a known sequence. The goal is finding an optimal control command \mathbf{u} , switching instances T_i and final time T_f which minimises the composite cost $J = J_1 + \dots + J_{K+1}$.

depending on the particularities of the implementation. In general the servo motor dynamics can be modelled as a critically damped second order dynamical system:

$$\ddot{\mathbf{q}}_m + 2\alpha_i \dot{\mathbf{q}}_m + \alpha_i^2 \mathbf{q}_m = \alpha_i^2 \mathbf{u}, \quad (4.7)$$

where α determines the bandwidth of the servo motors¹ [Braun et al., 2012a]. We assume that the range of control command \mathbf{u} is limited between \mathbf{u}_{\min} and \mathbf{u}_{\max} as defined in (3.2).

Given the previously described robotic system with VSA, an adequate state-space representation would include both the robot dynamics (4.6) and the closed-loop motor dynamics (4.7). In this representation, the state-space representation (3.8) is reformulated as:

$$\dot{\mathbf{x}} = \mathbf{f}_i(\mathbf{x}, \mathbf{u}) = \begin{bmatrix} \mathbf{x}_2 \\ \mathbf{M}_i^{-1}(\mathbf{x}_1)(-\mathbf{C}_i(\mathbf{x}_1, \mathbf{x}_2)\mathbf{x}_2 - \mathbf{g}_i(\mathbf{x}_1) - \mathbf{D}_i\dot{\mathbf{x}}_2 + \tau(\mathbf{x}_1, \mathbf{x}_3)) \\ \mathbf{x}_4 \\ -\alpha^2 \mathbf{x}_3 - 2\alpha \mathbf{x}_4 + \alpha^2 \mathbf{u} \end{bmatrix} \quad (4.8)$$

and $\mathbf{x} = [\mathbf{x}_1^T, \mathbf{x}_2^T, \mathbf{x}_3^T, \mathbf{x}_4^T]^T = [\mathbf{q}^T, \dot{\mathbf{q}}^T, \mathbf{q}_m^T, \dot{\mathbf{q}}_m^T]^T \in \mathbb{R}^{2(n+m)}$. By incorporating both the rigid body dynamics in (4.6) and the servo motor dynamics of the VSA (4.7), the state of the systems is fully represented.

¹ $\alpha = \text{diag}\{\alpha_1, \dots, \alpha_m\}$ and $\alpha^2 = \text{diag}\{\alpha_1^2, \dots, \alpha_m^2\}$ for notational convenience.

4.5.3 Multi Phase Movement Cost Function

For the given dynamics we consider the composite cost function:

$$J = \phi(\mathbf{x}(T_f)) + \sum_{j=1}^K \psi^j(\mathbf{x}(T_j^-)) + \int_{T_0}^{T_f} r(\mathbf{x}, \mathbf{u}) dt \quad (4.9)$$

to describe the full movement with multiple phases, where $\phi(\mathbf{x}(T_f))$ is the terminal cost, $\psi^j(\mathbf{x}(T_j^-))$ is the via-point cost at the j -th switching instance and $r(\mathbf{x}, \mathbf{u})$ is the running cost.

Using traditional OC approaches, each cost function can be (locally) optimised, however, the total cost $J = \sum_{j=1}^{K+1} J_j$ may be suboptimal. Under sequential optimisation of separate cost functions, the via-point states for the j -th sequence may not be necessarily optimal depending on the choice of the via-point cost ψ^j . This is because separate optimisation does not consider future consequences of optimisation beyond the time horizon of T_{j+1} when minimising the j -th cost function.

For a given set of plant dynamics (4.3) and state transition (4.4), the optimisation problem we consider is to i) find an optimal feedback control law $\mathbf{u} = \mathbf{u}(\mathbf{x}, t)$ which minimises the composite cost (4.9) and ii) simultaneously optimise the switching instances T_1, \dots, T_k and the final time T_f .

4.5.4 Developed Multi-Phase Optimisation Method

In this section, we first extend iLQR (deterministic version of iLQG presented in Sec. 2.3.1 [Li and Todorov, 2007]²) in order to incorporate timed switching dynamics with discrete and discontinuous state transitions. We then present a temporal optimisation algorithm to optimise the switching instances and the total movement duration.

A Optimal Control of Switching Dynamics and Discrete State Transition

The iLQR method solves an optimal control problem by performing a locally linear quadratic approximation of the nonlinear dynamics and the cost function around a nominal trajectory $\bar{\mathbf{x}}$ and control sequence $\bar{\mathbf{u}}$ in discrete time, and iteratively improves the solutions.

² In this work, we assume deterministic plant dynamics.

In order to incorporate switching dynamics and discrete state transition with a given switching sequence σ (4.5), the hybrid dynamics (4.3) and (4.4) are linearised in discrete time around the nominal trajectory and control sequence as:

$$\delta \mathbf{x}_{k+1} = \mathbf{A}_k \delta \mathbf{x}_k + \mathbf{B}_k \delta \mathbf{u}_k \quad (4.10)$$

$$\delta \mathbf{x}_{k_j}^+ = \Gamma_{k_j} \delta \mathbf{x}_{k_j}^- \quad (4.11)$$

$$\mathbf{A}_k = \mathbf{I} + \Delta t \left. \frac{\partial f_{i_j}}{\partial \mathbf{x}} \right|_{\mathbf{x}=\mathbf{x}_k}, \quad \mathbf{B}_k = \Delta t \left. \frac{\partial f_{i_j}}{\partial \mathbf{u}} \right|_{\mathbf{u}=\mathbf{u}_k} \quad (4.12)$$

$$\Gamma_{k_j} = \left. \frac{\partial \Delta}{\partial \mathbf{x}} \right|_{\mathbf{x}=\mathbf{x}_{k_j}^-} \quad (4.13)$$

where $\delta \mathbf{x}_k = \mathbf{x}_k - \bar{\mathbf{x}}_k$ and $\delta \mathbf{u}_k = \mathbf{u}_k - \bar{\mathbf{u}}_k$, k is the discrete time step and Δt is the sampling time, k_j is the j -th switching instance in the discredited time step.

The composite cost function (4.9) is locally approximated in a quadratic form as:

$$\begin{aligned} \delta J = & \delta \mathbf{x}_N^T \phi_x + \frac{1}{2} \delta \mathbf{x}_N^T \phi_{xx} \delta \mathbf{x}_N + \sum_{j=1}^K (\delta \mathbf{x}_{k_j}^T \psi_x^j + \sum_{k=1}^N (\delta \mathbf{x}_k^T r_x + \delta \mathbf{u}_k^T r_u \\ & + \frac{1}{2} \delta \mathbf{x}_k^T r_{xx} \delta \mathbf{x}_k + \frac{1}{2} \delta \mathbf{u}_k^T r_{uu} \delta \mathbf{u}_k + \delta \mathbf{u}_k r_{ux} \delta \mathbf{x}_k) \Delta t \end{aligned} \quad (4.14)$$

and the optimal cost-to-go function is locally approximated as:

$$v_k(\delta \mathbf{x}_k) = \frac{1}{2} \delta \mathbf{x}_k^T \mathbf{S}_k \delta \mathbf{x}_k + \delta \mathbf{x}_k^T \mathbf{s}_k. \quad (4.15)$$

The local control law:

$$\delta \mathbf{u}_k = \mathbf{I}_k + \mathbf{L}_k \delta \mathbf{x}_k \quad (4.16)$$

is obtained from the Bellman equation:

$$v_k(\delta \mathbf{x}_k) = \min_{\delta \mathbf{u}} \{r_k(\delta \mathbf{x}_k, \delta \mathbf{u}_k) + v_{k+1}(\delta \mathbf{x}_{k+1})\}, \quad (4.17)$$

by substituting (4.10) and (4.15) into the equation (4.17), where r_k is the local approximation of the running cost in (4.14).

The cost-to-go parameters $\mathbf{S}_k, \mathbf{s}_k$ in (4.15) are updated with a modified Riccati equations: ³

$$\mathbf{S}_k = r_{xx} + \mathbf{A}_k^T \mathbf{S}_{k+1} \mathbf{A}_k + \mathbf{L}_k^T \mathbf{H} \mathbf{L}_k + \mathbf{L}_k^T \mathbf{G} + \mathbf{G}^T \mathbf{L}_k, \quad (4.18)$$

$$\mathbf{s}_k = r_x + \mathbf{A}_k^T \mathbf{s}_{k+1} + \mathbf{L}_k^T \mathbf{H} \mathbf{l}_k + \mathbf{L}_k^T \mathbf{g} + \mathbf{G}^T \mathbf{l}_k, \quad (4.19)$$

with:

$$\mathbf{H} = r_{uu} + \mathbf{B}_k^T \mathbf{S}_{k+1} \mathbf{B}_k, \quad (4.20)$$

$$\mathbf{G} = r_{ux} + \mathbf{B}_k^T \mathbf{S}_{k+1} \mathbf{A}_k, \quad (4.21)$$

$$\mathbf{g} = r_u + \mathbf{B}_k^T \mathbf{s}_{k+1}. \quad (4.22)$$

At the instance of discrete state transition $k = k_j$, the following cost-to-go parameter update is added:

$$\mathbf{S}_{k_j}^- = \psi_{xx}^j + \Gamma_{k_j}^T \mathbf{S}_{k_j}^+ \Gamma_{k_j}, \quad (4.23)$$

$$\mathbf{s}_{k_j}^- = \psi_x^j + \Gamma_{k_j}^T \mathbf{s}_{k_j}^+, \quad (4.24)$$

which is derived from the Bellman equation:

$$v_{k_j}(\delta \mathbf{x}_{k_j}^-) = cost_{k_j} + v_{k_j}(\delta \mathbf{x}_{k_j}^+) \quad \text{at } k = k_j, \quad (4.25)$$

where $\psi_{xx}^j(\delta \mathbf{x}_{k_j}^-)$ is the local approximation of the via-point cost in (4.14).

Once we have a locally optimal control command $\delta \mathbf{u}$, the nominal control sequence is updated as $\bar{\mathbf{u}} \leftarrow \bar{\mathbf{u}} + \delta \mathbf{u}$. Then, the new nominal trajectory $\bar{\mathbf{x}}$ is computed by running the obtained control $\bar{\mathbf{u}}$ and the above process is iterated until convergence.

B Temporal Optimisation

In [Rawlik et al., 2010], a method is proposed for simultaneously optimising the control commands and temporal parameters. A mapping $\beta(t)$ from the real time t to a canonical time t' is introduced as:

$$t' = \int_0^t \frac{1}{\beta(s)} ds, \quad (4.26)$$

³ At the final time $k = N$, $\mathbf{S}_N = \phi_{xx}$ and $\mathbf{s}_N = \phi_x$.

and $\beta(t)$ is optimised to scale the temporal aspect of the movement. This algorithm is employed [Nakanishi et al., 2011; Nakanishi and Vijayakumar, 2013] for optimising the frequency of periodic movement and the movement duration of swing locomotion in a brachiation task. In these problems, (4.26) is dispensed with, under the assumption that $\beta(t)$ and Δt are constant throughout the movement as:

$$\Delta t' = \frac{1}{\beta} \Delta t. \quad (4.27)$$

In our work, in order to optimise the switching instance and the total movement duration, we introduce a scaling parameter and sampling time for each duration of between switching as:

$$\Delta t'_j = \frac{1}{\beta_j} \Delta t_j \quad \text{for } T_j \leq t < T_{j+1} \quad \text{where } j = 0, \dots, K. \quad (4.28)$$

By optimising the vector of temporal scaling factors $\boldsymbol{\beta} = [\beta_0, \dots, \beta_K]^T$ via gradient descent:

$$\boldsymbol{\beta} \leftarrow \boldsymbol{\beta} - \eta \nabla_{\boldsymbol{\beta}} J, \quad (4.29)$$

where $\eta > 0$ is a learning rate, or the Newton's method:

$$\boldsymbol{\beta} \leftarrow \boldsymbol{\beta} - (\nabla_{\boldsymbol{\beta}}^2 J)^{-1} \nabla_{\boldsymbol{\beta}} J, \quad (4.30)$$

we can obtain each switching instance:

$$T_{j+1} = (k_{j+1} - k_j) \Delta t'_j + T_j \quad (4.31)$$

and total movement duration:

$$T_f = \sum_{j=0}^K (k_{j+1} - k_j) \Delta t'_j + T_0, \quad (4.32)$$

where $k_0 = 1$ and $k_{K+1} = N$.

In the complete optimisation, computation of optimal feedback control law and temporal scaling parameter update are iteratively performed, until convergence, in an EM-like (expectation-maximisation) manner. A pseudocode of the complete procedure is summarised in Algorithm 2.

Algorithm 2 Complete optimisation algorithm for hybrid dynamics with temporal optimisation

- 1: **Input:**
 - Timed switching plant dynamics f_i (4.3), discrete state transition Δ (4.4) and switching sequence σ (4.5)
 - Composite cost function J (4.9)
 - 2: **Initialise:**
 - Nominal switching instance and final time T_1, \dots, T_K and T_f and corresponding discrete sampling time $\Delta t_0, \dots, \Delta t_K$ as in (4.31) and (4.32)
 - Nominal control sequence $\bar{\mathbf{u}}$ and corresponding $\bar{\mathbf{x}}$
 - 3: **repeat**
 - 4: **repeat**
 - 5: **Optimise control sequence $\bar{\mathbf{u}}$:**
 - Linearise dynamics (4.10) and state transition (4.11) around $\bar{\mathbf{x}}$ and $\bar{\mathbf{u}}$ in discrete time with current Δt_j
 - Compute quadratic approximation of the composite cost (4.14)
 - Solve local optimal control problem to obtain $\delta \mathbf{u}$ (4.16)
 - Apply $\delta \mathbf{u}$ to (4.10) and (4.14)
 - Update nominal control sequence $\bar{\mathbf{u}} \leftarrow \bar{\mathbf{u}} + \delta \mathbf{u}$, trajectory $\bar{\mathbf{x}}$ and cost J
 - 6: **until** convergence
 - 7: **Temporal optimisation: update Δt_j :**
 - Update the temporal scaling factor β and corresponding sampling time $\Delta t_0, \dots, \Delta t_K$ in (4.28) via gradient descent or Newton's method
 - Obtain the switching instances T_j (4.31) and final time T_f (4.32)
 - 8: **until** convergence
 - 9: **Output:**
 - Optimal feedback control law $\mathbf{u}(\mathbf{x}, t)$: feedforward optimal control sequence \mathbf{u}_{opt} , optimal trajectory $\mathbf{x}_{\text{opt}}(t)$ and optimal gain matrix $\mathbf{L}_{\text{opt}}(t)$: $\mathbf{u}(\mathbf{x}, t) = \mathbf{u}_{\text{opt}}(t) + \mathbf{L}_{\text{opt}}(t)(\mathbf{x}(t) - \mathbf{x}_{\text{opt}}(t))$
 - Optimal switching instance T_1, \dots, T_K and final time T_f
 - Optimal composite cost J
-

4.6 BRACHIATING ROBOT DYNAMICS WITH VSA

The equation of motion of the two-link brachiating robot shown in Fig.4.3 (See Sec. 4.9 for the description of the hardware platform) takes the standard form of rigid body dynamics (4.6) where only the second joint has actuation:

$$\mathbf{M}_i(\mathbf{q})\ddot{\mathbf{q}} + \mathbf{C}_i(\mathbf{q}, \dot{\mathbf{q}})\dot{\mathbf{q}} + \mathbf{g}_i(\mathbf{q}) + \mathbf{D}_i\dot{\mathbf{q}} = \begin{bmatrix} 0 \\ \tau(\mathbf{q}, \mathbf{q}_m) \end{bmatrix}, \quad (4.33)$$

with $\mathbf{q} = [q_1, q_2]^T$ the joint angle vector, maintaining the definitions given for (3.1) and (4.7). The index i denotes the link number corresponding to Table 4.1.

We use MACCEPA (Fig. 4.3) as our VSA implementation of choice 4. This actuator design has the desirable property that the joint can be passively compliant. This allows free swinging with a large range of movement by relaxing the spring, highly suitable for the brachiation task we consider. MACCEPA is equipped with two position controlled servo motors, $\mathbf{q}_m = [q_{m1}, q_{m2}]^T$, which control the equilibrium position and the spring pre-tension, respectively.

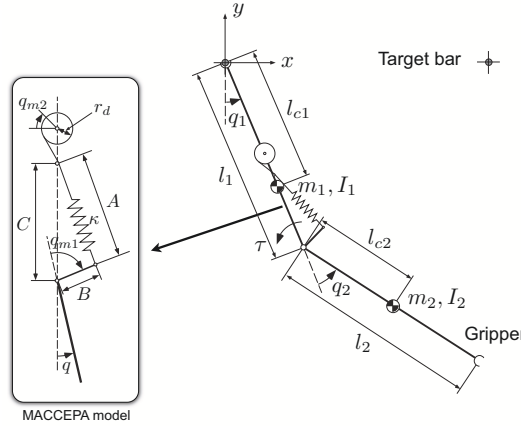


Figure 4.3: Two-link brachiating robot model with the VSA joint with the inertial and geometric parameters. See Sec. 4.9 for the description of the hardware platform. The parameters are given in Table 4.1 (The index i denotes the link number corresponding to Table 4.1).

We include position controlled servo motor dynamics, approximated by a second order system with a PD feedback control as defined in (4.7). For the motors bandwidth parameter we use $\alpha_1 = \alpha_2 = \text{diag}(20, 25)$. The range of the commands of the servo motors are limited as $u_1 \in [-\pi/2, \pi/2]$ and $u_2 \in [0, \pi/2]$.

The joint torque for this actuator model is given by:

$$\tau(q, q_{m1}, q_{m2}) = \kappa_s BC \sin \alpha \left(1 + \frac{r_d q_{m2} - (C - B)}{\sqrt{B^2 + C^2 - 2BC \cos \alpha}} \right). \quad (4.34)$$

and the joint stiffness can be computed as:

$$k(q, q_{m1}, q_{m2}) = -\frac{\partial \tau}{\partial q} = \kappa_s \cos(\alpha) BC \left(1 + \frac{\beta}{\gamma} \right) - \frac{\kappa_s \sin^2(\alpha) B^2 C^2 \beta}{\gamma^3}, \quad (4.35)$$

⁴ We introduced the MACCEPA design for the first time in 3.4.2. Here we reiterate the main elements, including realistic delays in the servomotors response (4.7). Hence in the behaviour of the MACCEPA $\mathbf{u} \rightarrow \mathbf{q}_m$.

where κ_s is the spring constant, r_d is the drum radius, $\alpha = q_{m1} - q$, $\beta = r_d q_{m2} - (C - B)$ and $\gamma = \sqrt{B^2 + C^2 - 2BC \cos(q_{m1} - q)}$ (see Fig. 4.3 for the definition of the model parameters). The spring tension is given by:

$$F = \kappa_s(l - l_0), \quad (4.36)$$

where $l = \sqrt{B^2 + C^2 - 2BC \cos(q_{m1} - q)} + r_d q_{m2}$ is the current spring length and $l_0 = C - B$ is the spring length at rest. The joint torque equation (4.34) can also be rearranged in terms of the moment arm and the spring tension as:

$$\tau = \frac{BC \sin(q_{m1} - q)}{\gamma} F. \quad (4.37)$$

We note that MACCEPA has a relatively simple configuration in terms of actuator design compared to other VSAs. However, the torque and stiffness relationships in (4.34) and (4.35) are dependent on the current joint angle and two servo motor angles in a complicated manner and its control is not straightforward.

We use the model parameters of the hardware platform (see Sec. 4.9 for details) shown in Table 4.1.

Robot parameters		i=1	i=2
Mass	m_i (kg)	1.390	0.527
Moment of inertia	I_i (kgm ²)	0.0297	0.0104
Link length	l_i (m)	0.46	0.46
COM location	l_{ci} (m)	0.362	0.233
Viscous friction	d_i (Nm/s)	0.03	0.035

MACCEPA parameters		value
Spring constant	κ_s (N/m)	771
Lever length	B (m)	0.03
Pin displacement	C (m)	0.125
Drum radius	r_d (m)	0.01

Table 4.1: Model parameters of the two-link brachiating robot and the variable stiffness actuator.

To formulate the optimisation, we use the state space representation in (4.8). At the transition at handhold, an affine discrete state transition $\mathbf{x}^+ = \Delta(\mathbf{x}^-) = \Gamma \mathbf{x}^- + \gamma$ is introduced to shift the coordinate system for the next handhold and reset the joint velocities of the robot to zero, which is defined as:

$$\Gamma = \text{diag}(\Gamma_1, \dots, \Gamma_4), \quad (4.38)$$

where:

$$\Gamma_1 = \begin{bmatrix} 1 & 1 \\ 0 & -1 \end{bmatrix}, \Gamma_2 = \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix}, \Gamma_3 = \Gamma_4 = \begin{bmatrix} -1 & 0 \\ 0 & 1 \end{bmatrix} \quad (4.39)$$

and $\gamma = [-\pi, 0, \dots, 0]^T$. Note that in this example, we have $\Delta = \Delta^{1,2} = \Delta^{2,1}$.

4.7 EXPLOITATION OF PASSIVE DYNAMICS WITH SPATIO-TEMPORAL OPTIMISATION OF STIFFNESS

In this section, we explore the benefits of simultaneous variable stiffness and temporal optimisation for tasks exploiting the intrinsic dynamics of the system. Brachiation is an example of a highly dynamic manoeuvre, requiring the use of passive dynamics for successful task execution. First, we discuss the implementation of a passive control strategy with variable stiffness actuation.

Our focus in this section is on how passive dynamics can be exploited in a single phase movement with the proposed OC framework. This is a particular instantiation of the OC formulation for the case of a single set of dynamics, without discrete state transition. Then, we present numerical simulations to demonstrate the benefits of temporal optimisation and time-varying stiffness modulation in a brachiation task. Later in Section 4.9, we report hardware experiments on a physical two-link brachiating robot with VSA.

4.7.1 *Passive Control Strategy in Swing Movement with Variable Stiffness Actuation*

A natural strategy would be to utilise gravity by making the joints passive and compliant. For example, while walking, humans seem to let the lower leg swing freely by relaxing the muscles controlling the knee joint during the swing phase and increase stiffness only when necessary. Stiffness modulation is also observed during a walking cycle in a cat locomotion study ⁵ [Akawaza et al., 1982]. This contrasts with the approach of high gain position controllers used for biped robots with stiff joints.

Consider the dynamics of a simplified single-link pendulum under the influence of gravity. If we consider an idealised VSA model of the form $\tau = -k(q - q_m)$,

⁵ To our knowledge, there are a large number of studies of stretch reflexes modulation in human walking, however, work specifically addressing stiffness modulation is very limited. For human arm cyclic movement, [Bennett et al., 1992] reported time-varying stiffness modulation in the elbow joint.

where q is the joint angle, τ is the joint torque, k is the stiffness and q_m is the equilibrium position of the elastic actuator, then the dynamics can be written as:

$$ml^2\ddot{q} + mgl \sin q = \tau = -k(q - q_m), \quad (4.40)$$

where m is the mass, l is the length and g is the gravitational constant. In this idealised VSA model, we assume that k and q_m are the control variables. From a viewpoint of position control, one way of looking at this system is as a manipulator with a flexible (elastic) joint, where we solve a tracking control problem [Spong, 1987]. In [Palli et al., 2008] a tracking control algorithm is proposed for such a flexible joint manipulator with variable stiffness actuation, in order to achieve asymptotic tracking of the desired joint and stiffness trajectories based on input-output linearisation. This constitutes an active cancellation of the intrinsic robot dynamics. Note that the main focus of [Palli et al., 2008] is the tracking control of the *given* joint and stiffness trajectories, and the problem of generating such desired trajectories for a given specific task is not addressed.

Alternatively, by rearranging the dynamics of (4.40) (for $\sin q \approx q$) as:

$$ml^2\ddot{q} + (mgl + k)q = v, \quad (4.41)$$

where $v = kq_m$, it can be observed that varying the stiffness of the actuator k in the second term of the left hand side effectively changes the dynamics property (e.g., the natural frequency of the pendulum). Thus, the control problem can be framed as finding an appropriate (preferably small) stiffness profile k to modulate the system dynamics (only when necessary) and compute the virtual equilibrium trajectory q_m [Shadmehr, 1990] to fulfil the specified task requirement, while maximally exploiting the natural dynamics.

In a realistic situation, obtaining the control solution to this effect is non-trivial, due to the complexity of the system dynamics, actuator mechanisms, the requirement of coordination of multiple degrees of freedom and redundancy in actuation. Next, we exploit the framework of optimal control and spatio-temporal optimisation of variable stiffness actuation to find appropriate control commands to implement the brachiation task.

4.7.2 Optimisation of a Single Phase Movement in Brachiation Task

In this section, we consider the task of swing locomotion from handhold to handhold on a ladder. Motivated by the discussions on passive control strategy in Sec. 4.7.1, we consider the following cost function to encode the task:

$$J = (\mathbf{y}(T) - \mathbf{y}^*)^T \mathbf{Q}_T (\mathbf{y}(T) - \mathbf{y}^*) + \int_0^T (\mathbf{u}^T \mathbf{R}_1 \mathbf{u} + R_2 F^2) dt, \quad (4.42)$$

where $\mathbf{y} = [\mathbf{r}, \dot{\mathbf{r}}]^T \in \mathbb{R}^4$ are the position and the velocity of the gripper in the Cartesian coordinates, \mathbf{y}^* contains the target values when reaching the target $\mathbf{y}^* = [\mathbf{r}^*, \mathbf{0}]^T$ and F is the spring tension in the VSA given in (4.36).

This objective function is designed in order to reach the target located at \mathbf{r}^* at the specified time T , while minimising the spring tension F in the VSA. The main component in the running cost is to minimise the spring tension F in the second term, while the first term $\mathbf{u}^T \mathbf{R}_1 \mathbf{u}$ is added for regularisation with a small choice of the weights in \mathbf{R}_1 . In practice, this is necessary since F is a function of the state and iLQG requires a control cost in its formulation to compute the optimal control law.

The actuator torque (4.37) can be expressed in the form:

$$\tau = -F \sin(q - q_{m1}) / \gamma', \quad (4.43)$$

where $\gamma' = \sqrt{B^2 + C^2 - 2BC \cos(q_{m1} - q)} / BC$. In (4.43), F can be interpreted as having a similar role to the stiffness parameter k in the simplified actuator model:

$$\tau = -k(q - q_m). \quad (4.44)$$

Another interpretation can be considered such that if we linearise (4.34) around the equilibrium position assuming that $\alpha = q_{m1} - q \ll 1$, the relationship between the joint stiffness k in (4.35) and the spring tension F in (4.36) can be approximated as:

$$k \approx \frac{1}{\sqrt{B^2 + C^2 - 2BC}} F. \quad (4.45)$$

Thus, effectively, minimising the spring tension F corresponds to minimising the stiffness k in an approximated way. It is possible to directly use k in the cost

function. However, in practice, first and second derivatives of k are needed to implement the iLQG algorithm, which become significantly more complex than those of F .

4.7.3 Benefit of Temporal Optimisation

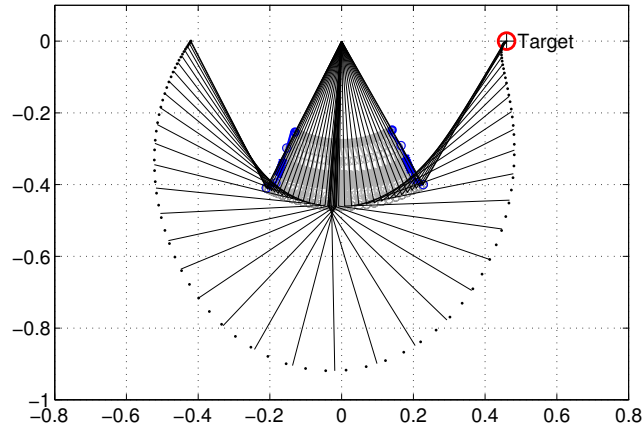
In this section, we numerically explore the benefit of temporal optimisation in exploiting the natural dynamics of the system. One of the disadvantages of the conventional optimal control formulation is that the time horizon needs to be specified in advance for a given task. In fully actuated systems, control can be used to enforce a pre-specified timing, while on underactuated ones, it is not always possible to choose an arbitrary time horizon. In a brachiation task, finding an appropriate movement horizon is essential for successful task execution with reduced control effort. This requires matching the movement duration corresponding to the property of the natural dynamics of the pendulum-like swing motion.

Consider the task of swing locomotion on a ladder with irregular intervals from $d_L = 0.42\text{m}$ to the target located at $d_R = 0.46\text{m}$. We optimise both the control command \mathbf{u} and the movement duration T . We use $\mathbf{Q}_T = \text{diag}\{10^4, 10^4, 10, 10\}$, $\mathbf{R}_1 = \text{diag}\{10^{-4}, 10^{-4}\}$ and $R_2 = 0.01$ for the cost function in (4.42)⁶. The optimised movement duration is $T = 0.806\text{s}$.

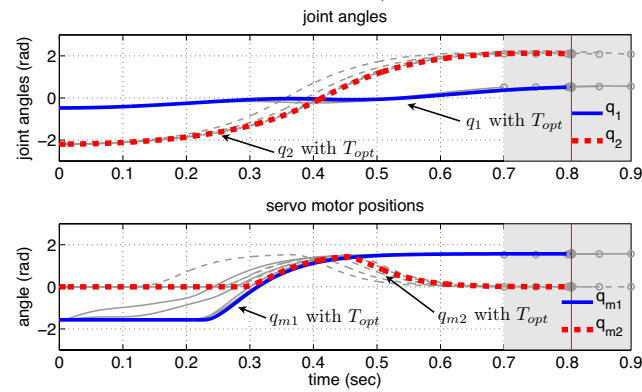
Fig. 4.4 displays the simulation result of (a) the optimised robot movement, (b) joint trajectories and servo motor positions, and (c) joint torque, spring tension and joint stiffness. Trajectories of the fixed time horizon ranging $T \in [0.7, 0.75, \dots, 0.9]\text{s}$ are overlaid for comparison alongside the case of the optimal movement duration $T = 0.806\text{s}$. In the movement with temporal optimisation, the spring tension and the joint stiffness are kept small at the beginning and end of the movement resulting in nearly zero joint torque, which allows the joint to swing passively. The joint torque is engaged only during the middle of the swing by increasing the spring tension as necessary.

By contrast, with non-optimal time horizon, larger joint torque and spring tension as well as higher joint stiffness can be observed, resulting in the requirement of larger control effort. This result suggests that the natural plant dynamics are fully exploited for the desirable task execution, based on the control strategy discussed in Sec. 4.7.1, with simultaneous stiffness and temporal optimisation.

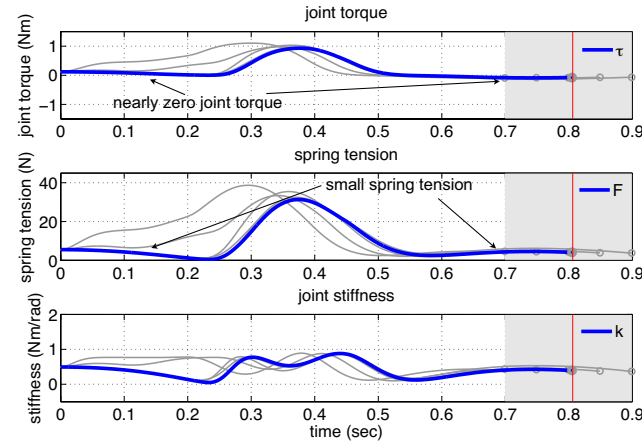
⁶ A brief investigation conducted by our collaborator, Dr. Nakanishi, into the possible options in defining a viable cost function has been included in Appendix C



(a) Movement of the robot (simulation) with optimal variable stiffness control (optimised duration $T = 0.806s$)



(b) Joint trajectories and servo motor positions



(c) Joint torque, spring tension and joint stiffness

Figure 4.4: Optimisation results of the swing locomotion task with temporal optimisation. In (b) and (c), grey thin lines show the plots for non-optimised T in the range of $T = [0.7, 0.75, \dots, 0.9]s$ and blue thick lines show the plots for optimised $T = 0.806s$. With temporal optimisation, at the beginning and the end of the movement, joint torque, spring tension and joint stiffness are kept small allowing the joint to swing passively in comparison to the non-optimal time cases.

In order to illustrate the effect of the servo motor dynamics characterised by (4.7), Fig. 4.5 shows the servo motor position commands and actual motor angles for the optimal movement duration, corresponding to the result in Fig. 4.4. Limitations in the bandwidth of the servo motor response can be observed in this plot. This suggests that the proposed optimal control framework can find appropriate control commands taking this effect into account.

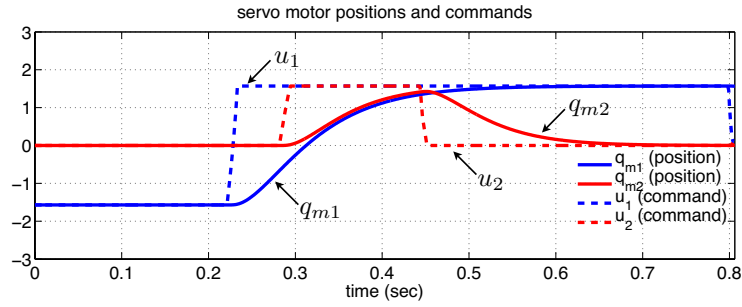


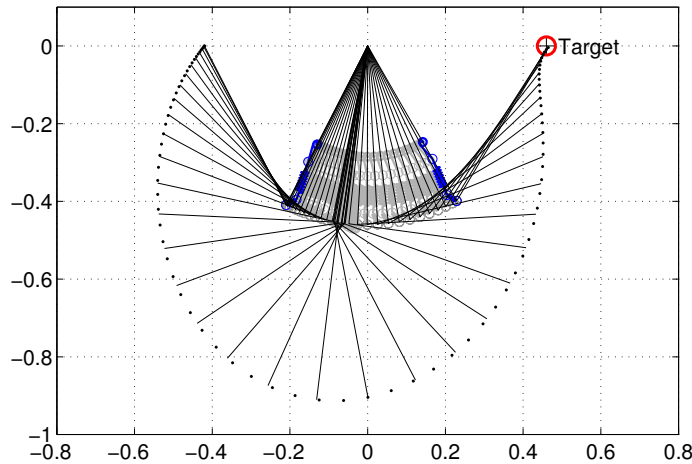
Figure 4.5: Servo motor commands u_i (dotted line) and actual angle q_{mi} (solid line) for the results with optimal movement duration $T = 0.806$ (Fig. 4.4 (b) bottom). Servo motor response delay can be observed characterised by the servo motor dynamics (4.7). The proposed optimal control framework finds appropriate control commands taking this effect into account.

4.7.4 Benefit of Stiffness Variation

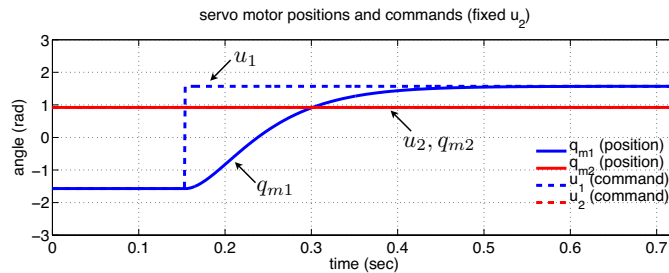
In this section, we investigate the benefit of time-varying stiffness modulation. One of the characteristics of VSAs is their ability to simultaneously modulate joint torque and stiffness. As discussed in Sec. 4.7.1, modulating stiffness effectively alters the properties of the system dynamics, such as natural frequency. Thus, the capability of modulating stiffness during the motion is potentially beneficial for improving the task performance.

We demonstrate the benefit of time-varying modulation of stiffness by comparing optimal *variable* stiffness control and optimal *fixed* stiffness control. In optimal *variable* stiffness control, both control commands in $\mathbf{u} = [u_1(t), u_2(t)]^T$ are optimised in a time-varying manner during the movement to independently control the joint torque and the joint stiffness. In optimal *constant* stiffness control, the command to the spring pre-tensioning servomotor is fixed to the optimal constant value throughout the movement (i.e., $\mathbf{u} = [u_1(t), u_2]^T$ where $u_2 = \text{const.}$). We note that a constant command to the pre-tensioning servomotor u_2 does not necessarily mean constant joint stiffness with MACCEPA [Braun et al., 2012a].

Fig. 4.6 shows the simulation result of a swing locomotion for the intervals $d_L = 0.42\text{m}$ and $d_R = 0.46\text{m}$ (the same configuration as in Sec. 4.7.3) with the optimal fixed stiffness control. Fig. 4.6 (a) illustrates the optimised movement of the robot and (b) shows the servomotor commands and positions, where u_2 and q_{m2} are kept at their optimal constant value during the movement. The optimal time horizon for this movement was $T = 0.723\text{s}$. The resultant movements in Fig. 4.4 (a) (optimal variable stiffness) and Fig. 4.6 (a) (optimal constant stiffness) look comparable. However, optimal constant stiffness control results in a higher cost $J = 9.528$ than that of the corresponding optimal variable stiffness case, which is $J = 2.979$.



(a) Movement of the robot (simulation) with optimal fixed stiffness control (optimised movement duration $T = 0.723\text{s}$)



(b) Servo motor commands u_i (dotted line) and actual angle q_{mi} (solid line) for the results with optimal fixed stiffness control. Note that the pre-tension motor the command u_2 and its actual position q_{m2} are kept to the optimal constant value throughout the movement.

Figure 4.6: Optimisation results of the swing locomotion task with optimal fixed stiffness control. The intervals between the bars correspond to the case with optimal variable stiffness control in Sec. 4.7.3.

A Stiffness Modulation and Range of Achievable Targets

Next, we compare the performance of variable and constant optimal stiffness control in terms of the range of distances that can be reached with the robot. Starting

from the bar at the nominal distance $d_L = 0.42\text{m}$, we vary the target positions d_R by 1cm and optimise the control commands and movement duration. When the end effector position at $t = T_f$ is within a tolerance of 1cm from the location of the target, we assume that the trial is successful.

With optimal variable stiffness control, the robot was able to reach the target in the range of $d_R \in [0.39, 0.59]\text{m}$ (the range of 20cm) while with optimal constant stiffness control, it was $d_R \in [0.42, 0.52]\text{m}$ (the range of 10cm). These numerical explorations illustrate the benefit of optimal *variable* stiffness control in terms of the cost and range of distances achieved in swing locomotion in comparison to optimal *constant* stiffness control.

4.8 SPATIO-TEMPORAL OPTIMISATION OF MULTIPLE SWINGS IN ROBOT BRACHIATION

To fully demonstrate the performance of our developed approach, we consider a brachiating task with multiple phases of the movement as follows: First, the robot swings up from the suspended posture to the target at $d_1 = 0.40\text{ m}$ and subsequently moves to the target located at $d_2 = 0.42$ and $d_3 = 0.46\text{ m}$, respectively. The composite cost function to encode this task is given by:

$$J = (\mathbf{y}(T_f) - \mathbf{y}_f^*)^T \mathbf{Q}_{T_f} (\mathbf{y}(T_f) - \mathbf{y}_f^*) + \sum_{j=1}^K (\mathbf{y}(T_j^-) - \mathbf{y}_j^*)^T \mathbf{Q}_{T_j} (\mathbf{y}(T_j^-) - \mathbf{y}_j^*) + \int_0^{T_f} (\mathbf{u}^T \mathbf{R}_1 \mathbf{u} + R_2 F^2) dt + w_T T_1, \quad (4.46)$$

where $K = 3$ is the number of phases, $\mathbf{y} = [\mathbf{r}, \dot{\mathbf{r}}]^T \in \mathbb{R}^4$ are the position and the velocity of the gripper in the Cartesian coordinates measured from the origin at current handhold, \mathbf{y}^* contains the target values when reaching the target $\mathbf{y}^* = [\mathbf{r}^*, \mathbf{0}]^T$ and F is the spring tension in the VSA. This cost function includes the time cost $w_T T_1$ for the swing up manoeuvre. We use $\mathbf{Q}_T = \mathbf{Q}_{T_j} = \text{diag}\{10^4, 10^4, 10, 10\}$, $\mathbf{R}_1 = \text{diag}\{10^{-4}, 10^{-4}\}$ and $R_2 = 0.01$ and $w_T = 1$. In addition, we impose constraints on the range of the angle of the second joint during the course of the swing up manoeuvre as $q_{2_{\min}} \leq q_2 \leq q_{2_{\max}}$, where $[q_{2_{\min}}, q_{2_{\max}}] = [-100, 100]\text{deg}$, by adding a penalty term to the cost (4.46). This is introduced considering the physical joint limit of the hardware platform used in this study.

Fig. 4.7 (a) shows the sequence of the multi-phase movement of the robot optimised using the proposed algorithm including temporal optimisation. The optimised switching instance and total movement duration are $T_1 = 5.259s$, $T_2 = 6.033s$ and $T_f = 6.835s$ and the total cost is $J = 37.815$. Fig. 4.7 (b) shows the optimised joint trajectories and servo motor positions. Note that at the instance of switching, denoted by vertical lines, discrete state transition can be observed in these trajectories due to the definition of the coordinates transformation.

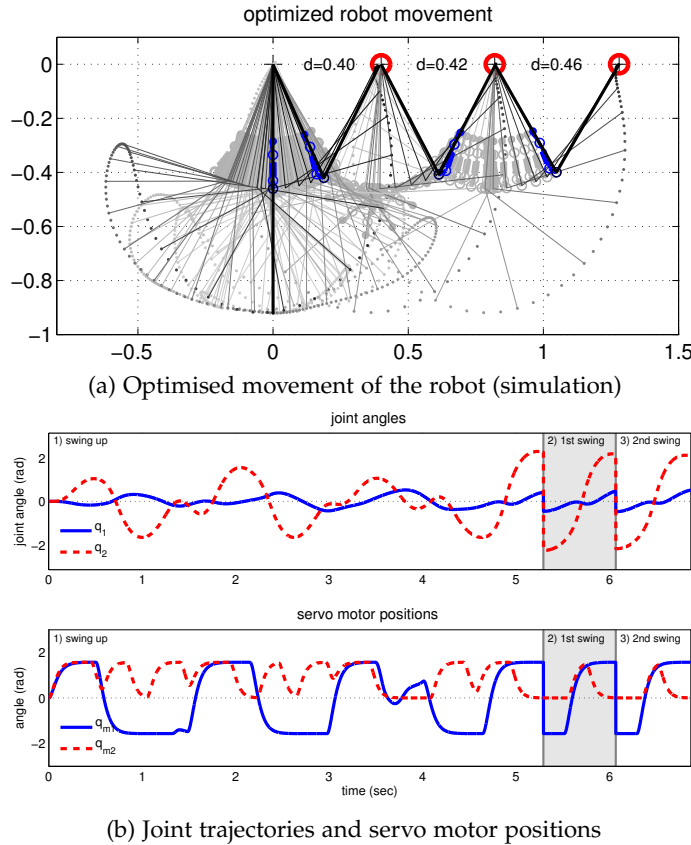


Figure 4.7: (a) Optimised multi-phase movement of the robot. (b) Joint trajectories and servo motor positions of complete optimisation with temporal optimisation.

To illustrate the benefit of the proposed multi-phase formulation, we performed a movement optimisation with a pre-specified nominal (fixed, non-optimal) time horizon with $T_1 = 5.2s$, $T_2 = 5.9s$ and $T_f = 6.7s$ using the same cost parameters in sequential and multi-phase optimisation. With sequential individual movement optimisation, at the end of the final phase movement, a large overshoot was observed (the distance from the target at $t = T_f$ was 0.0697 m) incurring a significantly large total cost of $J = 101.053$. By contrast, with multi-phase movement optimisation, for the same pre-specified time horizon, it was possible to find a

feasible solution within a tolerance of 1cm at the end of each phase at the bar. In this case, the total cost was $J = 50.228$. These results demonstrate the benefit of the multi-phase movement optimisation in finding optimal control commands and temporal aspect of the movement using the proposed method resulting in lower cost.

In the next section we briefly describe the developed hardware platform used to showcase the applicability of the proposed framework under realistic physical conditions. We then present the experimental results obtained on brachiation tasks for the presented two-link robotic system with VSA.

4.9 HARDWARE DEVELOPMENT

This section describes the hardware platform used in the experiments. Fig. 4.8 depicts the configuration of the brachiating robot with VSA. The elbow joint is fitted with a MACCEPA actuator equipped two servo motors (Hitec HS-7940TH).

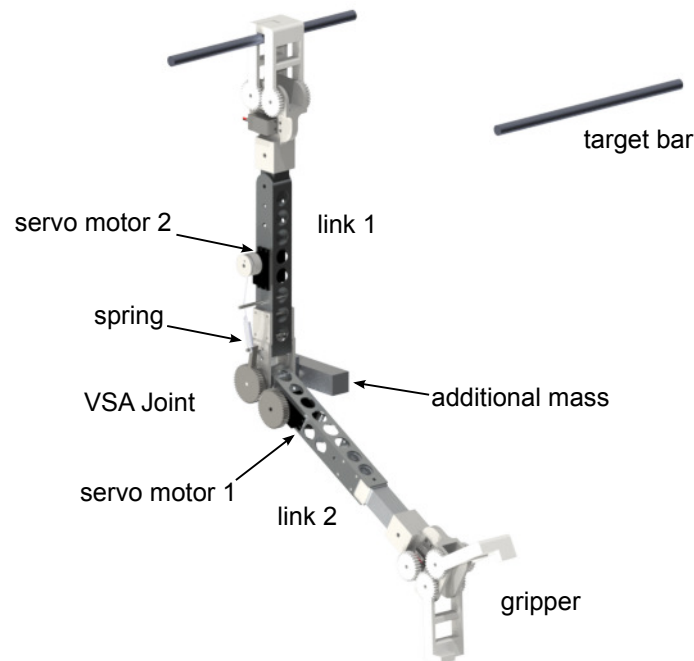


Figure 4.8: Configuration of the two-link brachiating robot with VSA in CAD drawing.

Each link is fitted with a gripper driven by a single servo motor (Hitec HSR-5990TG) which opens and closes it through a gear mechanism. The angle of the first link is measured by integrating its angular velocity obtained through an IMU unit (InvenSense MPU-6050) attached to the link. The angle of the second link is

measured by a rotary potentiometer (Alps RDC503013A) at the elbow joint. The servo motor positions are measured through direct access to its internal potentiometer. The operating frequency of control and measurement is 1 kHz.

The length of each link is 0.46 m and the total mass is about 1.92 kg. The link parameters are computed from the CAD model, while friction coefficients and the servo motor bandwidth parameters are estimated by fitting the actual responses of the robot.

Our numerical exploration showed that with an inadequate mass distribution, it was difficult to find an optimal solution in achieving desired swing locomotion behaviour under the constraint. Thus, the mass distribution of the robot was empirically adjusted to resemble the desirable natural dynamics required for the task (see Table 4.1 for the parameters). An overview of the hardware design evolution is provided in Appendix D.

The parameters in Table 4.1 were obtained from the CAD model of the system. We performed data collection from the hardware system, under various command input sets and compared the observed behaviour with those anticipated from the simulation. We used a similar approach to tune the value of the bandwidth parameters in the motor dynamics model, α . An overview of the details regarding these validation procedures is provided in Appendix E.

4.10 EVALUATION ON HARDWARE PLATFORM

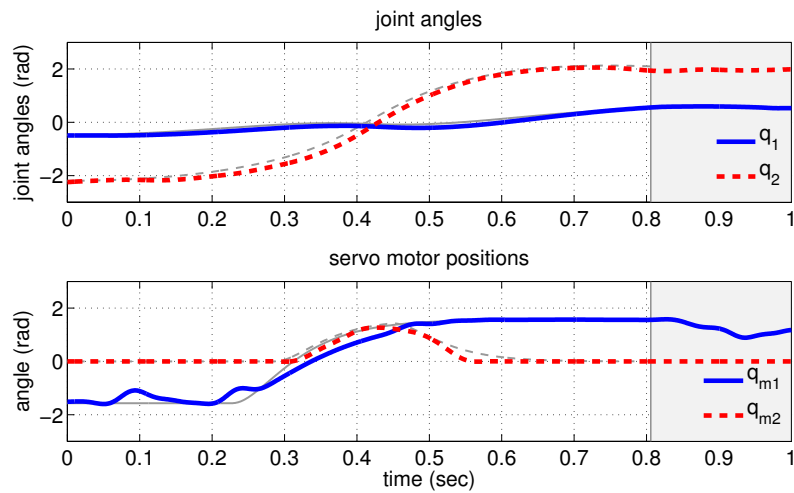
This section presents experimental results implemented on the two-link brachiating variable stiffness robot hardware corresponding to the numerical simulations reported in Sec. 4.7.3 and Sec. 4.8.

4.10.1 *Spatio-Temporal Optimisation for a Single-Phase Movement*

Fig. 4.9 shows the experimental result of a swing locomotion on the ladder with the intervals of $d_l = 0.42$ m and $d_r = 0.46$ m. The optimised control commands with the optimal movement duration obtained in the corresponding simulation in Section 4.7.3 are used. In Fig. 4.9 (a), the movement of the robot is depicted, while in Fig. 4.9 (b) the joint trajectories and servo motor positions are shown. This result corresponds to the simulation in Fig. 4.4 for the optimal movement duration. In the experiments, we only use the open-loop optimal control command to the servo motors without state feedback as in [Braun et al., 2012a].



(a) Optimised movement of the robot (single phase locomotion)



(b) Joint trajectories and servo motor positions

Figure 4.9: Experimental result of the swing locomotion task on the brachiation robot hardware. In (b), red and blue thick lines show the experimental data, and grey thin lines show the corresponding simulation results with the optimised planned movement duration $T = 0.806$ s presented in Sec. 4.7.3.

4.10.2 Multi-Phase Movement with Switching Dynamics and Discrete State Transitions

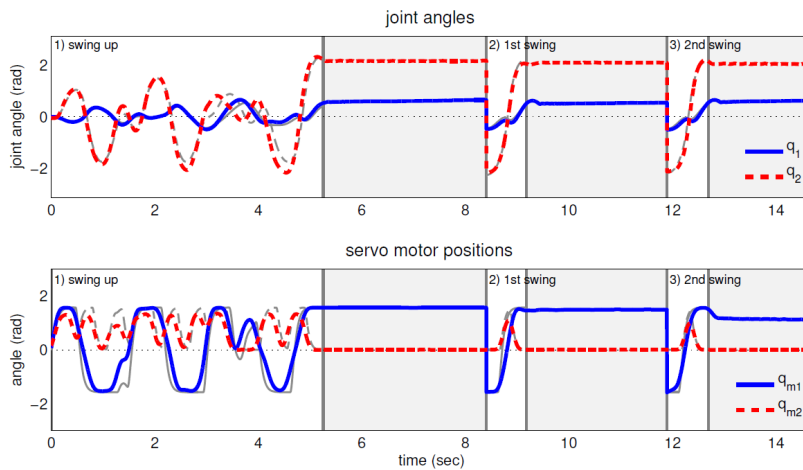
Fig. 4.10 shows the experimental result of multi-phase movements consisting of swing-up followed by two additional swings which corresponds to the simulation in Fig. 4.7 (Sec. 4.8). The intervals between the target bars are $d_1 = 0.40$, $d_2 = 0.42$ and $d_3 = 0.46$ m. Fig. 4.10 (a) illustrates the movement of the robot moving from left to right and Fig. 4.10 (b) shows the joint trajectories and servo motor positions.

The joint trajectories in the experiment closely match the planned movement in the simulation. The discrepancy observed during parts of the movement can be mainly attributed to mismatch in the model of the hardware. Note that at the end

of each phase of the movement, switching to the next phase is manually done by confirming firm grasping of the bar in order to avoid falling off from the ladder at run-time ⁷.



(a) Movement of the robot (experiment)



(b) Joint trajectories and servo motor positions

Figure 4.10: (a) Experimental results of multi-phase locomotion task with the brachiating robot hardware. In (b), red and blue lines show the actual hardware behaviour, and grey lines show the corresponding simulation results presented in Sections 4.8.

These experimental results demonstrate the effectiveness and feasibility of the proposed framework in achieving highly dynamic tasks in compliantly actuated robots with variable stiffness capabilities under real conditions. (The results presented are depicted in the additional video material).

⁷ The minor model discrepancies observed in the joint trajectories are suppressed, at the end of each phase, by the grippers' ability to reach and stabilise at the target position on the bar.

4.11 DISCUSSION AND CONCLUSIONS

In this chapter, we presented the application of our systematic methodology for movement optimisation with multiple phases and switching dynamics. We address this in the context of robotic systems with VIA, with the focus on exploiting the intrinsic dynamics of the system.

Tasks including switching dynamics and interaction with the environment are approximately modelled as a hybrid dynamical system with time-based switching. We showed the benefit of simultaneous temporal and variable stiffness optimisation leading to a lowered control effort and improved performance. We demonstrated the effectiveness of the proposed approach in example tasks in both numerical simulations and in a hardware implementation of a brachiating robot with VSA.

In Sec. 4.7.3, we highlight the benefit of temporal optimisation for exploiting the natural dynamics of the system. We use for this an example of a swing locomotion task on a brachiation system with VSA capabilities. The selection of an optimal time is crucial for lowering the required control effort. The results presented are closely related to the work in [Nakanishi et al., 2011], which showed how, for a periodic motion of a brachiation system, finding an appropriate frequency matching the natural one of the system, can reduce the control effort.

We showed the benefit of multi-phase movement optimisation on the brachiation system in terms of cost and performance, compared to individual phase optimisation. This is due to the multi-phase optimisation taking into consideration future goals. This result might be less evident in the brachiation example used, as the dynamics of the system define a reset in the joint velocities every time the robot grasps the bar. However, the effect can be observed in the positions of the VSA servomotors, which are appropriately adjusted at the end of each phase, considering the next phase of the movement.

A significant practical consideration for time-based switching approaches is ensuring the feasibility and approximation of the switching condition. This is generally determined by the given task and the associated cost function, as the switching condition is imposed as a soft constraint by the via-point and terminal costs. In the case of our brachiation system, given the design of the gripper in the physical robot, a small error at the endpoint was tolerated considering that the robot was able to grasp the bar.

The current methodology assumes the existence of a predefined sequence of switching, where the switching conditions can be incorporated in the cost function. If such a sequence is not available, a possible alternative approach can be provided by one of the methods discussed in Sec. 4.2.3. Performing a temporal optimisation in these scenarios would be more difficult as the switching time depends on the trajectory of the system [Xu and Antsaklis, 2004].

In the next chapter we extend the evaluation of our proposed approach to a monopod hopper. This constitutes a significantly more challenging platform in terms of contacts (i.e., more complex impact map) and individual phase dynamics (i.e., existence of a flight phase). We present numerical evaluations showcasing the benefits of multi-phase optimisation on this increasingly complicated system.

HOPPING ROBOT WITH IMPEDANCE MODULATION

In this chapter we further demonstrate the effectiveness of the proposed approach, introduced in Chapter 4. First, we provide a brief overview of the related work regarding the control of the platform used in this chapter, the monopod hopper. We then apply our methodology and present numerical evaluations on the challenging task of hopping locomotion.

The results presented show the feasibility of the developed framework, highlighting the benefits of multi-phase optimisation and the robustness of the obtained optimal feedback controller.

5.1 PREVIOUS WORK ON CONTROL FOR HOPPER SYSTEM

The monopod hopper is a well known platform, previously used in a variety of studies on dynamics and control. A locomotion task for such a system includes switching of different modes of dynamics (flight and stance) and complex discontinuous state transition caused by the touch down impact.

During the stance phase the system is underactuated, while in the flight phase the absence of contact with the ground does not allow control the system using the contact forces. An additional difficulty in this task is the temporal optimisation aspect. Finding the flight and stance times for successful task execution is highly restricted by the intrinsic dynamics and the desired task specifications.

Thompson and Raibert explored passive dynamic locomotion on a one legged hopper with hip compliance [Thompson and Raibert, 1990] and show that the system can achieve a wide range of locomotion speeds. Many approaches were based on Raibert's control [Raibert and Tello, 1986]. The most engaging results revolved around the work of Martin Buehler et al. [Ahmadi and Buehler, 1997; Gregorio et al., 1997; Sato and Buehler, 2004]. In their first attempts they adapted the Raibert controller to their particular hopper design (the electrically actuated ARL Monopod) [Gregorio et al., 1997]. In further studies they progressed by considering compliant elements in the hip and leg [Ahmadi and Buehler, 1997]. A similar controller that exploits the passive dynamics, achieving stabilisation of a param-

eterised passive dynamic motion was proposed. In later work [Sato and Buehler, 2004] the Spring Loaded Inverted Pendulum (SLIP) model was considered, due to being naturally stable, and thus the control problem was simplified.

Alternatives to the Raibert-based controllers were developed. In [François and Samson, 1998], natural regimes were regarded as periodic orbits which could be mapped and stabilised, an approach borrowed from [McGeer, 1990]. The method involved a discrete-time controller, that applied impulsive forces computed at each time step. Based on this method, [Hyon and Emura, 2002] devised a controller that applied piecewise-constant forces. It used the intrinsic dynamics of the hybrid system, instead of the pre-planned periodic solutions.

5.2 EXPERIMENTS: HOPPING ROBOT WITH VIA

In our work we augment the hopping model from [Hyon and Emura, 2004] with variable stiffness capabilities in the hip and leg actuators. We demonstrate the feasibility of the proposed approach on this increasingly challenging platform. Given the difficulties raised by the dynamics of the hopper system, a successful task execution will rely heavily on a proper exploitation of the variable impedance elements and the passive dynamics.

5.2.1 Dynamics Model of a Hopping Robot

The hopping dynamics switch between the flight phase ($i = 1$) and stance phase ($i = 2$) at the switching instance T_k when either touch-down or lift-off conditions are met. The configuration vector of the system is defined as $\mathbf{q} = [x_{\text{com}}, y_{\text{com}}, \theta, \phi, r]^T$ (see Fig. 5.1).

The flight dynamics ($i = 1$) are given as :

$$\begin{cases} M\ddot{x}_{\text{com}} = 0, & M\ddot{y}_{\text{com}} = -Mg, \\ J_l\ddot{\theta} = -\tau_{\text{hip}}, & J_b\ddot{\phi} = \tau_{\text{hip}}, \end{cases} \quad (5.1)$$

with $r = r_0$. The stance dynamics ($i = 2$) are described as:

$$\begin{cases} (J_l + Mr^2)\ddot{\theta} + 2Mr\dot{\theta} - Mrg \sin \theta = -\tau_{\text{hip}}, \\ J_b\ddot{\theta} = \tau_{\text{hip}}, \\ M\ddot{r} - Mr\dot{\theta}^2 + Mg \cos \theta = \tau_{\text{leg}}, \end{cases} \quad (5.2)$$

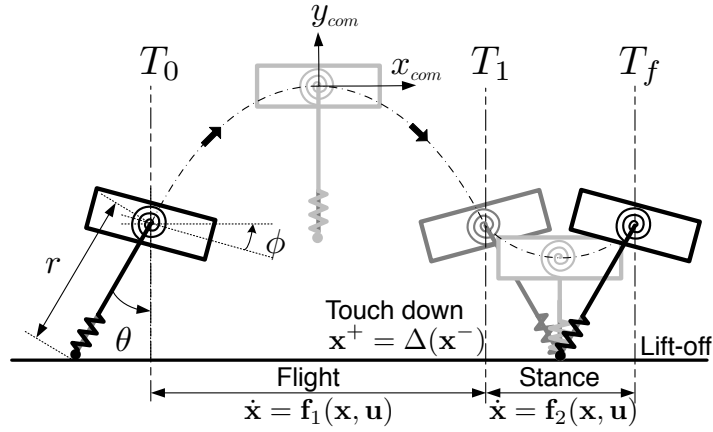


Figure 5.1: Hybrid dynamics of a hopping robot model and locomotion phases during one cycle

with $x = -r \sin \theta + x_{\text{foot}}$ and $y = r \cos \theta + y_{\text{foot}}$. M is the mass of the body, J_l is the leg inertia, J_b is the body inertia, g is the gravitational constant, r_0 is the nominal length of the leg spring. τ_{hip} and τ_{leg} are the torque applied to the hip joint and the force applied to the leg by the VSAs as given in (5.4) and (5.5) below, respectively. We use the parameters $M = 11.0$ kg, $J_b = 2.5$ kgm², $J_l = 0.25$ kgm² and $r_0 = 0.7$ m adopted from [Ahmadi and Buehler, 1997].

The general equation of the system dynamics (4.6) can be formulated in this case as:

$$\mathbf{M}_i \ddot{\mathbf{q}}_i + \mathbf{C}_i(\mathbf{q}_i, \dot{\mathbf{q}}_i) \dot{\mathbf{q}}_i + \mathbf{g}_i(\mathbf{q}_i) = \boldsymbol{\tau}_i(\mathbf{q}_i, \mathbf{u}_i), \quad (5.3)$$

where $\mathbf{q}_1 = [x_{\text{com}}, y_{\text{com}}, \theta, \phi]^T$ and $\mathbf{q}_2 = [\theta, \phi, r]^T$ are the partial state vector for the flight and stance phase, respectively.

$\boldsymbol{\tau}_1(\mathbf{q}, \mathbf{u}) = [0, 0, -\tau_{\text{hip}}, \tau_{\text{hip}}]^T$ and $\boldsymbol{\tau}_2(\mathbf{q}, \mathbf{u}) = [-\tau_{\text{hip}}, \tau_{\text{hip}}, \tau_{\text{leg}}]^T$ are the VSA torque/force applied to each degree of freedom where :

$$\tau_{\text{hip}}(\mathbf{q}, \mathbf{u}) = u_1 - u_3(\phi - \theta), \quad (5.4)$$

$$\tau_{\text{leg}}(\mathbf{q}, \mathbf{u}) = u_2 - u_4(r - r_0), \quad (5.5)$$

and $\mathbf{u} = [u_1, u_2, u_3, u_4]^T$ is the control command vector defined as in Table 5.1.

The range of the position and stiffness control is limited as $\mathbf{u}_{\min} \preceq \mathbf{u} \preceq \mathbf{u}_{\max}$ where:

$$\begin{aligned}\mathbf{u}_{\min} &= [-100, -100, 10, 2000]^T, \\ \mathbf{u}_{\max} &= [100, 100, 150, 25000]^T.\end{aligned}\quad (5.6)$$

For the purpose of optimisation, the dynamics are formulated in a state space representation of the form of $\dot{\mathbf{x}} = \mathbf{f}_i(\mathbf{x}, \mathbf{u})$ as in (3.8) with the full state vector $\mathbf{x} = [\mathbf{q}, \dot{\mathbf{q}}]^T$ and $\mathbf{q} = [x_{\text{com}}, y_{\text{com}}, \theta, \phi, r]^T$. Note that in this hopping robot, we consider a simplified parallel elastic VSA model with direct force/torque and stiffness control as in [Hyon and Emura, 2004], which does not include the motor dynamics¹ in (4.7).

Command	Meaning	Unit
u_1	torque applied at the hip	Nm
u_2	hip joint stiffness	N/m
u_3	force applied in the leg	N
u_4	leg stiffness	Nm/rad

Table 5.1: Commands for the hopping robot model

The discontinuous impact map at touch-down is defined as:

$$\mathbf{x}^+ = \Delta(\mathbf{x}^-) = \begin{bmatrix} \mathbf{I} & \mathbf{0} \\ \mathbf{0} & \Gamma(\mathbf{q}) \end{bmatrix} \mathbf{x}^-, \quad (5.7)$$

where $\mathbf{x} = [\mathbf{q}, \dot{\mathbf{q}}]^T$, $\mathbf{q}^+ = \mathbf{q}^-$ and $\dot{\mathbf{q}}^+ = \Gamma(\mathbf{q})\dot{\mathbf{q}}^-$. The matrix $\Gamma(\mathbf{q})$ is the transition map from the pre-impact to post-impact velocities based on a rigid body collision model [Grizzle et al., 2001; Rosa Jr. et al., 2012]:

$$\Gamma(\mathbf{q}) = \mathbf{I} - \mathbf{M}_1^{-1} \mathbf{J}^T (\mathbf{J} \mathbf{M}_1^{-1} \mathbf{J}^T)^{-1} \mathbf{J}, \quad (5.8)$$

with \mathbf{J} the Jacobean satisfying the contact constraint $\mathbf{J}(\mathbf{q}^-)\dot{\mathbf{q}}^+ = \mathbf{0}$. This form of the velocity transition map is given in the study of passive running with an additional analysis of energy dissipation at impact [Hyon and Emura, 2004]. At lift-off, the only discontinuity occurs in the velocity of the leg which is reset to zero as $\dot{r} = 0$ at $r = r_0$.

¹ This form of VSA in (5.4) and (5.5) can be equivalently written in the form of (4.7) with a change of coordinates (see discussions in Section II-D in [Nakanishi et al., 2011]) assuming that motors have ideal responses, (i.e., infinitely high bandwidth $\alpha \rightarrow \infty$).

5.2.2 Design of the Composite Cost Function

A stable locomotion of the system for an indefinitely long time consists of a repeated periodic hopping behaviour, with the period of one hopping cycle. Thus, an optimised hopping behaviour can be reduced to a periodic one cycle trajectory.

In our experiment we consider directly obtaining this behaviour, while exploiting the natural dynamics and the benefits of modulating stiffness. For this purpose, first, we design a composite cost function for one hopping cycle, including both the flight and stance phases and the touch-down condition. Then, the obtained controller is applied to achieve multiple cycles.

We consider the following cost function:

$$J = (\mathbf{x}(T_f) - \mathbf{x}_0)^T \mathbf{Q}_{T_f} (\mathbf{x}(T_f) - \mathbf{x}_0) + \Psi(\mathbf{x}(T_1^-)) + \int_0^{T_f} \mathbf{u}^T \mathbf{R} \mathbf{u} dt, \quad (5.9)$$

where \mathbf{Q}_{T_f} is a positive semi-definite matrix and \mathbf{R} is a positive definite matrix. The purpose of the first term is to achieve periodicity of the trajectory with \mathbf{x}_0 the initial state. The second term consists of two criteria:

$$\Psi(\mathbf{x}(T_1^-)) = Q_{T_1,1} (y_{\text{foot}}^- - 0)^2 + Q_{T_1,2} (\mu^-)^2, \quad (5.10)$$

where $Q_{T_1,1}$ and $Q_{T_1,2}$ are positive weights. The term $Q_{T_1,1} (y_{\text{foot}}^- - 0)^2$ penalises the height of the foot from the ground $y = 0$ to approximately encode the touch-down condition in a time-based formulation. Note that in order to minimise this term, it is important to find an appropriate flight time T_1 , as the trajectory of the centre of mass cannot be directly controlled in the flight phase for a given initial lift-off condition. The term $Q_{T_1,2} (\mu^-)^2$ minimises energy loss at touch-down where μ^- is the *energy dissipation coefficient* [Hyon and Emura, 2004]:

$$\mu^- = \dot{x}_{\text{com}}^- \cos \theta^- + \dot{y}_{\text{com}}^- \sin \theta^- + r_0 \dot{\theta}^- = \dot{x}_{\text{foot}}^- \cos \theta^- + \dot{y}_{\text{foot}}^- \sin \theta^-. \quad (5.11)$$

This encoding is motivated from the study of passive running [Hyon and Emura, 2004] in order to find an appropriate leg angle and foot velocity relative to the ground at touch-down. Note that for $\mu^- = 0$, there is no energy loss at impact. In the running cost, we minimise the control effort u_1 and u_2 , while small weights are used for penalising the magnitude of stiffness u_3 and u_4 for regularisation. The range of the control and stiffness is limited as $u_{i,\text{min}} \leq u_i \leq u_{i,\text{max}}$ (5.6).

The initial conditions were obtained by choosing the desired initial horizontal velocity and lift-off angle of the leg (\dot{x}_0, θ_0) and computing the remaining parameters under the assumption of synchronisation between the oscillatory movements of the leg swing and compression in the passive running model [Ahmadi and Buehler, 1997; Hyon and Emura, 2004].

5.2.3 Simulation Results

We choose the desired initial condition at lift off as $\dot{x}_0 = 2.0\text{m}$ and $\theta_0 = -6.0\text{deg}$. Note that with this initial condition, the passive dynamics (i.e., no control is applied) of the robot can achieve several steps of running as reported in [Ahmadi and Buehler, 1997; Hyon and Emura, 2004], however, it will eventually fail since passive running is intrinsically unstable.

Using the proposed method, we simultaneously obtained the optimal feedback control for the control commands (u_1, u_2) and stiffness (u_3, u_4), and found the flight time T_1 and the duration of the complete cycle T_f . The optimised flight time and one hopping cycle are $T_1 = 0.410\text{s}$ and $T_f = 0.490\text{s}$. The total cost is $J_{\text{comp}} = 1.624$. The obtained controller is rooted in a time-based switching assumption for one cycle. There could be some mismatch in the exact timing in the switching condition when applied to realistic event-based switching dynamics for multiple cycles of locomotion. One of the benefits of our approach is that it provides a locally optimal *feedback* control law, hence deviations from the optimal trajectory can be corrected. The results presented are featured in the additional video material.

A Comparison to Individual Phase Optimisation

For comparison, we optimise the control command, stiffness and the movement duration in a sequential manner individually for the flight phase, followed by the stance phase, for one cycle of the movement. The optimised movement duration for the flight phase is $T_{1,\text{ind}} = 0.410\text{s}$ and for the stance is $T_{\text{stance,ind}} = 0.080\text{s}$, resulting in the total duration $T_{f,\text{ind}} = 0.490\text{s}$. The total cost for this individual optimisation is $J_{\text{ind}} = 1.686$, which is comparable to the complete optimisation case $J_{\text{comp}} = 1.624$ mentioned above. The optimised trajectories, control commands and stiffness profiles are similar between these two cases. However, interestingly, we found a notable difference in the robustness of the controllers when applied

to the event-based switching dynamics, where the role of the *feedback* control becomes prominent.

The controller with complete cycle optimisation is able to achieve continuous stable running over multiple cycles. By contrast, with the controller obtained by individual optimisation, the robot fails after 25 steps of hopping. Although this is an empirical observation, this difference could be justified by the difference in the optimal *feedback* gains. In the complete cycle optimisation, these gains take the future goal until the end of the hopping cycle into account, including both the flight and stance phases with the via-point and terminal costs. However, in the individual optimisation, corrections are made only considering the immediate goal specified by the terminal cost in each phase. This result highlights the benefits of optimising the whole cycle of the movement, in comparison to individually optimising the movement in a sequential manner.

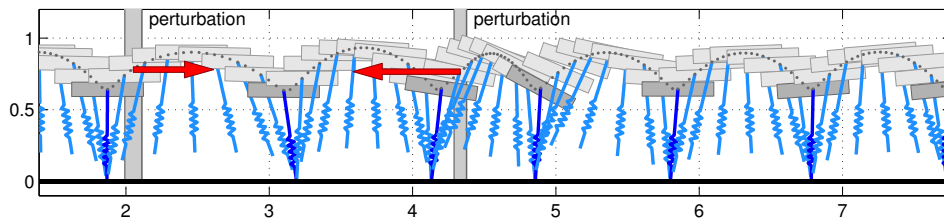
B Robustness against Perturbations

In the following simulation, we evaluate the robustness of the obtained optimal feedback controller by applying external perturbations while running. At $t = 1.0\text{s}$, the robot is pushed forward with $F_x = 150\text{N}$ for 0.05s , and at $t = 2.0\text{s}$, a backward perturbation is applied $F_x = -250\text{ (N)}$ for the same duration of 0.05s .

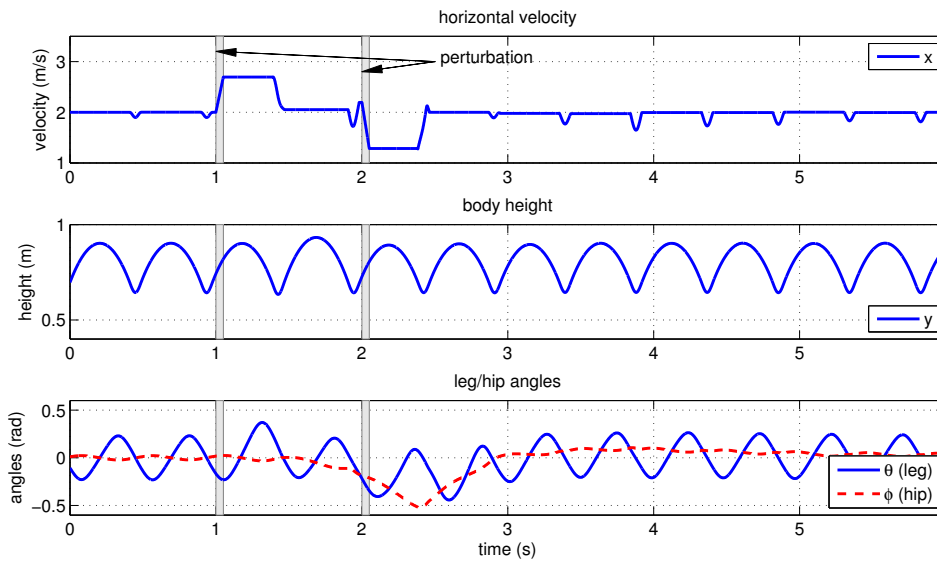
Figure 5.2 (a) depicts the movement of the robot from $t = 0.7 \sim 3.9\text{s}$. Figure 5.2 (b) shows the forward velocity \dot{x} (top), body height y_{com} (middle), and leg angle θ and hip angle ϕ (bottom) from $t = 0$ to $t = 6\text{s}$. Figure 5.2 (c) shows the control commands u_1 and u_2 (top), hip stiffness u_3 (centre) and leg stiffness u_4 (bottom). The simulation results illustrate that, after the perturbations were applied, the robot was able to stabilise the periodic running behaviour without falling over. This demonstrates the robustness of the optimal feedback controller and the feasibility of the proposed approach in this setting.

5.3 DISCUSSION AND CONCLUSIONS

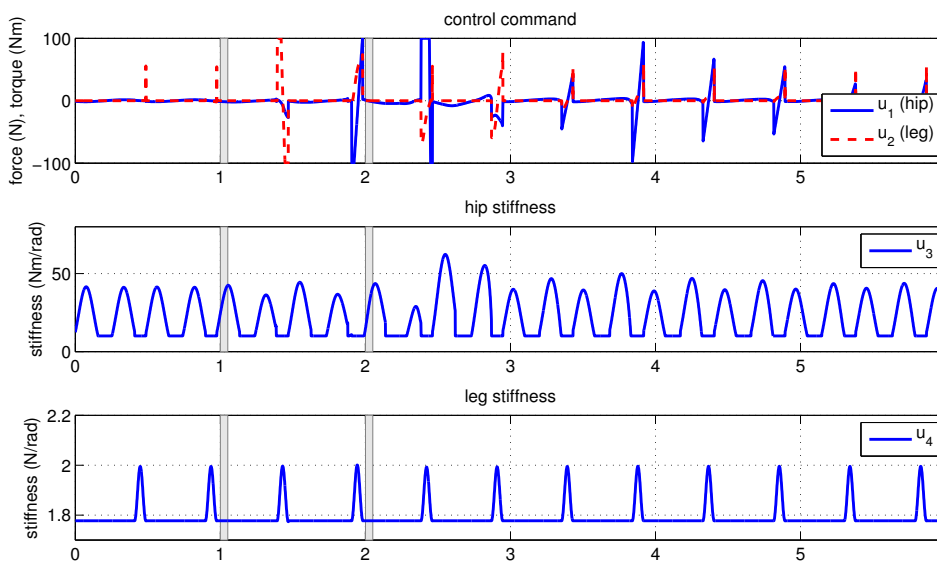
In this chapter we presented further applications of our systematic methodology for movement optimisation on the increasingly difficult task of hopping. With an appropriate choice of the composite cost function to encode the task, we demonstrated the effectiveness of the developed methodology in numerical simulations. The benefits of multi-phase optimisation and the robustness of the feedback controller obtained are highlighted.



(a) Movement of the hopping robot (simulation)



(b) Forward velocity, body height, joint angles



(c) Control commands, and hip and leg stiffness

Figure 5.2: Application of the optimised controller to achieve steady state hopping with multiple cycles. External perturbations were applied in order to evaluate the robustness of the obtained optimal *feedback* control law. The robot was able to continue hopping after the application of the perturbations, without falling over.

In addition to the discussion concerning the developed approach presented in Sec. 4.11, we note that in the hopping example enforcing the feasibility of the switching condition at touch-down is more critical. Hence, the weights of the cost were empirically adjusted in order to reduce the mismatch. The temporal optimisation also plays a crucial role, as finding a suitable movement duration is effective in reducing the error. The results obtained showcase how certain mismatches in the switching conditions can be overcome by using the feedback controller. Further investigation is required on the extent of these capabilities.

In the next chapter we extend our approach to incorporate model learning in order to address modelling uncertainties of the system dynamics [Mitrovic, 2010], which pose a challenging problem, especially in the case of underactuated systems.

In the previous chapters we presented a systematic methodology for movement optimisation with multiple phases and switching dynamics in robotic systems with variable stiffness actuation. We showed the effectiveness of the proposed approach in numerical simulations and on a hardware platform.

The investigation thus far assumes that the dynamics of the controlled systems are given as an analytic rigid body dynamics model and that model is not changing over time. For the purpose of our previous tasks, this proved to be an effective technique. In the current chapter we extend our approach to tasks where the dynamics are unknown, too complex to be accurately described by a rigid body dynamics model and/or subject to methodical changes.

We formulate an online adaptation scheme capable of systematically improving the multi-phase plans (stiffness modulation and position command) and switching instances. By incorporating a locally weighted non-parametric learning method to model the discrepancies in the system dynamics, the dynamics model is improved on the fly. This is demonstrated on a realistic model of a VSA brachiating system with excellent adaptation results.

6.1 INTRODUCTION

The accuracy of model-based control is significantly dependent on that of the models themselves. Traditional robotics employs models obtained from mechanical engineering insights. Kinematic equations will provide accurate information about the evolution of a rigid body configuration, given a precise knowledge of its geometry. Similarly, the dynamics equation can incorporate well modelled factors such as inertia, Coriolis and centrifugal effect or external forces.

However, there are certain elements that cannot be fully captured by these models, such as friction from the joints or resulting from cable movement [Siciliano and Khatib, 2008], which can vary in time. The introduction of flexible elements in the structure of a system increases the complexity of the model and makes the identification of accurate dynamics significantly more difficult. Additionally, dur-

ing operation, the robot can suffer changes in its mechanical structure due to wear and tear or due to the use of a tool (which modifies the mechanical chain structure) [Sigaud et al., 2011].

On-line adaptation of models can provide a solution for capturing all these properties. Early approaches, such as on-line *parameter identification* [Sciavicco and Villani, 2009], which tunes the parameters of a predefined model (dictated by the mechanical structure) using data collected during operation, proved sufficient for accurate control and remained a popular approach for a long time [Atkeson et al., 1986; Khalil and Dombre, 2004]. The increased complexity of latest robotic systems demands novel approaches capable of accommodating significantly non-linear and unmodelled robot dynamics. Successful *non-parametric model learning* methods use supervised learning to perform system identification with only limited prior information about its structure - removing the restriction to a fixed model structure, allowing the model complexity to adapt in a data driven manner.

In this chapter, we build on prior efforts to engage this techniques for robot control [Mitrovic et al., 2010a; Nakanishi et al., 2005; Nguyen-Tuong and Peters, 2011] and apply this in the context of multiphase variable impedance movements. Indeed, adaptive model learning has been used successfully in a wide range of scenarios such as inverse dynamics control [Nguyen-Tuong et al., 2009; Vijayakumar et al., 2002], inverse kinematics [Herbort et al., 2010; Ting et al., 2009], robot manipulation and locomotion [Kalakrishnan et al., 2009; Schaal et al., 2002].

6.1.1 Adaptive Learning for Optimal Control

Classical OC is formulated using an analytic dynamics model, but recent work [Atkeson et al., 1997; Mitrovic et al., 2010d] has shown that combining OC with dynamics learning can produce a powerful and principled control strategy for complex systems with redundant actuation.

Depending on which type of data is observed, the models obtained can be: forward models, inverse models, mixed models and multi-step prediction models [Nguyen-Tuong and Peters, 2011]. There are various control architectures that integrate learning mechanisms (e.g. direct modelling, indirect modelling, distal teacher learning), depending on the way in which the model is being adapted while in use. *Direct modelling* represents a direct mapping between the inputs and the outputs in order to improve trajectory tracking performance by predicting the required motor commands for a desired target state. In *indirect modelling* the learn-

ing is driven by an error, such as the output of a feedback controller, while in the *distal teacher learning* scenario the learning of an inverse model (used for control) is guided by a forward model. The last two approaches are goal-directed, and can be used in cases where the mapping problem is ill-posed (e.g. for a redundantly actuated system). While direct modelling can be applied both on-line and off-line, indirect modelling and distal teacher learning have to be performed on-line. The indirect modelling approach learns a particular output solution, which has a minimal feedback error, while the distal teacher provides a globally consistent model [Nguyen-Tuong and Peters, 2011].

Most of these examples are typically based on a desired trajectory that has to be tracked (as accurately as possible). In the context of optimal feedback control, forward dynamics models are used to obtain both the optimal trajectory and control law. Sustained effort has been made to obtain reliable non-linear forward dynamics models in the context of adaptive control theory [Atkeson and Morimoto, 2002; Nakanishi and Schaal, 2004].

In [Mitrovic et al., 2010d], using online (non-parametric) supervised learning methods, an adaptive internal model of the system dynamics is learned. The model is used afterwards to derive an optimal control law. This approach, named (*iLQG*) with learned dynamics (*iLQG-LD*), proved efficient in a variety of realistic scenarios, including problems where the analytic dynamics model is difficult to estimate accurately or subject to changes and the system is affected by noise [Mitrovic et al., 2008, 2010d].

In *iLQG-LD* the update of the dynamics model takes place on a trial-by-trial basis [Mitrovic et al., 2008]. The operating principle (depicted in Fig. 6.1) is to (i) compute the *iLQG* solution, (ii) run the obtained control law on the plant and collect data, (iii) use the plant data to update the dynamics model.

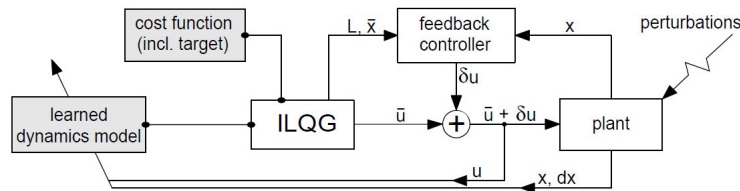


Figure 6.1: The *iLQG-LD* learning and control scheme [Mitrovic et al., 2008]

The initial state and the cost function (which includes the desired final state) are provided to the *iLQG* planner, alongside a preliminary model of the dynamics. An initial (locally optimal) command sequence \bar{u} is generated, together with the

corresponding state sequence $\bar{\mathbf{x}}$ and feedback correction gains \mathbf{L} . Applying the feedback controller scheme, at each time step the control command is corrected by $\delta \mathbf{u} = \mathbf{L}(\mathbf{x} - \bar{\mathbf{x}})$, where \mathbf{x} is the true state of the plant. The model of the dynamics is updated using the information provided by the applied command $\mathbf{u} + \delta \mathbf{u}$ and observed state \mathbf{x} .

6.1.2 Locally Weighted Projection Regression (LWPR)

This methodology employs a *Locally Weighted Learning (LWL)* method, called the *Locally Weighted Projection Regression (LWPR)*, to train a model of the dynamics in an incremental fashion.

LWL algorithms are nonparametric local learning methods that proved successful in the context of (online) motor learning scenarios [Atkeson et al., 1997]. Incremental LWL was proposed in [Schaal and Atkeson, 1998] (as the *receptive field weighted regression (RFWR)* method) in order to provide fast learning rates and computational efficiency. RFWR works by allocating resources in a data driven fashion, allowing on-line adaptation to changes in the behaviour.

Since advanced mechanical systems can output data with potentially irrelevant and/or redundant dimensions, it was shown that the motion data from such system can often be locally represented using low dimensional distributions [Atkeson et al., 1997]. LWPR [Vijayakumar et al., 2005] exploits this property by finding locally low dimensional projections that are used to perform a more efficient local function approximation. The regression function is constructed by combining local linear models. During training the parameters of the local models (locality and fit) are updated using incremental *partial least squares (PLS)*. PLS projects the input onto a small number of directions in the input space along the directions of maximal correlation with the output and then performs linear regression on the projected inputs. This makes LWPR suitable for high dimensional input spaces. Local models can be pruned or added on an as-need basis (e.g., when training data is generated in previously unexplored regions). The areas of validity (receptive fields) of each local model are modelled by Gaussian kernels. LWPR keeps a number of variables that hold sufficient statistics for the algorithm to perform the required calculations incrementally. The method proved to be a robust and efficient technique for incremental learning of nonlinear models in high dimensions [Mitrovic et al., 2010d].

We incorporate the iLQG-LD scheme into our approach involving learning the dynamics of a brachiation system with VSA capabilities and employing it in planning for locomotion tasks. The method proved capable of adapting to changes in the system's properties and provided a better accuracy performance, than the deterministic optimisation. Based on these results, iLQG-LD could be a strong candidate for optimal control strategy for more complex hardware systems.

6.1.3 *Notes on practical issues concerning model learning and adaptive control methods*

In the previous sections we provided a brief outline of model learning techniques for adaptive (optimal) control with a focus on the LWPR method in the context of iLQG-LD. Employing this methodology for a specific task is not always straightforward. In this section we present an overview of the problems encountered in customising a machine learning method for a robotics application. We describe some of the associated approaches for addressing them, including those employed in the context of our work.

A *Model Learning*

The accuracy of a model is strongly dependent on the training data provided, which needs to be sufficiently rich for generalisation and has to cover the state space associated with the scope of the model.

Many approaches for obtaining the required data have been devised, such as sampling trajectories generated by random point-to-point movements [Swevers et al., 1997] or by constrained optimisation [Gautier and Khalil, 1992]. To enhance the richness of the data additional excitation (persistent excitation [Narendra and Annaswamy, 1987]) is often required. This excitation can be naturally provided by the nature of the system (e.g., perturbations affecting an aircraft) or artificially generated (e.g., adding random perturbations to a system). In our work we employ a similar technique by adding random perturbations to the solution obtained at each step of the iLQG-LD and collecting data from the resulting trajectory.

Special attention is necessary to ensure the data is sufficiently smooth, a requirement of most machine learning methods [Nguyen-Tuong and Peters, 2011]. Discontinuities of non-smooth functions can be approximated by kernel methods [Rasmussen, 2006] or by local models and learning how to switch between these models [Toussaint and Vijayakumar, 2005].

Robotic systems with high DOFs pose a challenge, due to the increased dimensionality of the training data. This is normally handled by employing dimensionality reduction approaches, which have been studied in machine learning for a long time [Tenenbaum et al., 2000]. While some methods require special preprocessing to eliminate the irrelevant and/or redundant information [Nguyen-Tuong and Peters, 2010b], the LWPR method is able to handle both redundant and irrelevant training data [Vijayakumar et al., 2005].

If prior knowledge is available (in the form of a parametric model), this can be incorporated into the learning method, rendering a *semiparametric model*. In [Nguyen-Tuong and Peters, 2010a] this approach was used to learn an accurate model with limited training data. In our experiments we use a similar technique by exploiting the information available in the form of a preliminary (erroneous) analytic model of the dynamics.

Most machine learning methods are designed for offline learning and run in batch mode [Nguyen-Tuong and Peters, 2011]. The major motivation for online learning, (which LWPR permits), is that one cannot cover the regions of interest in advance, but these can be discovered during execution. This allows for incremental learning and partial forgetting of previous information. It constitutes a move towards continuous adaptation, required in order to allow robots increased autonomy [Thrun and Mitchell, 1995]. Modelling nonstationary systems is a necessity for modern robotic applications confronted with time dependent dynamics [Petkos et al., 2006].

Other practical issues include feature selection and missing data (for robustness and reliability of the prediction). Probabilistic inference methods can deduce the missing information, based on the training data. An advantage of recent methods, such as LWPR, is that feature selection is incorporated within the algorithm [Vijayakumar et al., 2005].

In [Nakanishi et al., 2005] the authors provide empirical evidence of the performance benefits of LWL over cooperative learning approaches for non-linear function approximation. They show how the former can provide a much more accurate fit even with a limited number of local models. When the number of models increases, LWL does not suffer from the overfitting issues exhibited by cooperative learning approaches. In addition, the study demonstrated that the global accuracy of a LWL method upper is bounded by that of its individual local linear models. Since these are updated based on the curvature of the function learned, this ensures that the upper bound on the model is kept small.

B Adaptive control

Controllers able to adapt to changes in the behaviour of the systems are a desirable attribute for many robotic applications. Compared to their non-adaptive counterparts, which use just a priori information [Dumont et al., 2002], adaptive controllers incorporate data from the systems behaviour in order to change their properties or those of the system's.

Classic adaptive controllers react to sudden changes in the expected behaviour by updating the control parameters and/or the dynamics model, in order to enhance performance, such as trajectory tracking [Narendra and Annaswamy, 2012]. By contrast *iterative learning controllers (ILC)* make use of the long term history of the behaviour, gradually improving the performance of the system. They are based on the assumption that the same task will be performed repeatedly over time. Alternatively, *reinforcement learning (RL)* approaches achieve learning by trial and error, without specific instructions on which actions to take. By using the information from previous trials to build a reward function, the system learns to select the actions providing maximal reward.

In [Mitrovic, 2010] the dual nature of iLQG-LD as a hybrid between ILC and RL approaches is indicated. The method performs iterative improvements based on data from each trial (similar to ILC), but is also less concerned with tracking the previous trajectory and delivers an new optimal feedback law at each iteration (thus, resembling the RL methodology).

In terms of performance, previous work has shown empirically that iLQG-LD can deliver a stable performance in the presence of noise and can adjust to complex systematic changes in the (non-linear) dynamics [Mitrovic et al., 2010d]. Most challenges relate to the magnitude of the non-linearity of the dynamics, the tuning of the LWPR parameters and the fact that the size of training data requirement increases with the dimensionality of the system. As previous studies have suggested [Mitrovic, 2010], the latter of these issues can be alleviated by incorporating existing knowledge (analytic dynamics model) in combination with the LWPR learning.

Previous applications of iLQG-LD have focused on relatively low dimensional, well-tempered dynamics, mostly in simulation environments. As remarked in [Mitrovic, 2010], the direct application to a robotic platform raises a few difficulties. Data collection on a hardware is more demanding (even without considering potential hardware failure) and the noise affecting the system can be more pronounced and less predictable. In cases where discontinuities occur, the method is

not equipped with mechanisms that will deal directly with them. Also, successful use of LWPR in practice requires a certain amount of experience and good intuition (e.g., parameter tuning).

The main challenge is directing the learning on the specific area required for the task and obtaining a reliable model which is as accurate as possible. Additional practical challenges relate to the necessity of adequate and reliable sensing. In our experiments we address these aspects by devising an exploration strategy which pre-trains the model with information within the region of interest for the given task. We also provide a set of sensors that allow the LWPR method to learn the effect of the changes in the model structure and recover the system model.

So far we discussed the relevant background on model learning and adaptive control, the associated challenges and the approaches of addressing them, with a focus on our chosen methodology. In the next section we present the formulation used and the experimental setup of our investigation.

6.2 MULTI-PHASE OPTIMISATION WITH ADAPTIVE DYNAMICS LEARNING

In Sec. 4.5, we introduced a general formulation of optimal control problems for tasks with multiple phase movements including switching dynamics and discrete state transition arising from iterations with an environment. Given a rigid body dynamics formulation of a robot with a VSA model, a hybrid dynamics representation with a composite cost function is introduced to describe such a task (Sec. 4.5). In this section we introduce the changes dictated by the use of the LWPR method in the context of iLQG-LD, for integration within our approach.

6.2.1 *Hybrid Robot Dynamics with Variable Stiffness Actuation for Time-based Switching and Discrete State Transition*

We use the hybrid dynamics representation (Sec. 4.5.1) to model multi-phase dynamics including contacts with the environment [Caldwell and Murphey, 2011b], as depicted in Fig. 4.2

To describe multi-phase movements, we consider multiple sets of robot dynamics, as described by (4.3), (4.4). An individual rigid body dynamics model is defined for each associated phase of the movement as a subsystem (4.6). The servo motor dynamics in the VSA are modelled as a critically damped second order dynamics (4.7). We assume that the range of control command \mathbf{u} is limited be-

tween \mathbf{u}_{\min} and \mathbf{u}_{\max} . As in Sec. 4.5, we consider a VSA model in which the joint torques are given in the form (3.3).

We consider the state space representation as the combined plant dynamics consisting of the rigid body dynamics (4.6) and the servo motor dynamics (4.7), with $\mathbf{x} = [\mathbf{x}_1^T, \mathbf{x}_2^T, \mathbf{x}_3^T, \mathbf{x}_4^T]^T = [\mathbf{q}^T, \dot{\mathbf{q}}^T, \mathbf{q}_m^T, \dot{\mathbf{q}}_m^T]^T \in \mathbb{R}^{2(n+m)}$ the state vector consisting of the robot state and the servo motor state.

Employing the iLQG-LD framework, we aim to create an accurate model of the dynamics model for the real hardware using supervised learning. We assume the existence of a preliminary analytic dynamics model that takes the form presented in (4.6, 4.7), which is inaccurate (due to various factors such as: the inability of the rigid body dynamics to incorporate all the elements of the system's behaviour or changes suffered during operation).

We use the LWPR method to model the error between the true behaviour of the system and the initial model provided. Thus we replace the dynamics \mathbf{f}_i in (4.8) with the composite dynamics model \mathbf{f}_{ci} :

$$\dot{\mathbf{x}} = \mathbf{f}_{ci}(\mathbf{x}, \mathbf{u}) = \tilde{\mathbf{f}}_i(\mathbf{x}, \mathbf{u}) + \bar{\mathbf{f}}_i(\mathbf{x}, \mathbf{u}), \quad (6.1)$$

where $\tilde{\mathbf{f}}_i \in \mathbb{R}^{2(n+m)}$ is the initial inaccurate model and $\bar{\mathbf{f}} \in \mathbb{R}^{2(n+m)}$ is the LWPR model mapping the discrepancy between $\tilde{\mathbf{f}}_i \in \mathbb{R}^{2(n+m)}$ and the behaviour of the system. We note that the changes introduced by iLQG-LD only affect the dynamics modelling in (4.3), while the instantaneous state transition mapped by Δ in (4.4) remains unchanged.

For the given hybrid dynamics, in order to describe the full movement with multiple phases, we consider the same composite cost function, as in (4.9).

6.2.2 Optimal Control of Switching Dynamics and Discrete State Transition

In brief, the iLQR method (Sec. 2.3.1), upon which our developed methodology (Sec. 4.5) is built, solves an optimal control problem for the locally linear quadratic approximation of the nonlinear dynamics and the cost function around a nominal trajectory $\bar{\mathbf{x}}$ and control sequence $\bar{\mathbf{u}}$ in discrete time, and iteratively improves the solutions.

In order to incorporate switching dynamics and discrete state transition with a given switching sequence, the hybrid dynamics (4.3) and (4.4) are linearised in discrete time around the nominal trajectory and control sequence as described by (4.10) : (4.13).

When using the composite model of the dynamics (\mathbf{f}_c) introduced in (6.1) the linearisation of the dynamics is provided in two parts. The linearisation of $\tilde{\mathbf{f}}$ is obtained by replacing \mathbf{f} with $\tilde{\mathbf{f}}$ in (4.10) and (4.12). The derivatives of the learned model ($\tilde{\mathbf{f}}$) are obtained analytically by differentiating with respect to the inputs $\mathbf{z} = (\mathbf{x}; \mathbf{u})$ as suggested in [Atkeson et al., 1997]:

$$\begin{aligned} \frac{\partial \tilde{\mathbf{f}}(\mathbf{z})}{\partial \mathbf{z}} &= \frac{1}{W} \sum_k \left(\frac{\partial w_k}{\partial \mathbf{z}} \psi_k(\mathbf{z}) + w_k \frac{\partial \psi_k}{\partial \mathbf{z}} \right) - \frac{1}{W^2} \sum_k w_k(\mathbf{z}) \psi_k(\mathbf{z}) \sum_l \frac{\partial w_l}{\partial \mathbf{z}} \\ &= \frac{1}{W} \sum_k (-\psi_k w_k \mathbf{D}_k(\mathbf{z} - \mathbf{c}_k) + w_k \mathbf{b}_k) + \frac{\tilde{\mathbf{f}}(\mathbf{z})}{W} \sum_k w_k \mathbf{D}_k(\mathbf{z} - \mathbf{c}_k), \end{aligned} \quad (6.2)$$

where:

$$\frac{\partial \tilde{\mathbf{f}}(\mathbf{z})}{\partial \mathbf{z}} = \begin{pmatrix} \partial \tilde{\mathbf{f}} / \partial \mathbf{x} \\ \partial \tilde{\mathbf{f}} / \partial \mathbf{u} \end{pmatrix}. \quad (6.3)$$

Since there are no changes on the encoding of the instantaneous state transition (4.4) the equations in (4.11) and (4.13) remain unchanged for the iLQG-LD framework. With the above modification (6.2), the developed optimisation methodology is applied as described in Sec. 4.5 to obtain the locally valid feedback control law.

6.3 BRACHIATION SYSTEM DYNAMICS

We evaluate the effectiveness of the approach on a robot brachiation task which incorporates switching dynamics and multiple phases of the movement in a realistic VSA actuator model. We consider a two-link underactuated brachiating robot with a MACCEPA [Ham et al., 2007] variable stiffness actuator, as introduced in Sec. 4.6.

The equation of motion of the system used takes the standard rigid body dynamics form (4.33). The servo motor dynamics are approximated by a second order system with a PD feedback control, as mentioned in (4.7). As in the previous study, the bandwidth of the actuator is $\alpha = \text{diag}(20, 25)$. The range of the commands of the servo motors are limited as $u_1 \in [-\pi/2, \pi/2]$ and $u_2 \in [0, \pi/2]$.

The model parameters are shown in Table 6.1 and the MACCEPA parameters with the spring constant $\kappa_s = 771$ N/m, the lever length $B = 0.03$ m, the pin displacement $C = 0.125$ m and the drum radius $r_d = 0.01$ m.¹

Robot parameters		i=1	i=2	i=1
Mass	m_i (kg)	1.390	0.527	1.240
Moment of inertia	I_i (kgm ²)	0.0297	0.0104	0.0278
Link length	l_i (m)	0.46	0.46	0.46
COM location	l_{ci} (m)	0.362	0.233	0.350
Viscous friction	d_i (Nm/s)	0.03	0.035	0.03

Table 6.1: Model parameters of the two-link brachiating robot. The final column shows the change of parameters of the first link of the system under the changed mass distribution described in Sec. 6.4

6.4 EXPERIMENTAL SETUP

To test the efficiency of our approach we create a scenario where the difference between the true and the assumed model is caused by a change in the mass (and implicitly mass distribution) on one the links (i.e. the mass of the true model is smaller by 150 g (located at the joint) on link $i = 1$). The changed model parameters are shown in the right column of Table 6.1.

The error in the dynamics manifests itself only in the joint accelerations. Thus, we require to map the error just in those two dimensions, reducing the dimension of the output of the LWPR model $\tilde{\mathbf{f}}$ from $n = 8$ to 2, where the predictions are added on the corresponding dimension of $\tilde{\mathbf{f}}$ within \mathbf{f}_c (6.1). Note that different discrepancies could necessitate estimation of the full 8-dimensional state error.

In line with previous work, we will demonstrate the effectiveness of the proposed approach on a multi-phase, asymmetric swing-up and brachiation task with a VSA while incorporating *continuous, online model learning*. Specifically, in the multi-phase task, the robot swings up from the suspended posture to the target at $d_1 = 0.40$ m and subsequently moves to the target located at $d_2 = 0.42$ m and $d_3 = 0.46$ m, respectively.

¹ The MACCEPA parameters are the same as described in Chapter 4.

Since the system has an asymmetric configuration and the state space of the swing-up task is significantly different from that of a brachiation movement we proceed by first learning a separate error model for each phase. The procedure used is briefly described in Algorithm 3. The initial exploration loop is performed in order to pre-train the LWPR model \tilde{f}_i (as an alternative to applying motor babbling), the latter loop is using iLQG-LD to refine the model in an online fashion. In our experiments the training data is obtained by using a simulated version of the true dynamics, which is an analytic model incorporating the discrepancy. For details on the set-up of the LWPR model, initialisation and parameter selection, we refer the interested reader to Appendix F.

Algorithm 3 Description of the learning and exploration procedure

Given:

- analytic dynamics for one configuration \tilde{f}_i and start state x_0
- thresholds for target reaching ε_T and model accuracy ε_M
- the associated cost function J (including desired target x_T)
- p number of initial exploration training episodes

Initialise

- $\tilde{f}_i(x, u)$; $f_{ci}(x, u) = \tilde{f}_i(x, u) + \bar{f}_i(x, u)$

repeat

- generate \bar{u}, \bar{x}, L using $\tilde{f}_i(x, u)$ for an randomised exploration target
- apply the solution to the true dynamics
- train the model on the collected data
- apply η perturbed versions of the solution to the true dynamics
- train the model on the collected data

until p training episodes have been performed

repeat

- apply iLQG-LD for target x_T

until ε_T and $nMSE$ conditions are met

6.4.1 Individual Phase Learning

Using the traditional OC framework in the presence of an accurate dynamics model, the multi-phase task described in Sec. 4.8 was achieved with a position error of just 0.002 m. Once the discrepancy detailed in Sec. 6.4 is introduced, the planned solution is no longer valid and the final position deviates from the desired target (Fig. 6.2, blue line). We deploy the iLQG-LD framework in order to learn the new behaviour of the system and recover the task performance.

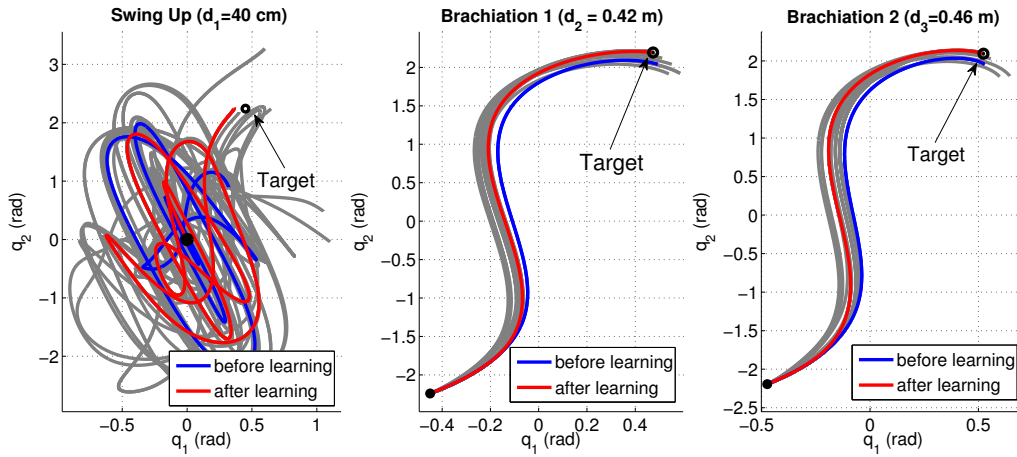


Figure 6.2: Individual phase learning: Comparison of the final position achieved (for each individual phase) when using the initial planning (erroneous model - blue) and the final planning (composite model - red). Intermediary solutions obtained at each step of the iLQG-LD run are depicted in grey.

As a measure of the model accuracy we use the *normalised mean square error* (nMSE) of the model prediction on the true optimal trajectory (if given access to the analytic form of the true dynamics of the system). The nMSE is defined as $nMSE(y, \tilde{y}) = \frac{1}{n\sigma_y^2} \sum_{i=1}^n (y_i - \tilde{y}_i)^2$ where y is the desired output data set of size n and \tilde{y} represents the LWPR predictions. The evolution of the nMSE at each stage of the training for every phase is shown in Fig. 6.3.

In the first part (pre-training phase in Fig. 6.3) we generate ($p = 7$) random targets around the desired x_T . A movement is planned for these targets using the assumed model (\tilde{f}). The obtained command solution is then applied to the simulated version of the true dynamics, using a closed loop control scheme. We repeat the procedure for a set of 10 noise contaminated versions of the commands. The collected data is used to train the model.

This pre-training phase seeds the model with information within the region of interest, prior to using it for planning. This reduces the load on the iLQG-LD by lowering the number of iterations required for convergence. For each phase of the movement, at the end of the procedure, the planned trajectory matched the behaviour obtained from running the command solution on the real plant (the final nMSE has an order of magnitude of 10^{-4}).

Overall the discrepancy is small enough to allow reaching the desired end effector position within a threshold of $\epsilon_T = 0.040$ m accuracy². Fig. 6.2 shows the effect

² The error in the swing up task is 0.033 m, while for brachiations the value is 0.004 m.

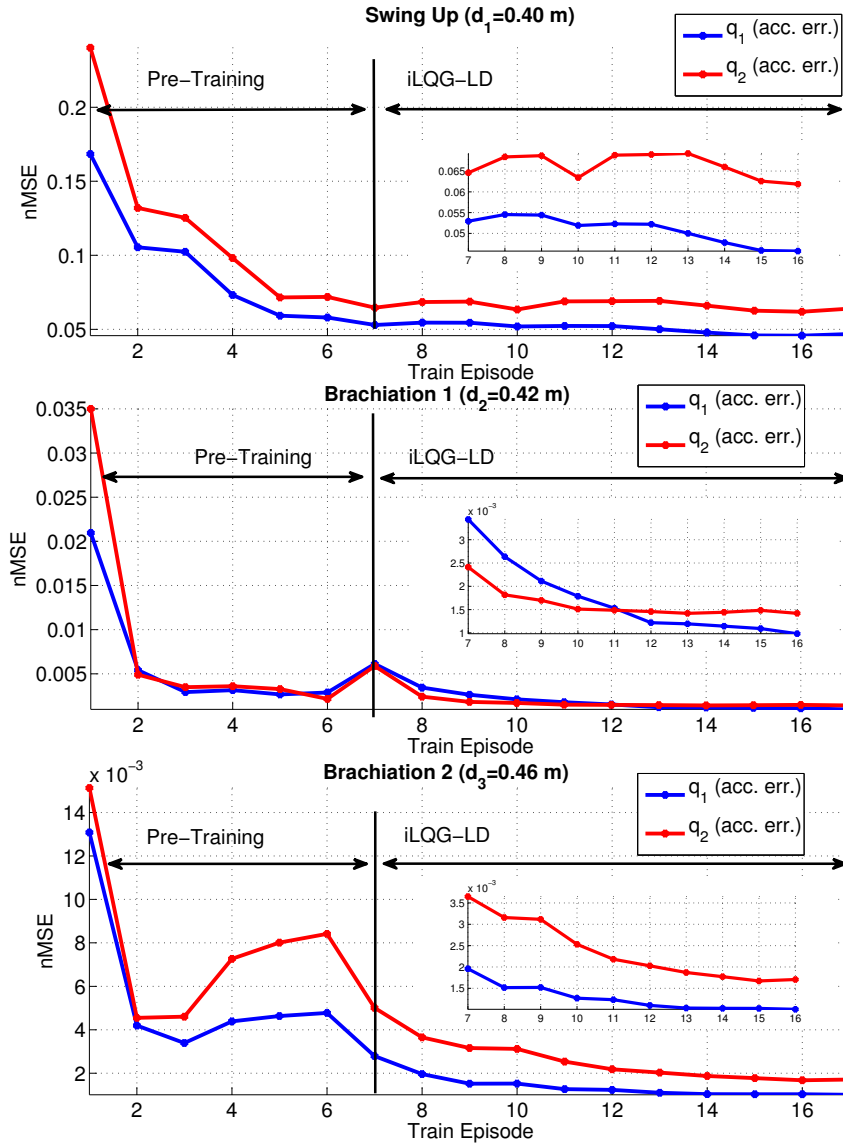


Figure 6.3: Individual phase learning: Evolution of the nMSE for each phase of the movement, at each episode.

of the learning by comparing the performance of the planning with the erroneous model and with the composite model obtained after training.

6.4.2 Multi-phase Performance

In the previous section we showed that our approach to iLQG-LD is able to cope with the requirements of the task in each phase of the movement. For a full solution we use the newly learned models from each phase to obtain the global solution for the multi-phase task wrt. the composite cost function J (4.46). We use

the phase optimal solutions obtained at the previous stage as the initial command sequence, the resulting behaviour is displayed in Fig. 6.4.

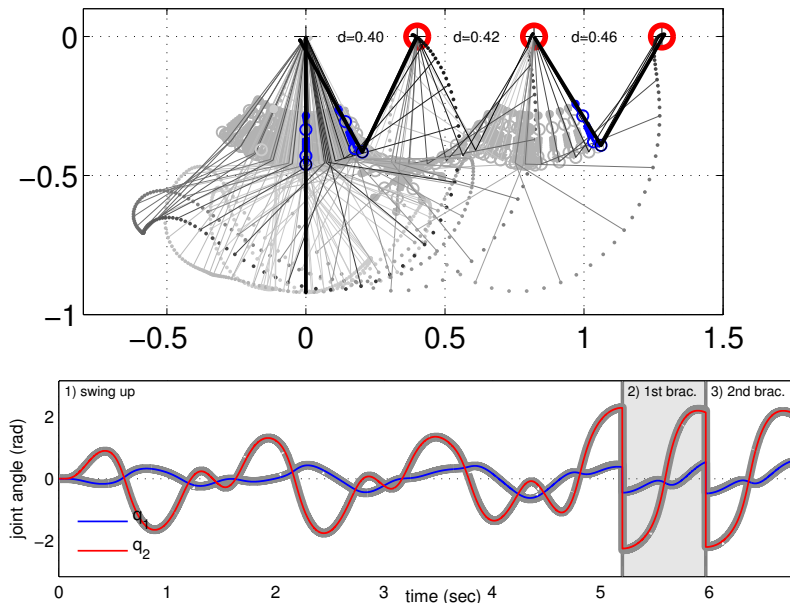


Figure 6.4: Performance of the fully optimised multi-phase task using the composite model. Thick grey lines: planned movement. Red and blue lines: actual system movement.

The planner is able to use the learned model to achieve the intermediary and final goals, while the expected behaviour provides a reliable match to the actual system's behaviour ³. The cost of multi-phase optimised solution ($J = 35.13$) is significantly lower than the sum of the costs of the individual phase solutions ($J = 44.45$) ⁴.

6.4.3 Performance of Learning

In the previous experiment, we investigated a single (arbitrarily chosen) mass distribution discrepancy. Next, we investigate the capacity of our approach to cope with a wider range of mismatched dynamics. For this, we consider the magnitude of the change that is bounded by the capability of the altered system to achieve all the phases of the movement presented in Fig. 6.4. We define these bounds as the limit values of mass change that allow the same accuracy in task execution, under

³ We consider that if the position at the end of each phase is within our prescribed threshold $\varepsilon_T = 0.040$ of the desired target the system is able to start the next phase from the ideal location, thus resembling the effect of the grabber on the hardware.

⁴ The cost incurred when applying the obtained solutions to the real system dynamics.

the same cost function (4.46). The corresponding values for these limits, found empirically, are -0.200 kg and $+0.300$ kg, respectively ⁵.

We apply our framework to the scenarios where the mass has been altered to these boundary values and demonstrate the result on just one of the phases, namely the first brachiation move, to study the relative effects.

Fig. 6.5 (a) shows the effect of the alteration introduced, when executing the commands resulting from the initial planning (using the erroneous model). After training and re-planning the model starts approximating the true behaviour of the system, such that in less than 10 iLQG-LD episodes, the system is once again able to reach the desired target, as depicted in Fig. 6.5 (b) and 6.5 (c).

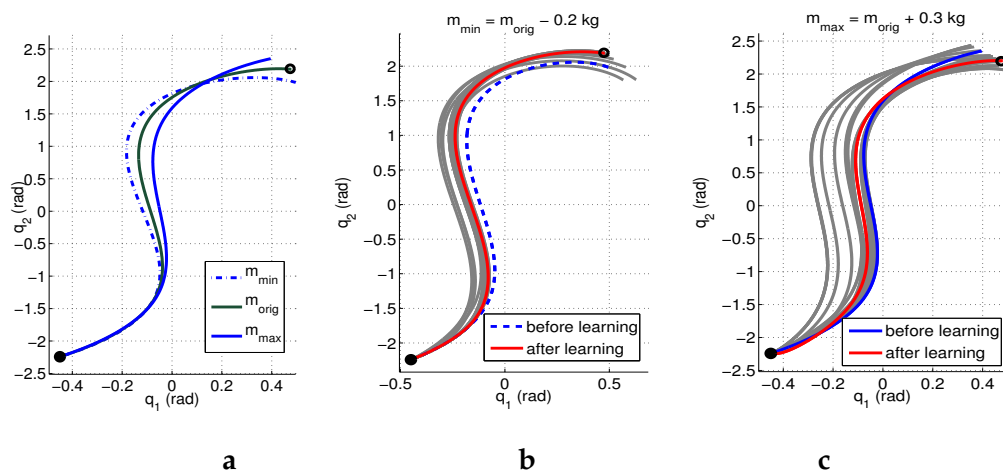


Figure 6.5: **(a)**: Effect of the discrepancy introduced by the limit values on the mass modulation (blue lines). The behaviour when the model match is correct is depicted for comparison (green line). **(b) - (c)**: Comparison of the final position achieved (for each individual phase) when using the initial planning (erroneous model - blue) and the final planning (composite model - red). Intermediary solutions obtained at each step of the iLQG-LD run are depicted in grey. Results for the boundary value discrepancies on the mass distribution.

6.5 IMPEDANCE DISCREPANCY

In the previous section we introduced a model discrepancy by changing the mass (and mass distribution) on one the links (i.e. the mass of the true model is smaller by 150 g (located at the joint) on link $i = 1$). We showed that our online adaptation

⁵ We note that in our experiments, we assume that modulating the mass distribution does not affect the motor dynamics – this represents a simplified scenario. In the real hardware, the speed of the motor dynamics (4.7) is indeed a function of the overall load distribution.

approach is able to learn the behaviour of the system and recover the performance in order to achieve the task at hand.

Next, we investigate the effect of a different type of discrepancy, where the difference between the true and the assumed model is caused by a change in the stiffness of the spring incorporated in the MACCEPA actuator (i.e., $\kappa_s=771(\text{N/m})$ is replaced by $\kappa_s=650(\text{N/m})$)⁶.

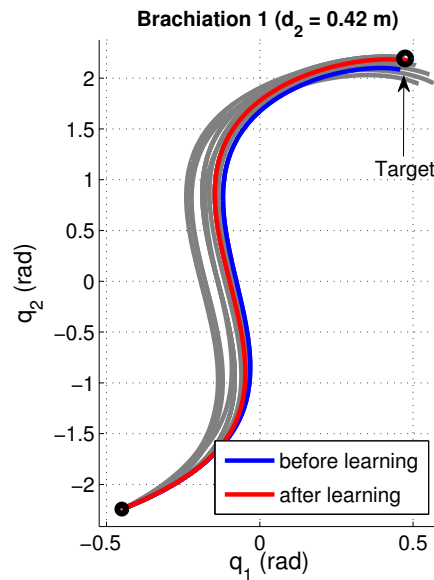


Figure 6.6: Changed spring stiffness κ_s : Comparison of the final position achieved (for a brachiation task) when using the initial planning (erroneous model - blue) and the final planning (composite model - red). Intermediary solutions obtained at each step of the iLQG-LD run are depicted in grey.

The discrepancy affects the behaviour of the system and the planned solution is no longer valid, the final position deviates from the desired target (Fig. 6.6, blue line). The final position of the end effector is located at approx. 0.054 m from the desired target. Applying the same methodology to this new scenario we observe the performance of our approach.

The evolution of the nMSE (of the model prediction on the true optimal trajectory, if given access to the analytic form of the true dynamics of the system) at each stage of the training for every phase is shown in Fig. 6.7. After the pre-training phase the nMSE has an order of magnitude of 10^{-2} . At the end of the procedure, the planned trajectory matched the behaviour obtained from running the command solution on the real plant ($\text{nMSE} \approx [10^{-3}, 10^{-4}]$) with the error of the

⁶ As before, we assume that the discrepancy does not affect the motor dynamics. In the real hardware, the speed of the motor dynamics (4.7) can be affected by a change in the spring stiffness constant κ_s .

final position for the end effector after training 0.008 m. Fig. 6.6 shows the effect of the learning by comparing the performance of the planning with the erroneous model and with the composite model obtained after training.

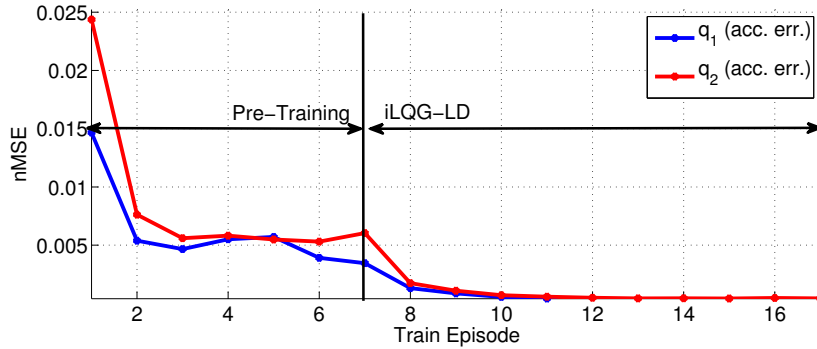


Figure 6.7: Changed spring stiffness κ_s : Evolution of the nMSE at each episode

The results obtained in this investigation suggest that the approach can be employed for several types of discrepancies, caused by a variety of changes in the system's structure.

6.6 TRANSFER LEARNING

In the investigation so far we build a model for each specific task of the multi-phase movement by collecting and training it on data near the domain of each task. This approach can be time consuming, especially for systems with large DOFs and it scales with the number of tasks introduced. A more efficient approach will use a singular (*general model*) and allow the knowledge gained in the context of previous tasks to be reused in improving performance on a new task.

6.6.1 General Background on Transfer Learning

The concept of transferring knowledge between different tasks is not novel. In *transfer learning* knowledge from previous task(s) (*source task(s)*) is applied to a new task (*target task*) in order to facilitate learning in the new context (compared to the scenario of learning for the new task in isolation). Transfer learning has been studied in the context of machine learning [Pan and Yang, 2010; Torrey and Shavlik, 2009] for classification, regression and clustering problems. In the field of robotics most transfer learning work has focused on reinforcement learning assignments [Barrett et al., 2010; Thrun and Mitchell, 1995].

Depending on the relation between the source and target tasks we distinguish two major categories. In i) *inductive transfer learning*, the target and source tasks are different [Ferguson and Mahadevan, 2006], while in ii) *transductive transfer learning* the target and source tasks are the same [Taylor and Stone, 2007b]. For a detailed look at the subcases of each category we refer the reader to the surveys in [Pan and Yang, 2010; Taylor and Stone, 2009]. Some methods in ii) assume that a mapping between the source and target tasks is provided [Taylor and Stone, 2007b], while a more general approach includes learning this mapping [Taylor and Stone, 2007a].

A major challenge is conducting the learning in such a way that will improve performance for related tasks, while avoiding *negative transfer* (i.e., deteriorating the performance for unrelated tasks) [Bonilla et al., 2007]. There is very little work addressing this topic, most of them focus on demonstrating how forcing transfer between two very different tasks can lower performance on the target task [Rosenstein et al., 2005] and analysing the relatedness of the two tasks [Ben-David and Schuller, 2003] in order to assist avoiding negative transfer.

6.6.2 Transfer Learning for Robotic Models

In the field of robotics, most transfer learning work has focused on reinforcement learning assignments [Barrett et al., 2010; Thrun and Mitchell, 1995]. There have been very few attempts to use transfer learning of robot models, but recent developments have demonstrated the effectiveness of such approaches [Bocsi et al., 2013; Um et al., 2014; Williams et al., 2009].

In [Williams et al., 2009] a multi-task learning problem is defined for the inverse dynamics of a robot, under different loads. Multiple Gaussian Processes, sharing the hyperparameters of the model, are used to encode the dynamics, thus sharing information across different tasks. By learning a shared covariance function on input-dependent features and a non-constrained covariance matrix encoding inter-task similarity, the performance is improved across the tasks.

Transfer learning between two distinct robot designs is studied in [Bocsi et al., 2013]. Here, both the source and task datasets are projected onto lower dimensional manifolds. A mapping between these manifolds is found by either i) defining a simple linear transformation (when there is a direct correspondence between the two task datasets) or ii) minimising the distance between the two datasets (e.g., assuming Gaussian distributions of both datasets and minimising the KL divergence). Once the mapping between the two robot systems is found, direct task

transfer is possible. Using previously available information is shown to improve the learning of forward kinematics.

In [Um et al., 2014] an alternative for task-to-task transfer learning is provided, whereby a model is learned for each joint independently, instead of a whole body model. This differs from a multi-task approach since the focus is not improving performance across all tasks, but rather focusing on the current target task. The method uses the modular Newton-Euler formulation of the dynamics to separate the data relevant to each joint. A Gaussian Process regression is then used for each joint to learn the model of the error between an assumed rigid body dynamics model and the true system behaviour.

6.6.3 *Transfer Learning : Forward Dynamics of Brachiation System*

In the previous section we introduced the topic of transfer learning. We showed that, in spite of a limited amount of research addressing robot models in this context, the available results show the potential benefits of using previous knowledge in building the model of the system within the context of a new task.

In this section we perform an investigation into transfer learning in the context of our multi-phase movement task. We use the brachiation system to analyse the effect of using knowledge from a previously available task on the performance within the current task for the LWPR based model.

Subsequently, we compare the results obtained for the multi-phase scenario in Sec. 6.4.2, with the resulting behaviour on the same task using a single *unified model* of the dynamics, instead of individual ones for each phase.

A *Task-to-task Transfer*

For the first transfer of knowledge investigation we define the source task as the swing-up motion and the target task as the final brachiation (i.e., the third phase of the multi-phase task described). In this scenario we know empirically that the two tasks share a segment of the domain space. We use the model learned during the source task, instead of training a new model just on the data from the brachiation task.

In the LWPR learning we set the forgetting factor variables (i.e., initial and final forgetting factors, and the annealing constant for adjusting the forgetting factor) to minimise information loss. Thus, we aim to reduce the effect of negative transfer in the inverse direction. By adding new information, the performance of the pre-

diction on the space explored in earlier stages should not deteriorate drastically. This is required since we wish to use a single unified model in planning for the multi-phase task.

Fig. 6.8 shows the performance of the model learned during the source task, before any learning is applied in the context of the target task. We compare this behaviour with the initial performance of the composite model for the target task without transfer learning. Based on the nMSE on the planned trajectory vs. the real system behaviour and the final pose at the end of the planned movement, we notice that the former performs better before any task specific learning is introduced.

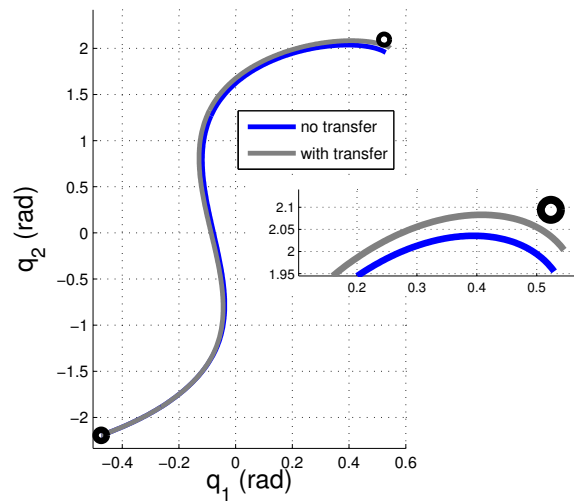


Figure 6.8: Task-to-task Transfer: Performance of planned trajectory on the system before task specific knowledge is applied. Comparison between performance with (grey) and without (blue) transfer learning.

We compare the evolution of the nMSE⁷ of the joint accelerations between the two cases during learning for the target task (Fig. 6.9). The stochasticity of the approach (i.e., represented in the exploration data) plays a significant role in this comparison. Hence, we used data averaged over 5 trials for each scenario and included the variance as the grey error bars along the curves.

The performance of the model in the transfer learning scenario starts off better than that of the model learning from zero. It quickly converges to a steady state performance which is comparable to that of the separate model (Fig. 6.9, insert). While the performance is not significantly better in the transfer scenario

⁷ The nMSE is computed for the prediction of the current model on the true optimal trajectory (if given access to the analytic form of the true dynamics of the system).

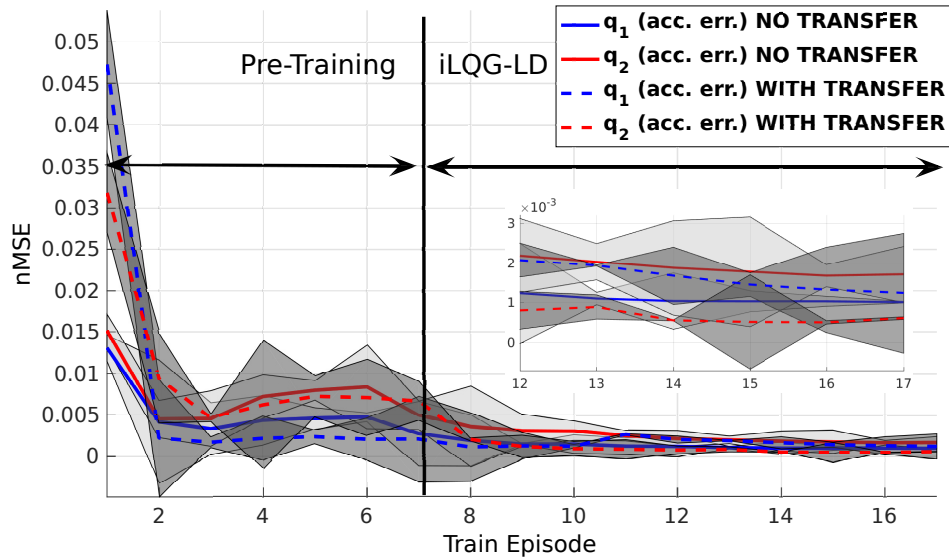


Figure 6.9: Task-to-task Transfer: Evolution of the nMSE at each episode. Comparison between performance with and without transfer learning. The results are averaged over 5 trials in each scenarios. The variance is included in the grey error bars along the curves.

we observe that the convergence to a lower nMSE level can be faster (e.g., episodes #8 : #10, Fig. 6.9). If the task permits, a more relaxed accuracy threshold, a transfer learning approach may lower the time spent in reaching the solution.

B *Inverted Model*

Having investigated task-to-task transfer within the previous experiments, we now turn our attention to transfer learning where the target task is the first brachiation movement (i.e., the second phase of the multi-phase task described). The source tasks are the previously explored phases. We note that in this case the source tasks both use the same handhold configuration, while the target task is using the alternative configuration.

We compare the evolution of the nMSE on the joint accelerations between the two cases during learning for the target task (Fig. 6.10). As in the previous investigation, the results presented are averaged over 5 trials for each scenario with the variance represented by the grey error bars along the curves.

The overall performance of the model in the transfer learning scenario is comparable that of the model learning from zero. The small improvement of the average performance of the model in the transfer learning scenario, including the asymptotic performance (Fig. 6.10, insert), can be justified by the stochastic nature of the

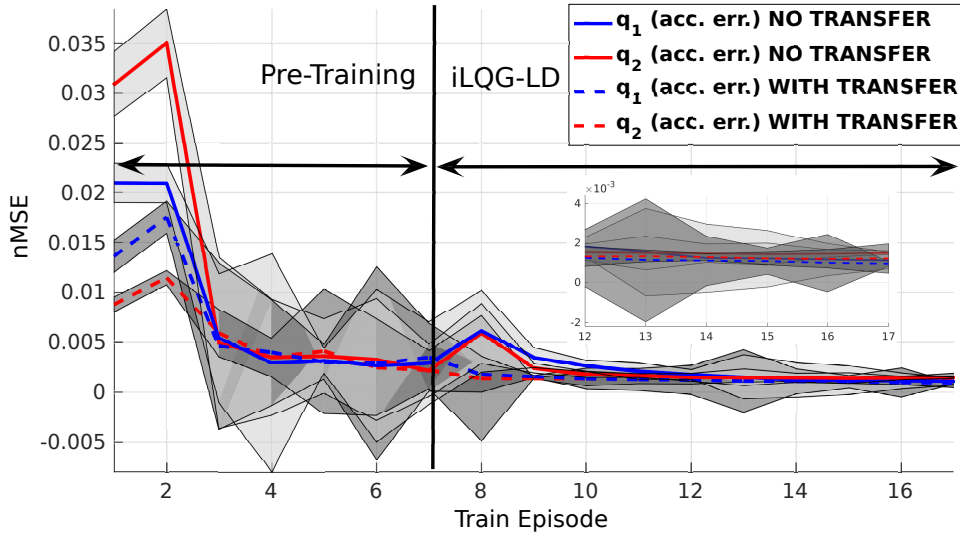


Figure 6.10: Inverted Model: Evolution of the nMSE at each episode. Comparison between performance with and without transfer learning. The results are averaged over 5 trials in each scenarios. The variance is included in the grey error bars along the curves.

exploration data, since the state space of the brachiation task in the alternative hand-hold configuration differs significantly from that of the source task's. We note that in both transfer learning scenarios, further exhaustive investigations are required, in order to establish a statistically significant qualitative comparison.

c Multi-phase Optimisation with Unified Model

In this section we present the results obtained on the multi-phase task with the unified model resulting from the transfer learning experiment, after all the phases of movement have been explored in this context using iLQG-LD (Fig. 6.11).

Compared to the optimisation result in Sec. 6.4.2, where individual models were used within each phase we note that the two cases have similar performances. The costs associated with the individual and unified model, when applying the obtained solutions on the true dynamics are 35.13 and 37.25, respectively. The overall functionality of the solution is preserved (i.e., the task can be achieved within the prescribed accuracy demands).

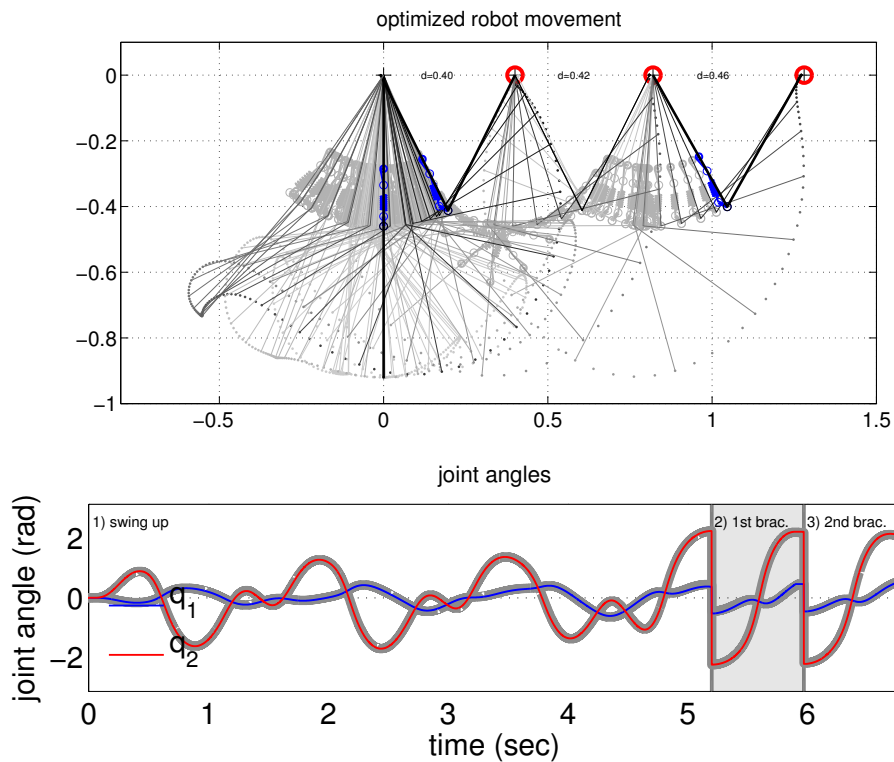


Figure 6.11: Performance of the fully optimised multi-phase task using the composite model. Thick grey lines: planned movement. Red and blue lines: actual system movement.

6.7 HARDWARE EXPERIMENTS: INDIVIDUAL PHASE LEARNING FOR A BRACHIATION ROBOT

In this section we perform a hardware validation of the simulation results obtained in Sec. 6.4. We deploy the developed strategy, once the mass distribution discrepancy has been introduced, in order to learn the new behaviour of the system and recover the task performance.

In the hardware experiment set-up we only have access to readings from two IMU units attached to each link, a potentiometer mounted on the main joint and the internal potentiometers of the actuating motors⁸. The former are fairly accurate and, with adequate filtering, provide reliable readings for estimating the positions and velocities of the robot links. The motors' internal potentiometers however, suffer from high noise; with filtering, their readings can be used as an estimation of the motor position, but they are too unreliable for deriving the motor accel-

⁸ For details please see Sec. 4.9.

erations. For this reason, we scale the dimensionality of the input for the model approximation from 10 ($[\mathbf{q}^T, \dot{\mathbf{q}}^T, \mathbf{q}_m^T, \dot{\mathbf{q}}_m^T, \mathbf{u}^T]^T$) to 8 ($[\mathbf{q}^T, \dot{\mathbf{q}}^T, \mathbf{q}_m^T, \mathbf{u}^T]^T$)⁹.

As previously, in the pre-training phase we generate random targets around the desired x_T and plan movements for those targets using the assumed model ($\tilde{\mathbf{f}}$). We apply the solution to the hardware, alongside a set of 10 noise contaminated versions of the commands. The collected data is used to train the model. In the second phase we apply iLQG-LD as described in Sec. 6.4. The evolution of the nMSE at each stage of the training for every phase is shown in Fig. 6.12.

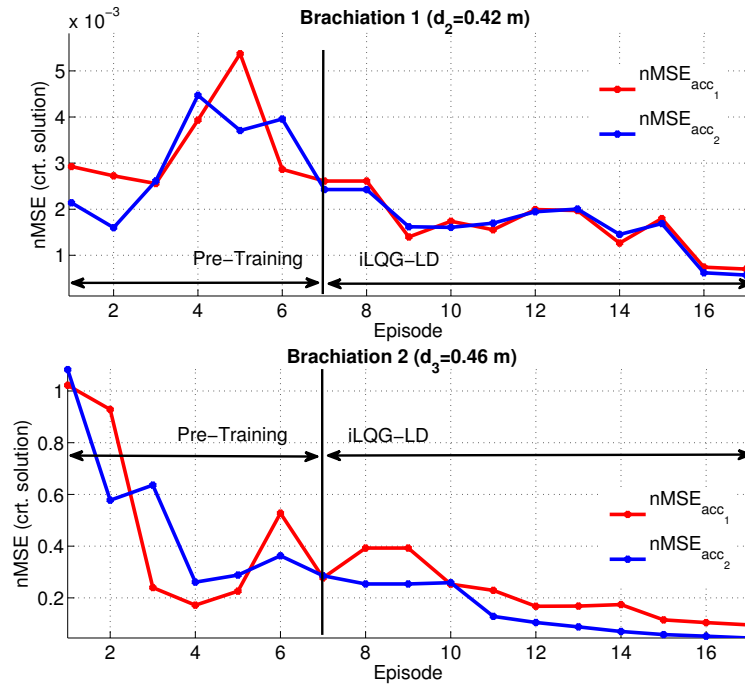


Figure 6.12: Individual Phase Learning in Hardware (mass discrepancy): Evolution of the nMSE for each phase of the movement, at each episode

For each phase of the movement, at the end of the procedure, the planned trajectory matched the behaviour obtained from running the command solution on the real plant (the final nMSE has an order of magnitude of 10^{-4} and 10^{-2} , respectively).

In Fig. 6.13 we compare the performance of the system under the i) solution obtained from the erroneous analytic model (blue) and ii) solution obtained after training the LWPR model (red). Thus, we can observe that the error in the pose

⁹ For this scenario we conducted an investigation to ensure that the input-output relationship for a brachiation movement fulfils the function requirement (idempotence). This is not valid in the case of the swing-up motion, which is the primary reason why we could not attempt that particular task in this context.

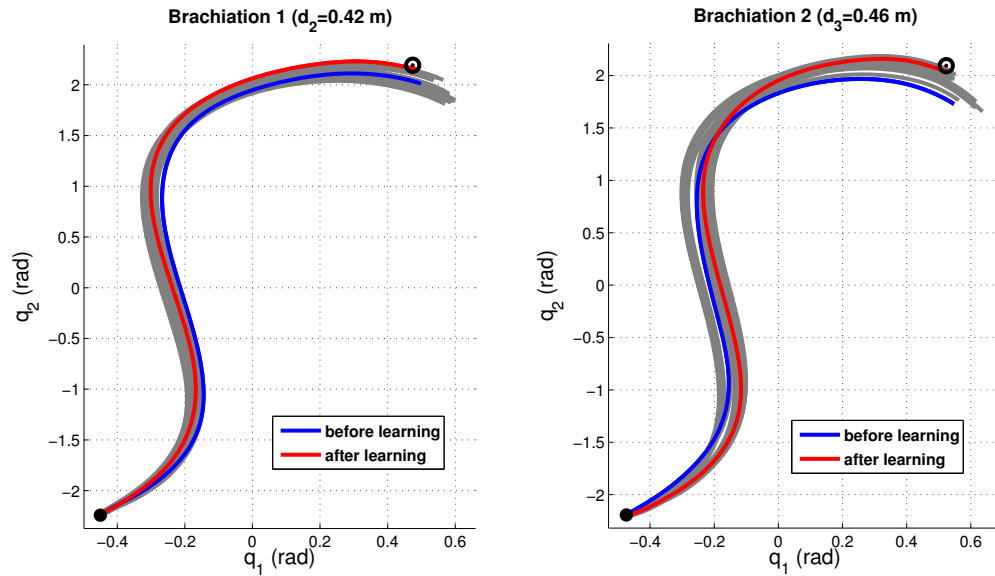


Figure 6.13: Individual Phase Learning in Hardware (mass discrepancy): Comparison of the final position achieved (for each individual phase) when using the initial planning (erroneous model - blue) and the final planning (composite model - red). Intermediary solutions obtained at each step of the iLQG-LD run and exploration trajectories are depicted in grey.

of the end effector (i.e., open gripper) at the end of the allocated time improves significantly in both brachiation tasks from i) 0.0867m and 0.1605m to ii) 0.0161m and 0.0233m, respectively. The final poses are close enough to allow the gripper to compensate for the rest of the error, thus resulting in the final distance of just 0.004. We note that the true poses are actually at the target as the gripper has securely locked on the target bar. The error comes from the variability of the sensor readings.

The experiments conducted address the second and third phases from the multi-phase experiment, described in the previous section (Sec. 6.6.3.c), which correspond to two brachiation movements, one for each hand-hold configuration.

We perform an additional experiment in which, besides the mass change (described above) we also modify the stiffness of the spring in the MACCEPA actuator from $\kappa_s=771(\text{N/m})$ to $\kappa_s=650(\text{N/m})$ as in Sec. 6.5.

As before, at the end of the procedure, the planned trajectory matched the behaviour obtained from running the command solution on the real plant (the final nMSE has an order of magnitude of 10^{-2} , Fig. 6.14).

In Fig. 6.15 we observe the improvement in performance from an initial reaching error (at the end of the planned time) of i) 0.1265m (blue line) to ii) 0.0167m (red line). As before, this allows the gripper to correct for the rest of the error

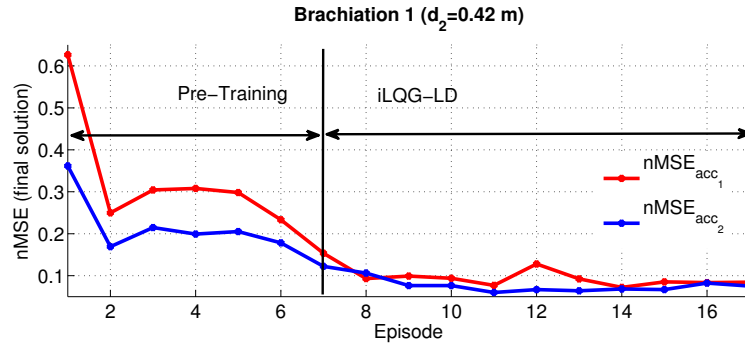


Figure 6.14: Individual Phase Learning in Hardware (mass and stiffness discrepancy): Evolution of the nMSE for each phase of the movement, at each episode

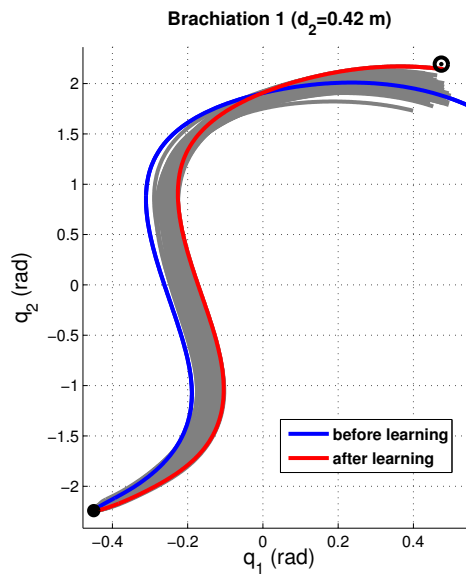


Figure 6.15: Individual Phase Learning in Hardware (mass and stiffness discrepancy): Comparison of the final position achieved when using the initial planning (erroneous model - blue) and the final planning (composite model - red). Intermediary solutions obtained at each step of the iLQG-LD run and exploration trajectories are depicted in grey.

and grab the bar located at the desired target. The results obtained demonstrate that the developed framework is directly applicable for real-life systems with a performance comparable to that of the simulation scenario. (The results presented are depicted in the additional video material).

6.8 DISCUSSION AND CONCLUSIONS

In this chapter we presented an extension of our methodology for movement optimisation with multiple phases and switching dynamics, including variable

impedance actuators. We broaden the approach by incorporating adaptive learning, which allows for adjustments to the dynamics model, based on changes occurred to the system's behaviour, or when the behaviour cannot be fully captured by a rigid body dynamics formulation.

The method employed (in the form of the LWPR algorithm) is particularly suited for certain regression situations, such as non-linear function learning, with the requirement of incremental learning. However, since this method is limited to approximating smooth enough functions we restrict the range of the discrepancies introduced in our experiments accordingly. At the same time, in the case of the brachiation system, the impact map is impervious to changes in the phase dynamics. This is not always true, for example, in the case of the hopper system it depends on the inertia of the system. For such systems, a change in the model parameters would affect not only the phase dynamics, but also the instantaneous impact map, requiring a relearning of both behaviours.

Within our investigation, we showed that the augmented developed methodology can be successfully applied in the case of underactuated systems, such as the brachiation system. We provided results for a range of model discrepancies, in both numerical simulations and real hardware experiments. Future work aims to extend the presented investigate by covering a wider range of discrepancies and dynamics, as mentioned above.

CONCLUSIONS

In this final chapter we summarise the results of our investigations. We put our findings into perspective, discuss the limitations of our approach, and point to directions for future work.

OVERVIEW

In this thesis we started by addressing the concept of optimal control in systems with variable impedance actuators with a study into a little researched sub-field, that of variable damping (alongside stiffness modulation). To our knowledge this was the first time simultaneous and independent stiffness and damping modulations were used in a single actuator.

We showed that, in spite of the theoretical redundancies from a physical perspective, depending on the particularities of the hardware embodiment and the task at hand, there is a clear benefit of having variable damping capabilities. We outlined how variable physical damping can be exploited through optimal control in the context of rapid movement tasks. The simulation and hardware experiments conducted indicate a significant benefit of (i) variable damping over fixed damping schemes and (ii) variable damping over variable stiffness when considering the relative speed at which the two can be modulated.

Next we built up on these results by addressing the optimal control problem of robotic systems with variable impedance actuation including switching dynamics and discontinuous state transitions. We developed and presented a systematic methodology for movement optimisation for such structures. The systems, facing tasks including switching dynamics and interaction with an environment, are approximately modelled as hybrid dynamical systems with time-based switching. We demonstrated the effectiveness of the proposed approach, emphasising the benefits of temporal and impedance modulation optimisation in brachiation. We presented both numerical evaluations and experimental results, for validation in a physical scenario, of brachiation tasks in a two-link brachiating robot with variable stiffness actuator developed in our laboratory.

We then extended the evaluation of our proposed approach to a monopod hopper system. This constitutes a significantly more challenging platform in terms of contacts (i.e., more complex impact map) and individual phase dynamics (i.e., existence of a flight phase). With an appropriate choice of the composite cost function to encode the task, we demonstrated the effectiveness of the proposed approach on the hopper system and showcased the robustness of the optimal feedback controller for a continuous hopping task.

Next, we extended our approach by incorporating adaptive learning, which allows for adjustments to the dynamics model, based on changes occurred to the system's behaviour, or when the behaviour cannot be fully captured by a rigid body dynamics formulation. We showed that the developed methodology can be successfully applied in the case of underactuated systems, such as the brachiation platform for a range of discrepancies, in both numerical simulations and in real hardware experiments.

FUTURE WORK

The focus of our investigation into variable damping was on a relatively simple task for a single link system. We have not addressed the role of damping in multi-link variable impedance control, nor have we explored the relative benefits of using variable damping versus variable stiffness in the context of energy-optimal behaviour. Future work on these aspects could reveal how the benefits of variable damping capabilities relate to the complexity of the system. This could potentially aid the design process of robotic systems (actuators could be equipped with different impedance capabilities depending on the scope of the system being developed).

Similarly, we applied our methodology for optimal control of robotic systems with variable impedance actuation including switching dynamics and discontinuous state transitions on two realistic systems. We demonstrated the benefit of simultaneous temporal and variable stiffness for a variety of tasks in numerical simulations and hardware implementation of a brachiating robot with VSA. In future work, we would like to apply this approach to more complex locomotion systems, such as bipedal or quadrupedal robotics systems with impedance modulation capabilities. As the current framework assumes a predefined phase sequence, this would require the development of a method for determining the best sequence for a given task, for systems where there are multiple candidates.

Furthermore, the increased complexity of such platforms would require a faster method for computing the desired control solutions, preferably on-line. Recently, OC was combined successfully with deep learning approaches [Levine, 2013; Watter et al., 2015], able transform the high dimensional state space to a significantly reduced one, where OC methods can be applied efficiently to complex systems. This methodology could provide a viable solution for extending the presented work to such systems.

Finally, in our adaptive learning approach we focused on learning the discrepancy between the real and assumed model of a system, for a range of mismatches. The method employed (in the form of the LWPR algorithm) is particularly suited for certain regression situations, such as non-linear function learning, in the presence of large amounts of data, with the requirement of incremental learning, (when the model might require continuous adaptation). The LWPR is limited to approximating smooth enough functions and the discrepancies introduced in our experiments fall within this category. In the hardware experiments we omitted the swing-up experiment due to a deficit of adequate sensor information, which could be rectified in future hardware embodiments.

As expected from theory, our investigation into transfer learning in the context of multi-phase movement shows that the use knowledge from previously available tasks can provide advantages in certain scenarios. It is of future interest to address this aspect in greater detail and evaluate the robustness and statistical significance of the results obtained. The impact map for the brachiation system is dependent only on the pre-impact state of the system, and thus unaffected by the dynamics discrepancy. In a different formulation, the impact map can dependent on the robot dynamics parameters (e.g., in the hopper system, it depends on the inertia matrix of the system). Introducing similar discrepancies in such a system would affect not only the individual phase dynamics but also the impact map, which would require learning. Due to the discontinuous nature of the impact map, LWPR would not be a suitable approach. A method for learning the inverse dynamics including contacts was presented in [Calandra et al., 2015]. By learning separate models for each contact position, multiple contact models can be activated, using a learned gating system. The performance of the method is due to an extensive set of reliable sensors. In future work, we would like to extend our methodology to include a similar approach for the contact dynamics. Handling non-smooth interactions in a safe and robust manner would provide an improved performance of complex locomotion systems on difficult terrains and human designed environments.

Part I

APPENDIX

APPENDIX: DAMPING IN VARIABLE STIFFNESS ACTUATORS

The one-link system in (3.9) can be regarded as a harmonic oscillator. In the absence of any damping, given fixed commands for the equilibrium position and stiffness the system displays a simple harmonic motion [Tipler and Mosca, 2007] (i.e., the position and velocity oscillate around the desired values, with constant amplitudes and frequencies - Fig. A.1, light blue line).

Newton's second law for damped harmonic oscillators is defined as:

$$M\ddot{q} + b\dot{q} + kq = 0 \Rightarrow \ddot{q} + 2\xi\omega_0\dot{q} + \omega_0^2q = 0, \quad (\text{A.1})$$

where $\omega_0 = (2\pi)/T = \sqrt{k/M}$ is the *undamped angular frequency of the oscillator* and $\xi = b/(2M\omega_0)$ is called the *damping ratio* (with k stiffness, b damping and T the period).

The value of the damping ratio ξ , determines the behaviour of the system. By adding variable damping capabilities to the variable stiffness system, ξ can be modified in order to achieve the desired effect, which can be:

- Overdamped response ($\xi > 1$): The system returns (exponentially decays) to equilibrium without oscillating. The decay speed is inversely proportional to ξ .
- Critically damped response ($\xi = 1$): The system returns to equilibrium in the shortest time possible without oscillating (which is often the desired scenario).
- Underdamped response ($0 < \xi < 1$): The system oscillates (with a slightly different frequency than the undamped case) with the amplitude gradually decreasing to zero.

Using the dynamics equation of a harmonic oscillator (A.1) and given a fixed set of values for M and k , the value for the critical damping can be obtained.

Fig. A.1 and A.2 showcase the behaviour of a nominal one-link system, comparing the no damping, critical damping, over, and under damping responses. The

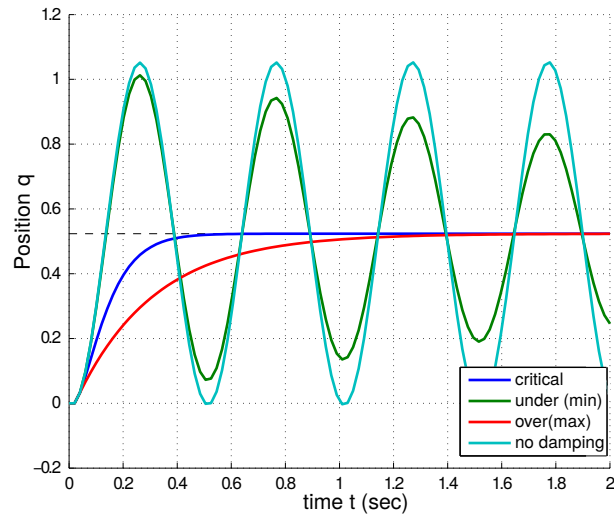


Figure A.1: Nominal one-link system. Comparison of position between no damping, critical damping, over and under damping.

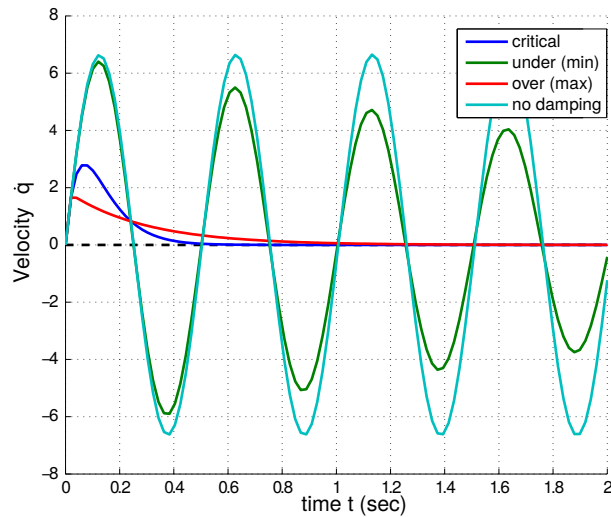


Figure A.2: Nominal one-link system. Comparison of velocity between no damping, critical damping, over and under damping.

values for the parameters of the system were selected (mainly for visualisation purpose) as $T = 0.125\text{sec}$ and $k = 0.9633\text{Nm/rad}$. The limit minimum and maximum value of the damping command (corresponding to the under- and over-damped limits) were chosen with respect to the value of the critical damping ($b_{\xi=1} = 0.1533\text{Nmsec/rad}$) such that the effects are observable.

APPENDIX: SYSTEM IDENTIFICATION AND ESTIMATION OF DAMPING FOR THE MACCEPA-VD

An approximation of the damping torque as a function of the equivalent circuit resistance R_e is given in (3.16). However, in our implementation of the MACCEPA-VD we employ PWM (pulse-width modulation) switching to emulate variation of the resistance. This permits the exploitation of the full damping range, but, it produces non-linearities in the relation between damping and the commanded duty cycle u_3 . This is due to various effects such as transients when switching between the open and closed configurations of the circuit. Hence, we use a data-driven approach for modelling the damping function $b = b(u_3)$.

For this we construct a test rig (displayed inside the inset of Fig. B.1) to collect data on the damping response of the motor used in the MACCEPA-VD.

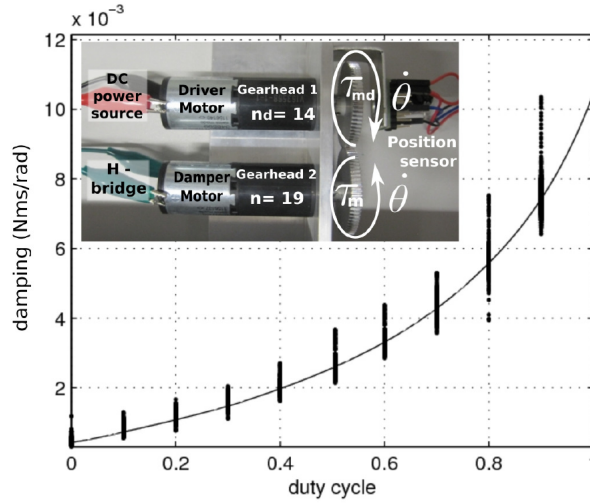


Figure B.1: Damping measurement rig (inset) and model fit.

Specifically, a DC motor (Maxon A-max 22/110125) is interconnected with the damper motor and used to drive the system as shown in Fig. B.1 (inset). The equation of motion of this system is given by:

$$M_m \ddot{\theta} + \underbrace{(b_{m,d} + b_m + b_e(u_3))}_{b(u_3)} \dot{\theta} = \underbrace{n_d k_{\tau d} I_d}_{\tau_{m,d}} \quad (\text{B.1})$$

where M_m is the moment of inertia of the interconnected motor-gearbox system, $b = b(u_3)$ is the total damping of the rig. This includes the mechanical damping of the two motors $b(0) = b_{md} + b_m$, when $u_3 = 0$, and the additional electrically generated damping $b_e(u_3)$, when $u_3 \in (0, 1]$. The terms on the right hand side of the equation represent the torques generated by the driver motor according to (3.14). The quantities: n_d , $\kappa_{\tau d}$ and I_d denote the gear ratio, torque constant, and the current of the driver motor, respectively.

For the experiments the driving motor is supplied with a constant voltage $V_d = 24$ V while the duty cycle of the PWM signal $u_3 \in \{0, 0.1, \dots, 1\}$ is modified by ten percent every 5s.

At each of these stages, the system quickly converges to a steady rotation speed ($\ddot{\theta} = 0$). We measure the rotational speed ($\dot{\theta} \neq 0$) and the current of the driver motor I_d . Using this data, we can compute the total damping: $b_i(u_3) = n_d \kappa_{\tau d} I_{di} / \dot{\theta}_i$ (based on (B.1) with $M_m \ddot{\theta} = 0$), for each data point $i \in \{1, 2, \dots, N\}$ for a given u_3 .

From these values the total mechanical damping of the system is estimated as a mean of the computed damping constants at $u_3 = 0$: $\hat{b}(0) = \frac{1}{N} \sum_1^N b_i(0)$. The latter is then used to estimate the electrical damping effect on the remaining data: $\hat{b}_{ei}(u_3) = b_i(u_3) - \hat{b}(0)$. Finally, using the quadratic scaling of the damping constant with the gear ratio (3.16), we define $\hat{b}_m = n^2 / (n^2 + n_d^2) \hat{b}(0)$. The latter is used to estimate the total damping coefficient as:

$$\hat{b}_i(u_3) = \hat{b}_m + \hat{b}_{ei}(u_3). \quad (\text{B.2})$$

Using the estimates from (B.2) as data, we fit a model with normalised Gaussian radial basis functions ¹ $\phi(u_3) \in \mathbf{R}^M$:

$$\hat{b}(u_3) = \mathbf{w}^T \phi(u_3), \quad (\text{B.3})$$

where $\mathbf{w} \in \mathbf{R}^M$ is a vector of weights. In Fig. B.1 we plot the damping as a function of duty cycle u_3 obtained through this process.

Finally, re-inserting the damper motor into the MACCEPA-VD, the joint damping torque is estimated as:

$$\hat{\tau}_b(\dot{q}, u_3) = -n_q^2 \hat{b}(u_3) \dot{q}, \quad (\text{B.4})$$

where $n_q = 1.8$ is the gear reduction between the link and the motor shaft.

¹ where $\phi_\mu = K_\mu(u_3) / \sum_{\nu=1}^M K_\nu(u_3)$, $K_\mu = \exp(-(u_3 - c_\mu)^2 / 2\sigma^2)$ and c_μ , $\mu = 1 \dots M$

APPENDIX: DESIGN AND SELECTION OF THE COST FUNCTION FOR THE BRACHIATION TASK

In this appendix we provide a brief investigation conducted by our collaborator, Dr. Nakanishi, into the possible options in defining a viable cost function for the brachiation task addressed in Sec. 4.7.3.

In OC, the usual approach is to encode an abstract representation of the task using a cost function. The design of the cost function is one of the most important and difficult parts for a successful application of the OC framework. For simple task and plant dynamics, an intuitive choice (typically a quadratic cost in the state and control as in an LQR setting) would suffice¹. For a highly dynamic task and complex plant dynamics, this becomes increasingly difficult and finding an appropriate choice of the cost terms which best encode the task remains an open issue.

In this section, we explore additional candidates for the cost function, in the context of the brachiation task described in Sec. 4.7.3. Alongside the cost function in (4.42), we consider the following prospective running cost functions $r = r(\mathbf{x}, \mathbf{u})$:

- quadratic cost in the control command (servo motor position command):

$$r = \mathbf{u}^T \mathbf{R} \mathbf{u}, \quad (\text{C.1})$$

- quadratic cost in the joint torque. The main term is the cost associated with the joint torque τ and $\mathbf{u}^T \mathbf{R}_1 \mathbf{u}$ is added for regularisation (small \mathbf{R}_1):

$$r = \mathbf{u}^T \mathbf{R}_1 \mathbf{u} + R_2 \tau^2. \quad (\text{C.2})$$

Fig. C.1 shows the results obtained using the running cost (C.1) with $\mathbf{R} = \text{diag}\{1, 1\}$. The computed optimal movement duration is $T = 0.604$ (sec). Fig. C.2 shows the results using the running cost (C.2) with $\mathbf{R}_1 = \text{diag}\{10^{-4}, 10^{-4}\}$ and $R_2 = 100$. The obtained optimal movement duration is $T = 0.620$ s. In both of these two cases, the same terminal cost parameters are used as in (4.42).

¹ The adjustment of the weights would still be required.

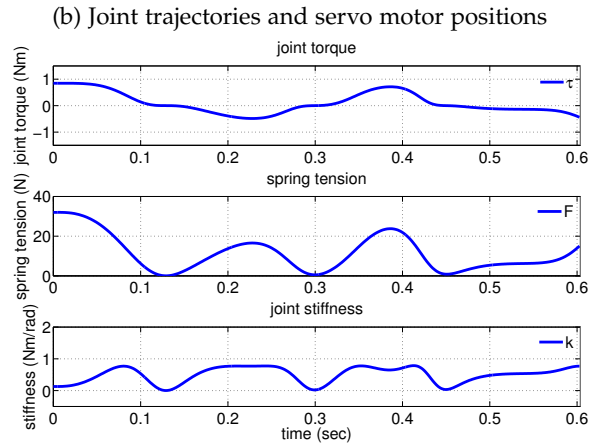
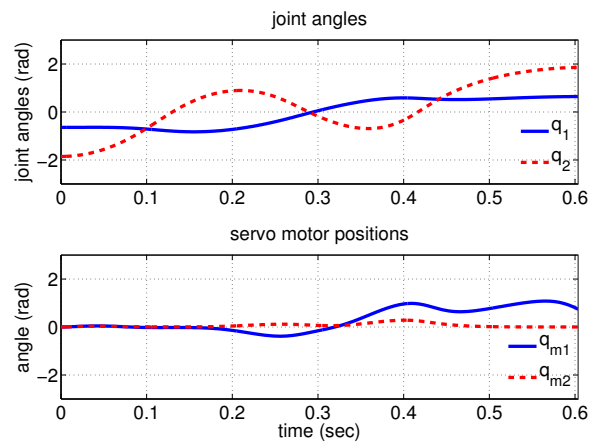
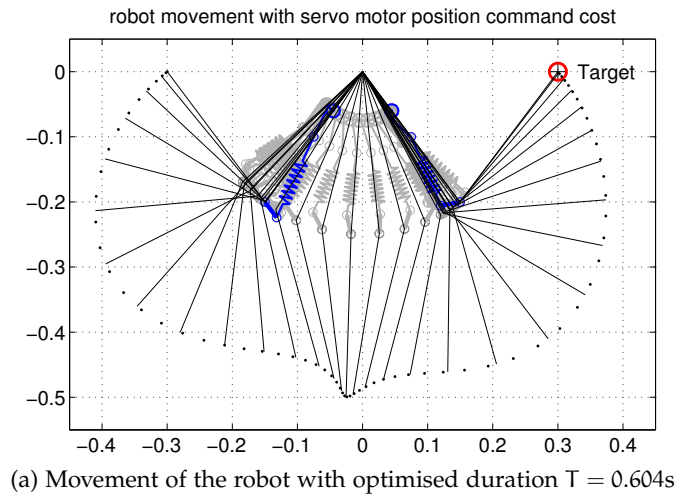
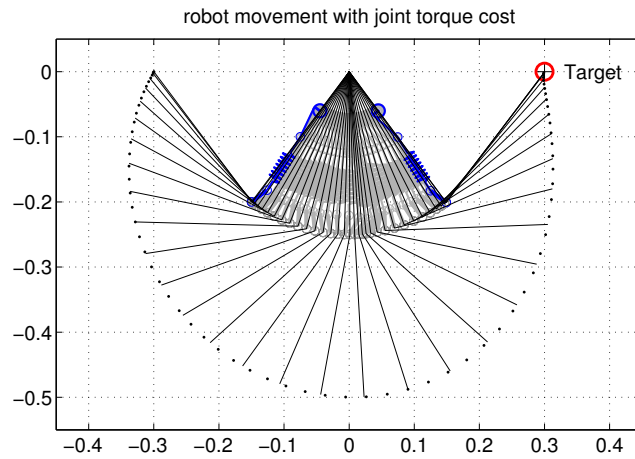
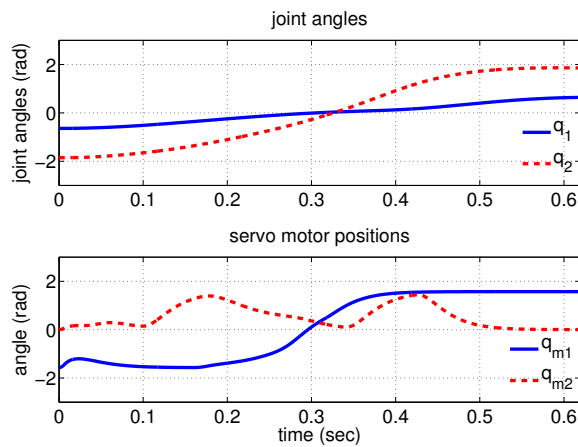


Figure C.1: Optimisation of the locomotion task using the cost (C.1).

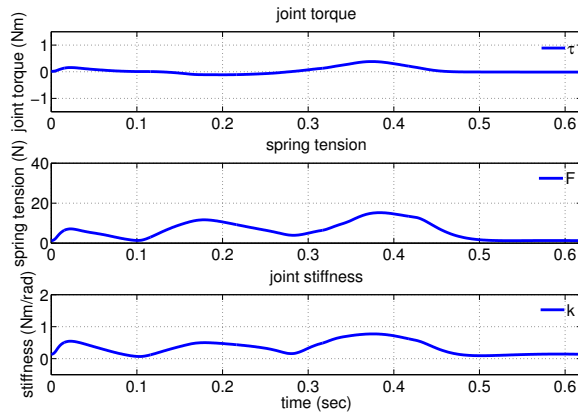
As demonstrated in Figs. C.1 and C.2, the robot is able to successfully reach the target bar by minimising each specific cost in addition to that defined in (4.42). Depending on the choice of running cost (C.1), significant difference in the resultant robot movement and much higher spring tension and joint stiffness can be found (Fig. C.1). As observed in Fig. C.2, when using the cost associated with the joint



(a) Movement of the robot with optimised duration $T = 0.620s$



(b) Joint trajectories and servo motor positions



(c) Joint torque, spring tension and joint stiffness

Figure C.2: Optimisation of the locomotion task using the cost (C.2).

torque (C.2), the resultant movement looks almost identical to the one obtained for the cost (4.42) (Fig. 4.4) and the joint torque profiles are comparable. However, we can observe that the spring tension and joint stiffness are larger than those for (4.42). This is due to the redundancy in the variable stiffness actuation, as the

results depend on how this is resolved by an appropriate choice of cost function. These results suggest that the selection of the cost function is crucial, however, the process is still non-intuitive.

We note that other options for cost functions are possible (e.g., energy consumption). In the brachiation task, in the absence of friction, the *mechanical* energy of the rigid body dynamics ($E = \int_0^T \tau \dot{q}_2 dt$) is conserved for the swing locomotion with the same intervals at the same height, starting and ending at zero velocity (if no potential energy is stored in the spring of the VSA at the end of the swing). Thus, in order to consider true energy consumption, it would be necessary to evaluate the *electrical* energy consumed at the motor level. However, this is not straightforward, since it requires a precise model of the mechanical and electrical motor dynamics. This includes all the associated factors such as motor efficiency and transmission loss, which can be rather complex to model in practice, and the control strategy would largely depend on the properties of the motors used.

APPENDIX: HARDWARE DESIGN DETAILS

In order to validate our method on a real world problem the construction of an adequate hardware platform was required. In this section we give a brief overview of the hardware design evolution.

D.1 INITIAL DESIGN

The first version of the robot configuration is displayed in Fig. D.1.

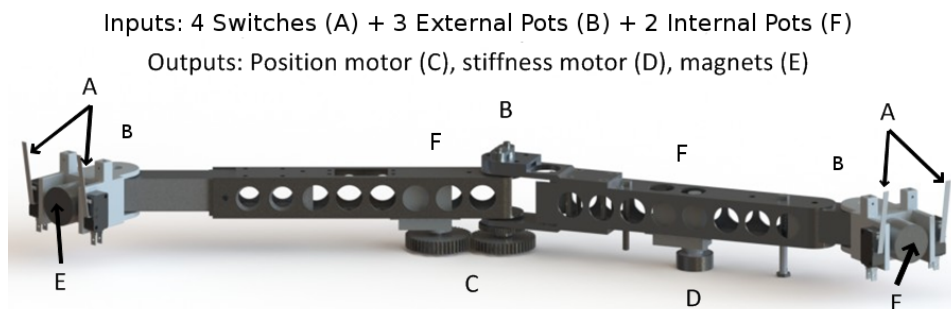


Figure D.1: The two link brachiation robot (Ver. 1.0)

The design makes use of a previously available mechanism, a multiple link throwing arm [Braun et al., 2012b]. For the purpose of our implementation this device was stripped down to a two link system, which was used as the core of the design.

We decided on building a system that will achieve brachiation on a continuous metallic surface, using magnets to attach itself to it. Thus, the hardware platform consists of two main links, with magnetic gripper mechanisms at the end. Each gripper consists of an additional link on which an electromagnet is mounted (Magnet-Schultz, GMHX030X00D02, Fig. D.1 - E).

These unactuated gripper links are connected via a string mechanism, which consists of two thin, nonelastic strings (depicted in black and blue in Fig. D.2), running along the length of the robot and cross-connecting the edges of the grippers. The strings are passing through the centre of rotation of the main link. The operating principle is that, this way, the angle of the free (swinging gripper) matches that

of the holding (fixed gripper). At the moment of contact the magnet is aligned with the brachiation surface, thus achieving a successful hold. Two switches mounted on each side of the magnet, along the brachiation direction (Fig. D.1 - A), provide additional information about the gripper's position and help the tuning of the passive mechanism. One of the limitations of this design is that brachiation is limited to same level surfaces.



Figure D.2: Depiction of the passive string mechanism, devised to ensure surface alignment at surface contact.

The two main links form an actuated joint that uses the mechanically adjustable series elastic actuators (MACCEPA) introduced in [Ham et al., 2007]. The MACCEPA actuator is designed to give simultaneous equilibrium position and joint stiffness control through use of two independently controlled servomotors.

The original design from [Braun et al., 2012b] was modified by upgrading the servos to the high speed HITEC HSR-7940TH model. The upgrade was dictated by the inability of the old motors to cope with the new load in successfully achieving a brachiation movement. The joint angles (including the unactuated links) are measured using by rotary potentiometers (Alps RDC503013A), which can cover an effective variable range of 320° . Information on the motor positions is given by direct access to the internal potentiometers of the servo motors.

The first implementation suffered from very low performance. The main issue encountered was that the mechanism required a perfect magnet-to-surface alignment at the point of contact, in order to operate successfully. A large discrepancy between the planned final position and the achieved one was observed, caused by a sliding behaviour of the gripper on the metallic surface towards the end of

the movement. Another difficulty was represented by the low repeatability rate of each movement.

Motivated by these discoveries, several solutions were investigated, such as i) replacing the one magnet gripper with a dual magnet one and ii) making the swinging arm's gripper compliant. These would both involve a significant re-design, without guarantee of success. After careful considerations a decision was made to adopt a different approach iii) to redesign the system, by replacing the magnetic end-effectors with grippers.

D.2 FINAL DESIGN

The final design of the brachiation robot can be observed in Fig. D.3 (equivalent to Fig. 4.8). It consists of the initial main link equipped with a MACCEPA actuator. The end-effectors are designed as two mechanical gripper actuated by servo motors. The grippers were design using SolidWorks 2013 (Student Edition) and manufactured using 3D printing technology, alongside with several connecting elements, required for attaching the grippers to the main link.

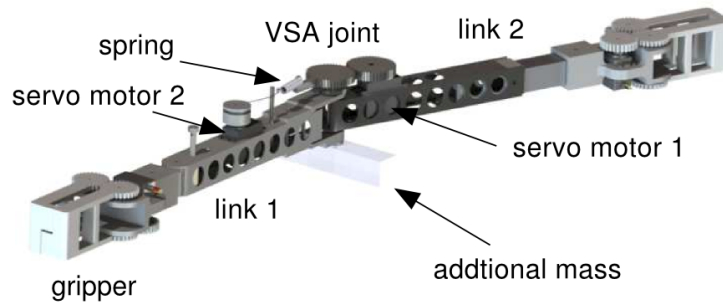


Figure D.3: The two link brachiation robot (Ver. 2.0)

As before, the elbow joint is actuated by a MACCEPA variable stiffness actuator [Ham et al., 2007] having two servo motors (Hitec HS-7940TH). Each link is equipped with a gripper driven by a single servo motor (Hitec HSR-5990TG) to open and close it through a gear mechanism. Since the new configuration eliminates the need for articulated joints at the end-effectors, the potentiometers associated with them were removed. Being left with only one active potentiometer (Alps RDC503013A, located at the elbow joint) the system required a new method of obtaining its absolute position during locomotion. Two IMU units (InvenSense MPU-6050) were placed on each link, allowing the measuring of the angle of the first link by integrating its angular velocity obtained from the IMU. The servo mo-

tor positions are measured through direct access to its internal potentiometer. The operating frequency of control and measurement is 1kHz.

Under the correct mass distribution a brachiation system should be able to achieve a successful movement with little effort, as various studies of biological systems adapted to this form of locomotion have indicated [Michilsens et al., 2009]. Our numerical exploration showed that with an inadequate mass distribution, it was difficult to find an optimal solution in achieving desired swing locomotion behaviour under the constraint.

The option of replacing the motors with similarly sized ones, but with a higher torque output was not available. Therefore, we decided on adjusting the mass distribution, in order to facilitate the movement by lowering the value of the additional torque the motors will have to provide to successfully reach the desired targets. For this reason, the mass distribution of the robot was empirically adjusted to resemble the desirable natural dynamics required for the task. We performed data collection from the hardware system, under various command input sets and compared the observed behaviour with those anticipated from the simulation. We used a similar approach to tune the friction parameters and the value of the bandwidth parameters in the motor dynamics model, α . (Table 4.1).

APPENDIX: MODELLING THE HARDWARE: PARAMETER FITTING

The hardware design detailed in Appendix D is modelled using SolidWorks 2013 [Lombard, 2013]. The software provides estimates of the inertial parameters of the components, based on the geometry of the parts and the density of the composing material. We use these estimates for our rigid body dynamics model. In order to ensure the accuracy of the model we performed a validation of the estimated parameters.

For this purpose we devise a series of experiments aimed at testing the behaviour of the system under various excitations. This procedure verifies that the parameters obtained provide an accurate model for planning, in order to obtain a corresponding performance on the hardware.

E.1 NATURAL DYNAMICS

The first set of experiments investigates the behaviour of the natural dynamics of the system. We perform the validation on each link individually. For this we disconnect the actuating motors and fix one of the links. We then let the free link swing under gravitational pull from various starting positions.

Fig. E.1 displays the behaviour of each of the two links on one set of these trajectories. The black and blue lines represent data collected from the hardware, from the free link's gyroscope sensor and the middle joint potentiometer, respectively. The oscillations of the hardware match the simulated behaviour (red lines), generated using the rigid body dynamics based on the estimated parameters.

E.2 ACTUATED DYNAMICS

In the second set of experiments we reconnect the actuating motors of the main link. We start by applying sinusoidal position commands for a fixed stiffness profile (i.e., fixed stiffness command), while keeping one of the links fixed, as in the previous section. We perform this experiment for various sinusoidal position com-

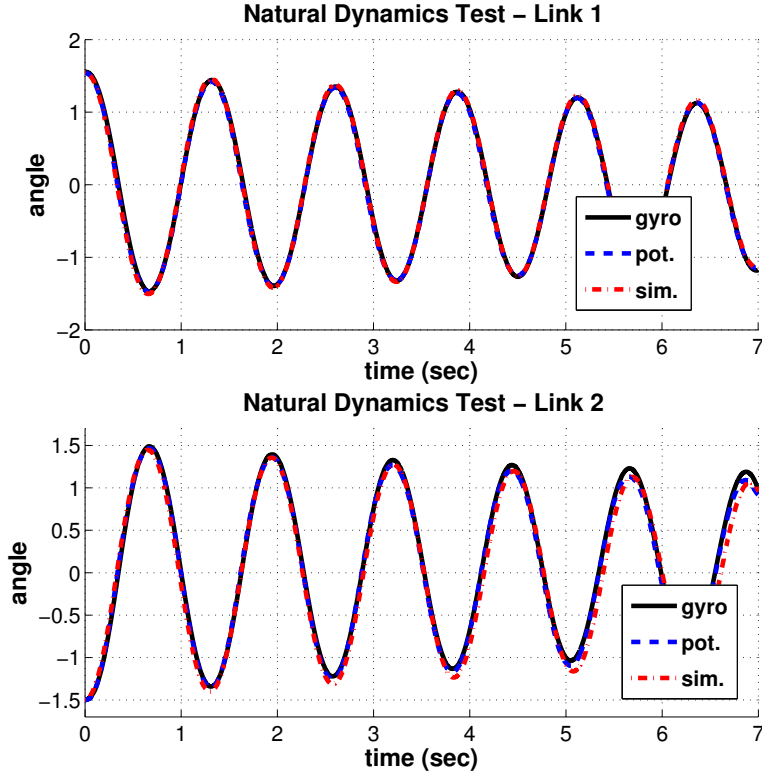


Figure E.1: Validation of the estimated parameters. Natural dynamics of each individual link.

mand waves (i.e., $A \in \{\pi/4, \pi/2\}$ rad, $f \in \{0.70, 0.85, 1.00\}$ Hz) and low, medium and high stiffness command levels (i.e., $u_2 = \{0, \pi/4, \pi/2\}$ rad).

Fig. E.2 displays the behaviour of each of the two links on one set of these trajectories. As before, the blue and black lines represent data collect from the hardware, from the middle joint potentiometer and the free link's gyroscope sensors, respectively. The oscillations of the hardware match the simulated behaviour (red lines), generated using the rigid body dynamics based on the estimated parameters, with the same input commands.

We know that, for this particular embodiment of the system dynamics, the optimal solutions for swing-up and brachiation tasks consist of mostly square wave commands (similar to a bang-bang controller behaviour). Thus, we also test the response of the system under such conditions.

Similarly to the sinusoidal excitation experiments described above, we analyse the behaviour of the individual links for a fixed stiffness profile (i.e., fixed low, medium and high stiffness commands, $u_2 = \{0, \pi/4, \pi/2\}$ rad), with a square wave command input ($A \in \{\pi/4, \pi/2\}$ rad, $f \in \{0.70\}$ Hz).

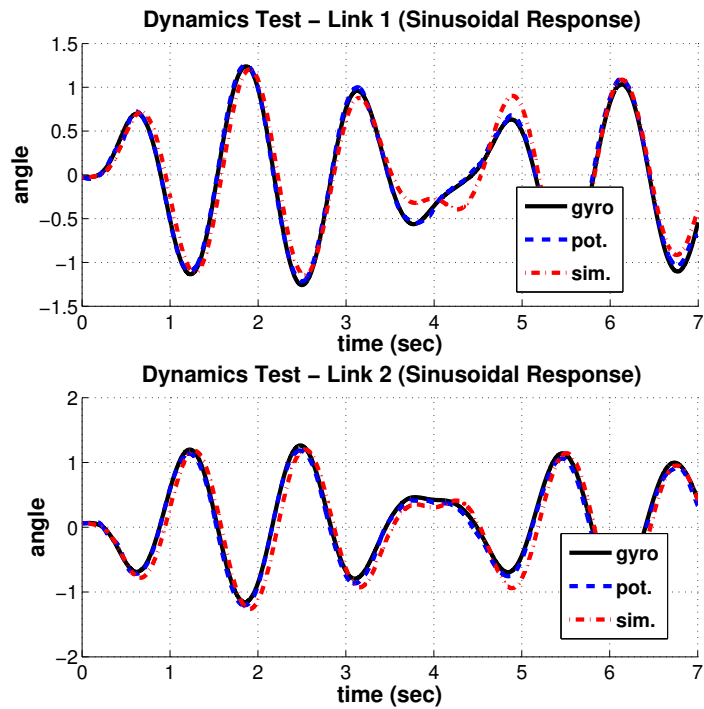


Figure E.2: Validation of the estimated parameters. Dynamics of each individual link under sinusoidal position commands with fixed stiffness command.

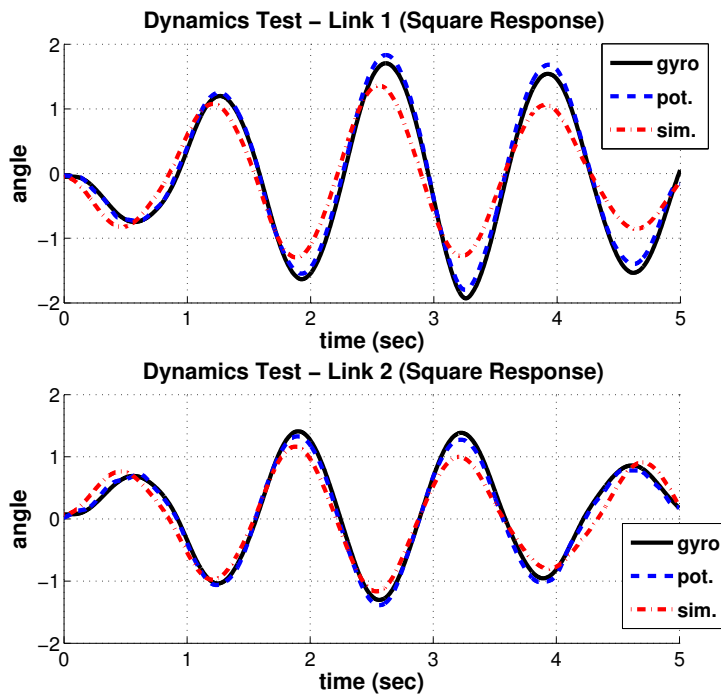


Figure E.3: Validation of the estimated parameters. Dynamics of each individual link under sinusoidal position commands with fixed stiffness command.

Fig. E.3 displays the behaviour of each of the two links on one set of these trajectories. As in the previous tests, the recorded hardware trajectories provide a good match to the simulated behaviour (red lines)¹.

We then proceed with extending this approach on the dynamics of the whole two-link system. Starting from a vertical, one hand-hold configuration, we observe the behaviour under sinusoidal and square wave excitations (for the position command), with a fixed stiffness profile (fixed stiffness command).

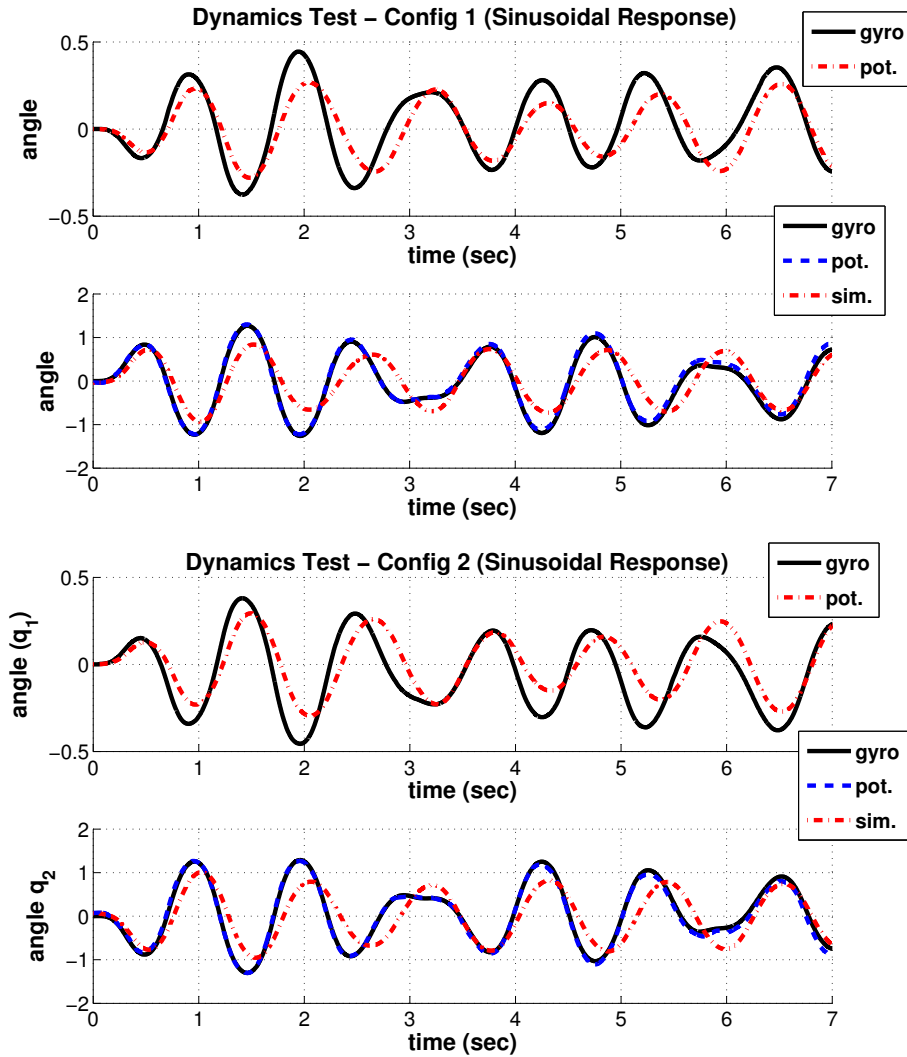


Figure E.4: Validation of the estimated parameters. Dynamics of the two-link system under sinusoidal position commands with fixed stiffness command.

¹ The slight discrepancy that can be observed in this case is caused by the presence of elastic components and non-deterministic effects, such as cable friction. These elements cannot be fully captured by standard rigid body dynamics equations.

Fig. E.4 shows the behaviour of the two link system for two sinusoidal trajectories. Fig. E.5 displays the corresponding behaviour of the two link system (both hand-hold configurations) for two square wave position command scenarios.

The sensor data from the hardware provides us with information on the position of the first link (q_1 , black line), by integrating the gyroscope readings. The position of the second link (q_2) is observed from the middle joint potentiometer and the free link's gyroscope sensors, blue and black lines, respectively.

As it can be observed, the shape of the movement is preserved, even if the model match is less accurate than in the individual link scenarios (Fig. E.4). The behaviour of the system under the bang-bang controller scenario (Fig. E.5) provides a good match to the simulated behaviour. Based on these experiments we are satisfied that the set of parameters obtained provide an good estimation of the true dynamics, allowing for accurate planning for our desired tasks.

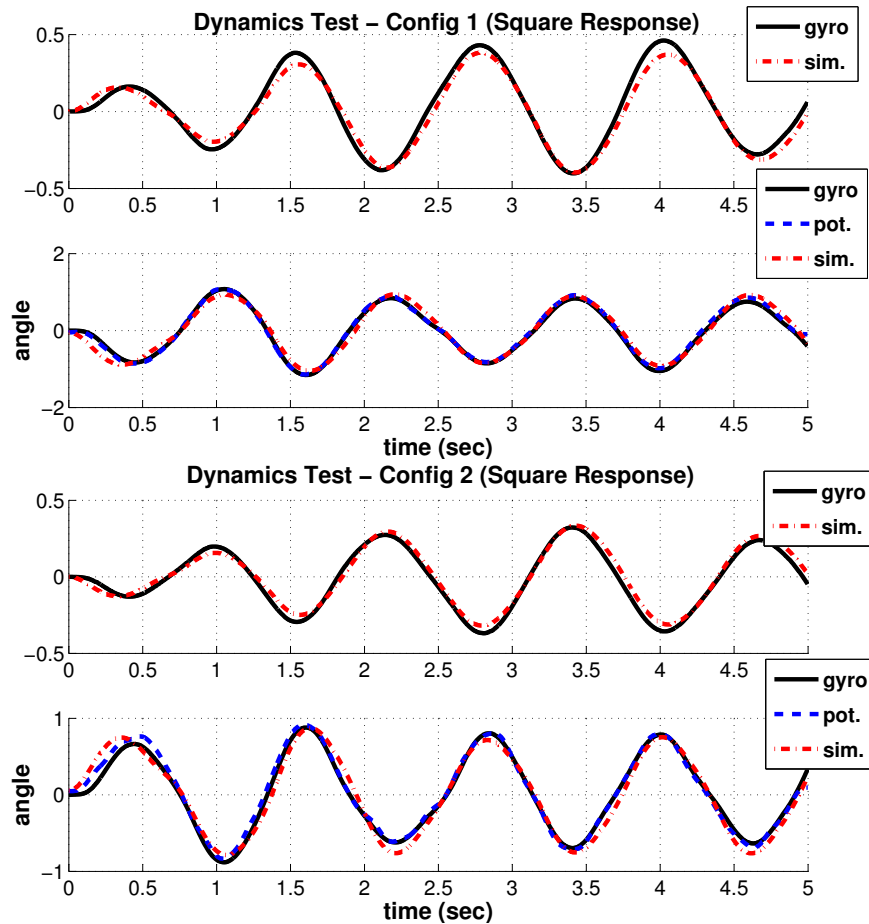


Figure E.5: Validation of the estimated parameters. Dynamics of the two-link system under square wave position commands with fixed stiffness command.

E.3 MOTOR DYNAMICS

In order for the behaviour of the simulated system to be consistent with the hardware response, we also tune the value of the bandwidth parameters in the motor dynamics model (4.7), α .

We use the response of the motors to a square wave to ensure α is able to approximate the speed of the dynamics ². Fig. E.6 shows the response of the two motors (qm_1 and qm_2 , grey lines ³) to a square wave command input (blue lines). The theoretical behaviour of the motors is obtained from the analytic form given in (4.7) using the displayed values for $\alpha = [20, 25]$. The parameters used provide a good match for the hardware behaviour.

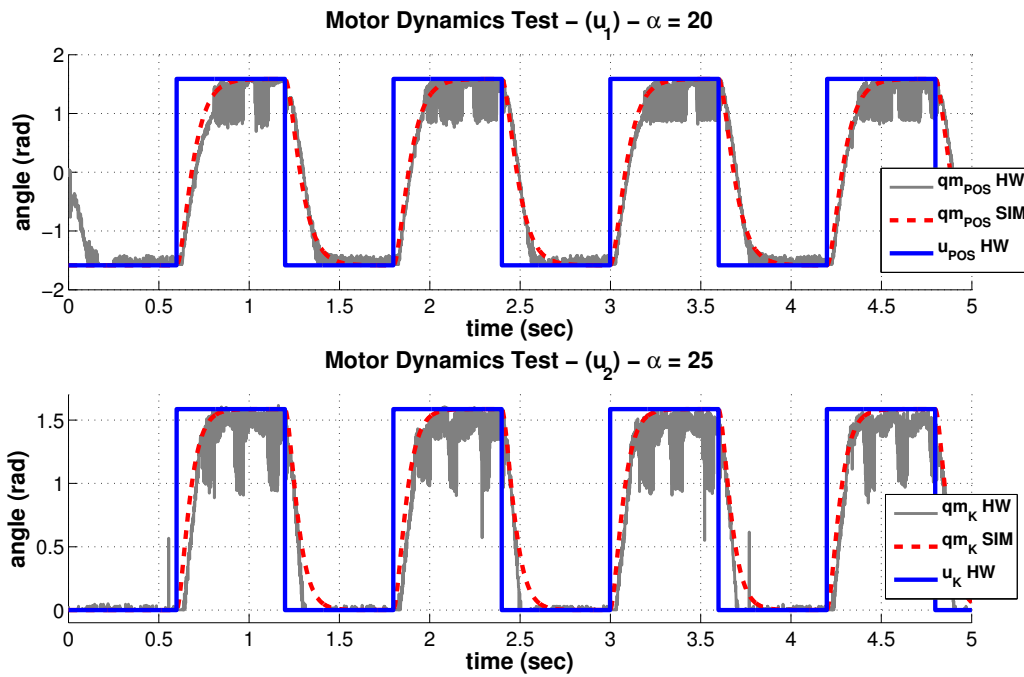


Figure E.6: Validation of the motor dynamics parameters. Motors response to a square wave input.

² This represents the scenario in which the response of the motor under load is tested at the limit.

³ The motor position values are obtained from the internal potentiometers' readings, which are presented here without filtering.

APPENDIX: LWPR SET-UP DETAILS

F.1 PARAMETER SELECTION

The LWPR algorithm is a robust method for incremental nonlinear function approximation in high-dimensional spaces. Despite its proven nice properties and efficiency [Vijayakumar et al., 2005], in practice, applying the algorithm to a specific learning task can be quite involved.

This requires the tuning of parameters to the specific task at hand. One of the most important such parameters is init_D , the distance metric assigned initially to newly created local models, which is also tightly connected to the normalisation of the input data (norm_{in}) [Klanke et al., 2008]. The latter is used in normalisation for making the inputs dimensionless. As recommended in [Klanke et al., 2008] we set these based on the expected range of possible inputs in each dimension. Two other parameters affecting the performance of the method are w_{gen} , the threshold for generating a new local model and w_{prune} , the threshold for eliminating one model when an overlap is detected.

Next, we analyse the effect of selecting a init_D , w_{gen} and w_{prune} values for learning the model discrepancy in a brachiation task. For this we create a testing scenario in which a solution is generated for a single swing task using on the assumed model. A set of 10 perturbed versions of the obtained command solution is generated, which will be used as a test set.

The original command and a different set of 10 perturbed versions are applied to the true dynamics. A LWPR model is trained incrementally on the data collected from these experiments. At each iteration, after the data from the current command version has been incorporated, we test the performance of the LWPR model on the test set.

Based on the obtained results we opted for $\text{init}_D = 25$, $w_{gen} = 0.2$ and $w_{prune} = 0.8$. Fig. F.1 shows the evolution of the nMSE and the number of local model allocated during the procedure described above, for several values of init_D for a set of $w_{gen} = 0.2$ and $w_{prune} = 0.8$.

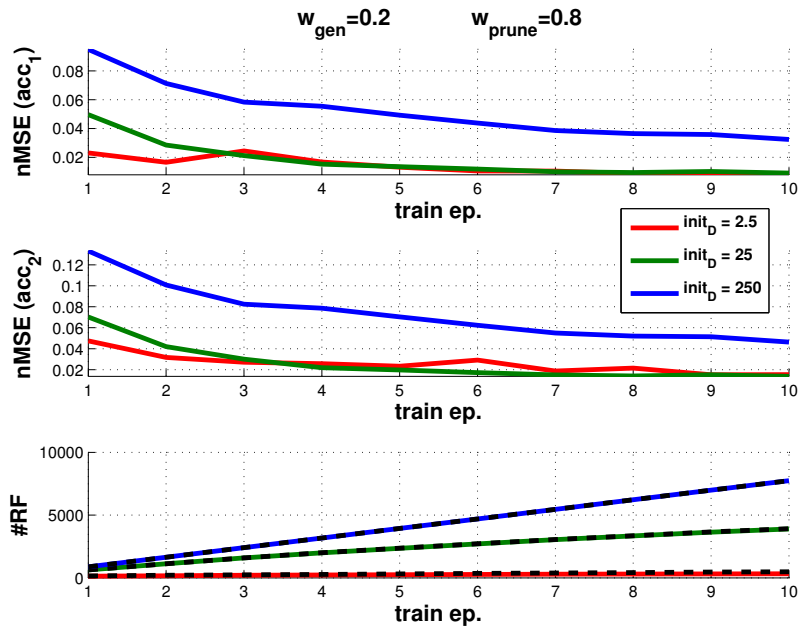


Figure F.1: Comparison of the effect of various init_D . Error (nMSE) on the training set (in the two joint positions and number of RF plotted against the number of training sets seen).

We tested this evolution for a wide range of combinations for these thresholds ($w_{\text{gen}} \in [0.1, 0.5]$ and $w_{\text{prune}} \in [0.6, 0.9]$), and the result displayed is representative of the overall behaviour. For comparison we also display the effect of using various values for w_{gen} and w_{prune} in Fig. F.2 and F.3, respectively, while keeping the other parameters to the same values.

A full detail of the parameters used for the initialisation of the LWPR model is given in the code below.

```

%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%% Initialise LWPR model %%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%%
model_lwpr = lwpr_init(10,2,'name','lwpr_brachiator');
model_lwpr = lwpr_set(model_lwpr,'norm_in',[1; 1; 5; 5; 1; 1; 10; 10; 1;
    1]);
model_lwpr = lwpr_set(model_lwpr,'init_D',eye(10)*25);
model_lwpr = lwpr_set(model_lwpr,'init_alpha',ones(10)*0.250);
model_lwpr = lwpr_set(model_lwpr,'diag_only',0);
model_lwpr = lwpr_set(model_lwpr,'w_gen',0.2);
model_lwpr = lwpr_set(model_lwpr,'w_prune',0.8);
model_lwpr = lwpr_set(model_lwpr,'meta',1);
model_lwpr = lwpr_set(model_lwpr,'meta_rate',250);
model_lwpr = lwpr_set(model_lwpr,'update_D',1);
model_lwpr = lwpr_set(model_lwpr,'kernel','Gaussian');

```

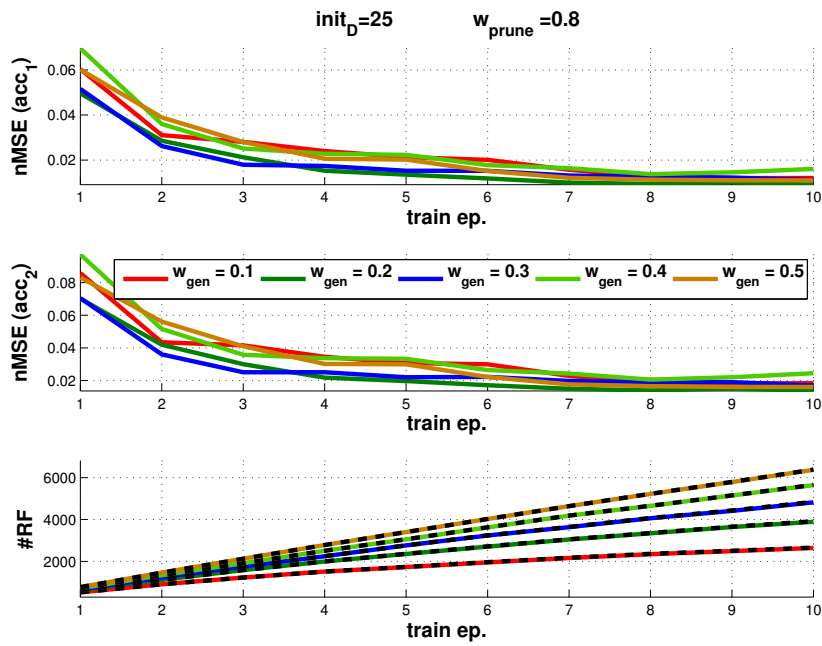


Figure F.2: Comparison of the effect of various w_{gen} . Error (nMSE) on the training set (in the two joint positions and number of RF plotted against the number of training sets seen.

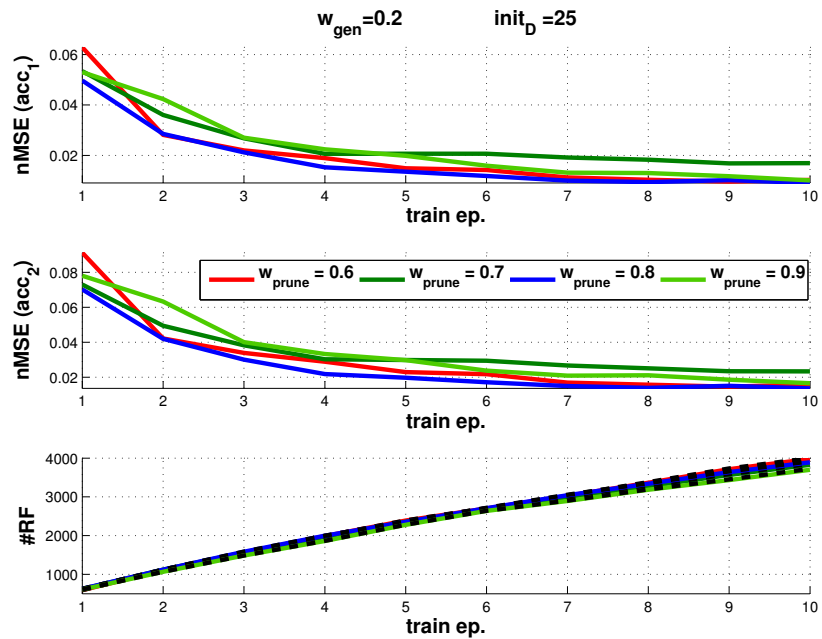


Figure F.3: Comparison of the effect of various w_{prune} . Error (nMSE) on the training set (in the two joint positions and number of RF plotted against the number of training sets seen.

F.2 NATURE OF THE DISCREPANCY

We note that the method employed (in the form of the LWPR algorithm) is particularly suited for certain regression situations, such as non-linear function learning, in the presence of large amounts of data, with the requirement of incremental learning, when the model might require adaptation. The LWPR is limited to approximating smooth enough functions and the discrepancies introduced in our experiments fall within this category. Fig. F.4 show the nature of this error at each state along a trajectory¹. The trajectory was generated by running a set of commands on the true dynamics model. The assumed model is queried at each step using these *true* states, giving us the *assumed* states. The error between the two dynamics (i.e., the quantity we are trying to model) is obtained as the difference between these two states.

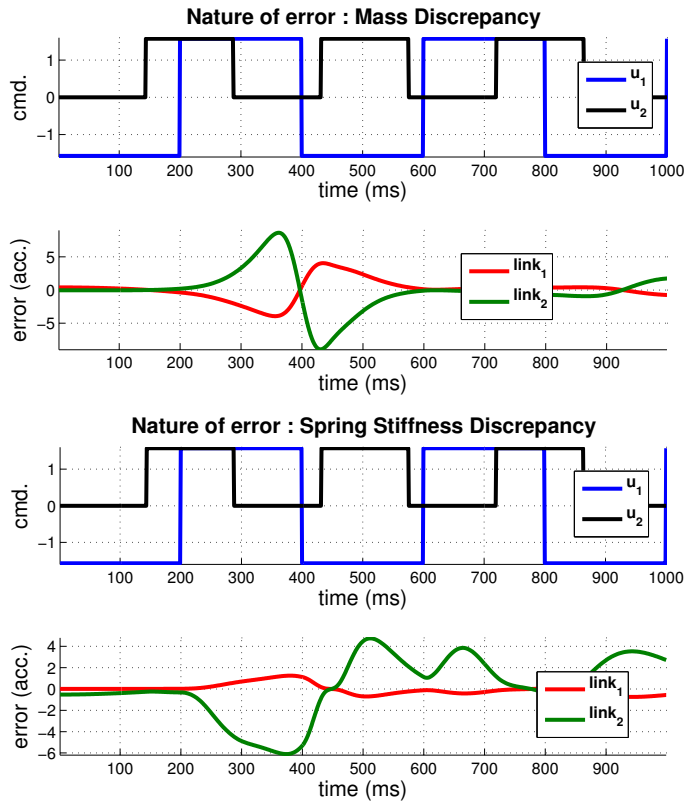


Figure F.4: Nature of the dynamics error along a trajectory for i) mass distribution and ii) spring stiffness discrepancies, respectively.

¹ The figure depicts only one example of the dynamics error, along a specific command trajectory, in our experiments we have investigated the nature of this discrepancy under several command scenarios and found the results consistent with those displayed here.

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